



SBOS499C – JANUARY 2011 – REVISED APRIL 2011

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# Gated-Class H, Dual-Port VDSL2 Line Driver

Check for Samples: THS6226

### **FEATURES**

- Digitally-Adjustable Quiescent Current: 7.6mA to 23.0mA
- 1.0mA Bias Current Step
- Independent Voltage Boost and Main Line
   Driver Disable
- Low-Power Line Termination Mode
- Full Capacitor Recharge: 3ms
- Low Input Voltage Noise Density: 6.3 nV/\/Hz Input-Referred Voltage Noise
- Low MTPR Distortion: 70dB with +19.8dBm G.993.2—Profile 8b
- –91dBc HD3 (1MHz, 60Ω Differential)
- High Output Current: (383mA into 60Ω)
- Wide Output Swing: 40V<sub>PP</sub> (+12V, 60Ω Differential Load with a 1:1.4 Transformer)
- Wide Bandwidth: 125MHz
- Port-to-Port Separation of 90dB at 1MHz
- PSRR: 70dB at 1MHz for Good Isolation

## **APPLICATIONS**

- Ideal for All VDSL2 Profiles
- Backwards-Compatible with ADSL/ADSL2+/ADSL2++ Systems

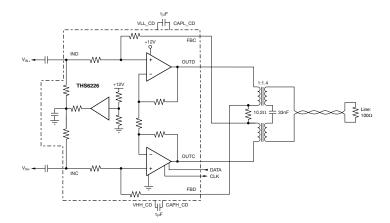
### DESCRIPTION

The THS6226 is а dual-port, class Η. current-feedback architecture, differential line driver amplifier system ideal for xDSL systems. The device is targeted for use in very-high-bit-rate digital subscriber line 2 (VDSL2) line driver systems that enable native DTM signals while supporting greater than +20.5dBm line power (up to 8.5MHz) with good linearity, supporting the G.993.2 VDSL2 8b profile. It also fast enough to support central-office is transmission of +14.5dBm line power up to 30MHz.

The unique architecture of the THS6226 allows quiescent current to be minimal while still achieving very high linearity. Differential distortion, under full bias conditions, is –91dBc at 1MHz and reduces to only –75dBc at 5MHz. Fixed multiple bias settings of the amplifiers offer enhanced power savings for line lengths where the full performance of the amplifier is not required. To allow for even more flexibility and power savings on all profiles, quiescent current is digitally adjustable from 7.6mA to 23mA with a bias current step of 1.0mA. For systems where additional power savings while not transmitting are desired, the THS6226 can be used in its line termination mode to maintain impedance matching.

The wide output swing on +12V power supplies, coupled with excellent current drive, allows for wide dynamic headroom, keeping distortion minimal.

The THS6226 is available in a QFN-32 PowerPAD™ package.



### Typical VDSL2 Line Driver Circuit Using One Port of the THS6226

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This integrated circuit can be damaged by ESD. Texas Instruments recommends that all integrated circuits be handled with appropriate precautions. Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

	ORDERING INFORMATION "						
PACKAGE PRODUCT <sup>(2)</sup> PACKAGE-LEAD PACKAGE DESIGNATOR PACKAGE MARKING TRANSPORT MEDIA, QUANTITY							
THS6226IRHBT		DUD	THS6226IRHB	Tape and Reel, 250			
THS6226IRHBR	VQFN-32	RHB	10302201800	Tape and Reel, 3000			

### ODDEDING INFORMATION(1)

For the most current package and ordering information, see the Package Option Addendum at the end of this document, or visit the (1)device product folder at www.ti.com.

The PowerPAD is electrically isolated from all other pins. (2)

### ABSOLUTE MAXIMUM RATINGS<sup>(1)</sup>

Over operating free-air temperature range, unless otherwise noted.

	PARAMETER	THS6226	UNIT
Supply vo	Itage, GND to $V_{S+}$ , class AB only	15	V
Supply vo	Itage, GND to $V_{S+}$ , class H only	12.5	V
Input volta	age, V <sub>I</sub>	15	V
Output cu	rrent, I <sub>O</sub> : static dc <sup>(2)</sup>	±100	mA
Continuou	us power dissipation	See Thermal Information table	
Normal storage temperature		-40 to +85	°C
Maximum junction temperature, any condition, T <sub>J</sub> <sup>(3)</sup>		+150	°C
Maximum junction temperature, continuous operation, long-term reliability, T <sub>J</sub> <sup>(4)</sup>		+130	°C
Storage te	emperature range, T <sub>STG</sub>	-65 to +150	°C
Human body model (HBM)		2000	V
ESD ratings:	Charged device model (CDM)	500	V
raango.	Machine model (MM)	100	V

- Stresses above those listed under absolute maximum ratings may cause permanent damage to the device. These are stress ratings (1)only, and functional operation of the device at these or any other conditions beyond those indicated under recommended operating conditions is not implied. Exposure to absolute maximum rated conditions for extended periods may degrade device reliability.
- The THS6226 incorporates a PowerPAD on the underside of the chip. This acts as a heatsink and must be connected to a thermally (2)dissipating plane for proper power dissipation. Failure to do so may result in exceeding the maximum junction temperature which could permanently damage the device. See TI Technical Brief SLMA002 for more information about utilizing the PowerPAD thermally-enhanced package. Under high-frequency ac operation (> 10kHz), the short-term output current capability is much greater than the continuous dc output current rating. This short-term output current rating is about 8.5x the dc capability, or approximately +850mA
- The absolute maximum junction temperature under any condition is limited by the constraints of the silicon process.
- The absolute maximum junction temperature for continuous operation is limited by the package constraints. Operation above this temperature may result in reduced reliability and/or lifetime of the device.

### THERMAL INFORMATION

	THERMAL METRIC <sup>(1)</sup>	RHB	UNITS
		32 PINS	
$\theta_{JA}$	Junction-to-ambient thermal resistance	35.1	
θ <sub>JCtop</sub>	Junction-to-case (top) thermal resistance	22.1	
$\theta_{JB}$	Junction-to-board thermal resistance	7.0	°C/W
$\Psi_{JT}$	Junction-to-top characterization parameter	0.3	C/W
Ψ <sub>JB</sub>	Junction-to-board characterization parameter	6.9	
$\theta_{JCbot}$	Junction-to-case (bottom) thermal resistance	1.3	

(1) For more information about traditional and new thermal metrics, see the IC Package Thermal Metrics application report, SPRA953.



## ELECTRICAL CHARACTERISTICS: V<sub>s</sub> = +12V

### Boldface limits are tested at +25°C.

At  $T_A = +25^{\circ}C$ , with  $R_{MATCH} = 10.2\Omega$ , transformer turn ratio 1:1.4,  $R_L = 100\Omega$  differential at transformer output, Full Bias Mode, and active impedance circuit configuration, unless otherwise noted. Each port is tested independently.

		٦	THS6226IRHB			TEST
PARAMETER	CONDITIONS	MIN	TYP	MAX	UNIT	LEVEL <sup>(1)</sup>
AC PERFORMANCE			1 1			
Small-signal bandwidth, -3dB         V <sub>O</sub> = 2V <sub>PP</sub> , differential at OUTCD and OUTAB, gain = 19V/V			125		MHz	С
0.1dB bandwidth flatness	$V_{O} = 2V_{PP}$		37		MHz	С
Large-signal bandwidth	$V_{O} = 10V_{PP}$		125		MHz	С
Slew rate (10% to 90% level)	V <sub>O</sub> = 15V step, differential		1500		V/µs	С
Rise and fall time	$V_{O} = 2V_{PP}$		2.8		ns	С
Harmonic distortion	$V_0 = 2V_{PP}, R_L = 60\Omega$ differential					С
Second harmonic	Full bias, f = 1MHz		-91		dBc	С
Third harmonic	Full bias, f = 1MHz		-91		dBc	С
	Full bias, f = 5MHz		-70		dBc	С
Second harmonic	Low bias, f = 5MHz		-64		dBc	С
	Full bias, f = 5MHz		-75		dBc	С
Third harmonic	Low bias, f = 5MHz		-47		dBc	С
Differential input voltage noise	f = 1MHz, input-referred		6.3		nV/√Hz	С
DC PERFORMANCE			I I I I I I I I I I I I I I I I I I I			
Differential gain			19		V/V	С
Differential gain error <sup>(2)</sup>				±2.5	%	А
· · · ·			±1	±5	mV	Α
Input offset voltage	-40°C to +85°C			±6	mV	В
Input offset voltage drift				15	µV/°C	В
Input offset voltage matching	Channels 1 to 2 and 3 to 4 only		±1	±5	mV	Α
INPUT CHARACTERISTICS						
Noninverting input resistance			500    2		kΩ    pF	С
Input bias voltage		5.8	6	6.2	V	А
OUTPUT CHARACTERISTICS			I			
Class H output voltage swing	$R_L = 60\Omega$ differential, class H operation <sup>(3)(4)</sup> , each output	+16/4	+17.5/-5.5		V	А
	-40°C to +85°C <sup>(3)(4)</sup>	+15.7/-3.7			V	В
	$R_L = 60\Omega$ differential, class H operation	±333	±383		mA	А
Class H output current (sourcing, sinking)	-40°C to +85°C	±323			mA	В
Class AB output voltage swing	$R_L = 60\Omega$ differential, normal operation <sup>(3)</sup> , each output	+9.9/+2.1	+10.1/+1.9		V	А
	-40°C to +85°C <sup>(3)</sup>	+9.8/+2.2			V	В
	$R_L = 60\Omega$ differential, normal operation	±130	±137		mA	A
Class AB output current (sourcing, sinking)	-40°C to +85°C	±126			mA	В
Short-circuit output current			1		А	С
Output impedance	f = 1MHz, differential		0.2		Ω	С
Crosstalk	$f = 1MHz$ , $V_{OUT} = 2V_{PP}$ , port 1 to port 2		-90		dB	С

Test levels: (A) 100% tested at +25°C. Over temperature limits set by characterization and simulation. (B) Limits set by characterization (1) and simulation. (C) Typical value only for information.

Negative feedback loop only. (2)

(3) (4) Measured at amplifier output (pin 17, 20, 21, and 24).

Capacitor fully charged, no droop.

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## ELECTRICAL CHARACTERISTICS: $V_s = +12V$ (continued)

### Boldface limits are tested at +25°C.

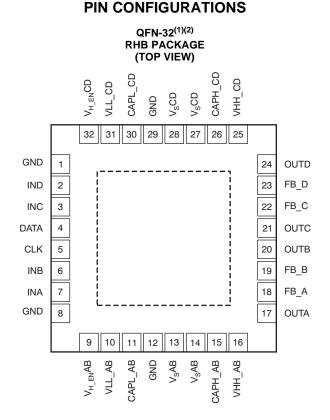
At  $T_A = +25^{\circ}$ C, with  $R_{MATCH} = 10.2\Omega$ , transformer turn ratio 1:1.4,  $R_L = 100\Omega$  differential at transformer output, Full Bias Mode, and active impedance circuit configuration, unless otherwise noted. Each port is tested independently.

			THS6226IRH	В		TEST
PARAMETER	CONDITIONS	MIN	TYP	MAX	UNIT	LEVEL <sup>(1)</sup>
POWER SUPPLY	1		1		1	
	Class AB	+10	+12	+15	V	А
•• •	-40°C to +85°C	+10		+15	V	В
Maximum operating voltage	Class H	+10	+12	+12.5	V	В
	-40°C to +85°C	+10		+12.5		В
	Per port, full bias, class H enable (power supply connected together)	22.5	23.5	24.5	mA	А
	_40°C to +85°C	21.8		25.2	mA	В
	Per port, full bias, class H disable (power supply connected together)	22.0	23.0	24.0	mA	А
	-40°C to +85°C	21.3		24.7	mA	В
	Bias current step		1.0		mA	С
I <sub>S+</sub> quiescent current	Per port, low bias, class H disable (power supply connected together)	7.2	7.6	8	mA	А
	-40°C to +85°C	6.9		8.3	mA	В
	Per port, line termination mode (B9 = B8 =B7 = B6 = 0) (power supply connected together)		4.4		mA	С
	Both ports, main amplifiers and class H disable (B9 = B8 = B7 = B6 = 0)		1.7	2.2	mA	А
	-40°C to +85°C			2.3	mA	В
Power-supply rejection (PSRR)	Differential, from +12V, GND	60	70		dB	А
	-40°C to +85°C	58			dB	В
LOGIC						
	Logic 1, with respect to GND <sup>(5)</sup>	1.9			V	С
Logic pin logic threshold	Logic 0, with respect to GND <sup>(5)</sup>			0.8	V	С
	Logic $X = 0.5V$ (logic 0)		10	25	μA	А
Logio pin quippont surrent	–40°C to +85°C			30	μA	В
Logic pin quiescent current	Logic X = 3.3V (logic 1)		66	125	μA	А
	-40°C to +85°C			130	μA	В
Turn-on time delay (t <sub>ON</sub> )	Time for I <sub>S</sub> to reach 50% of final value		1		μs	С
Turn-off time delay (t <sub>OFF</sub> )	Time for I <sub>S</sub> to reach 50% of final value		1		μs	С
Logic pin input impedance			50    1		kΩ∥pF	С

(5) The GND pin usable range is from  $V_{S-}$  to  $(V_{S+} - 5V)$ .

TEXAS INSTRUMENTS

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- (1) The PowerPAD is electrically isolated from all other pins and can be connected to any potential voltage range from  $V_{S-}$  to  $V_{S+}$ . Typically, the PowerPAD is connected to the GND plane because this plane tends to physically be the largest and is able to dissipate the most amount of heat.
- (2) The THS6226 defaults to the disabled mode at power-up.

**ISTRUMENTS** 

EXAS

T	PIN DESCRIPTIONS					
NAME	PIN	DESCRIPTION				
GND	1	Analog ground				
IND	2	Input D of amplifier CD				
INC	3	Input C of amplifier CD				
DATA	4	Serial interface data pin				
CLK	5	Serial interface CLK pin				
INB	6	Input B of amplifier AB				
INA	7	Input A of amplifier AB				
GND	8	Analog ground				
V <sub>H_EN</sub> AB	9	Class H mode control pin for amplifier AB				
VLL_AB	10	Amplifier AB low pump supply				
CAPL_AB	11	Amplifier AB negative voltage pump capacitor pin				
GND	12	Analog ground				
V <sub>S</sub> AB	13	Amplifier AB supply voltage				
V <sub>S</sub> AB	14	Amplifier AB supply voltage				
CAPH_AB	15	Amplifier AB positive voltage pump capacitor pin				
VHH_AB	16	Amplifier AB high pump supply				
OUTA	17	Output A of amplifier AB				
FB_A	18	Feedback for active output impedance of amplifier AB				
FB_B	19	Feedback for active output impedance of amplifier AB				
OUTB	20	Output B of amplifier AB				
OUTC	21	Output C of amplifier CD				
FB_C	22	Feedback for active output impedance of amplifier CD				
FB_D	23	Feedback for active output impedance of amplifier CD				
OUTD	24	Output D of amplifier CD				
VHH_CD	25	Amplifier CD high pump supply				
CAPH_CD	26	Amplifier CD positive voltage pump capacitor pin				
V <sub>S</sub> CD	27	Amplifier CD supply voltage				
V <sub>S</sub> CD	28	Amplifier CD supply voltage				
GND	29	Analog ground				
CAPL_CD	30	Amplifier CD negative voltage pump capacitor pin				
VLL_CD	31	Amplifier CD low pump supply				
V <sub>H_EN</sub> CD	32	Class H mode control pin for amplifier CD				

## TIMING CHARACTERISTICS

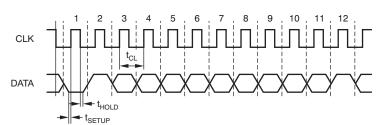


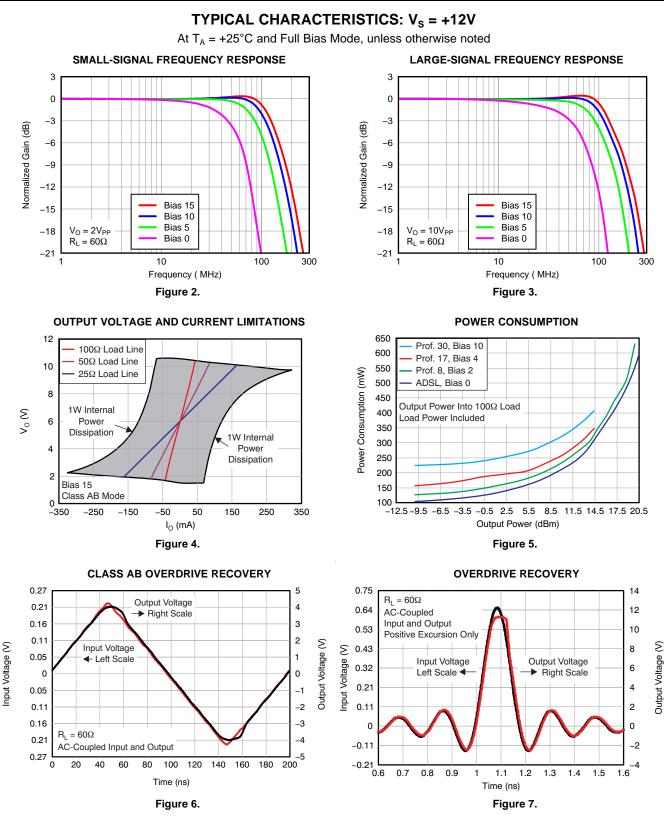
Figure 1. Serial Interface Timing

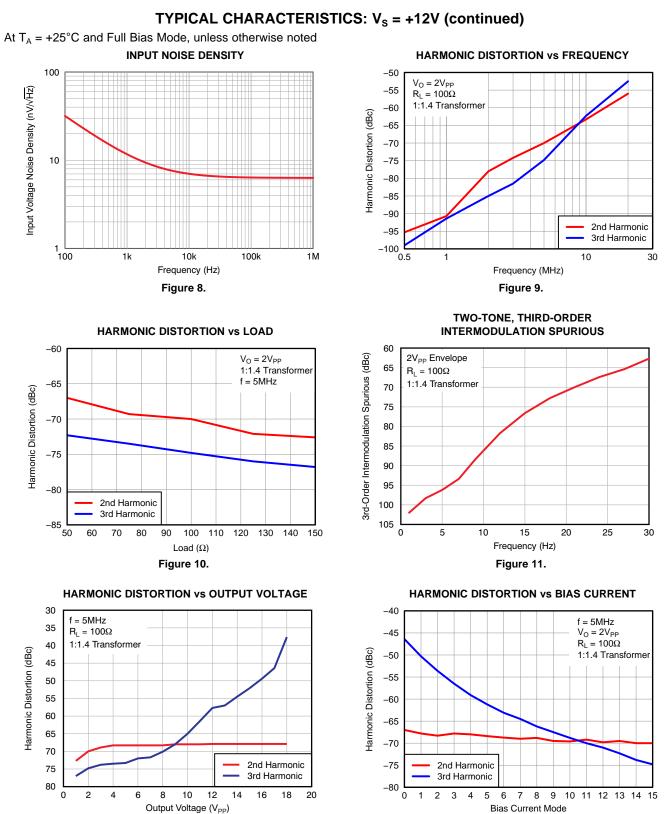
		THS		
PARAMETER	DESCRIPTION	MIN	MAX	UNITS
t <sub>SETUP</sub>	Setup time	3		ns
t <sub>HOLD</sub>	Hold time	0.5		ns
t <sub>CL</sub>	Clock period	200		ns





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Figure 12.

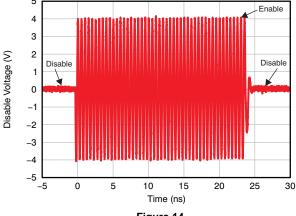
Figure 13.



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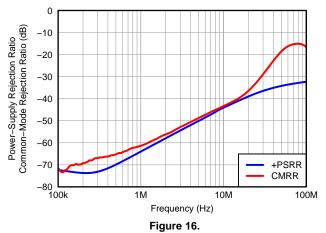
## TYPICAL CHARACTERISTICS: $V_s = +12V$ (continued)

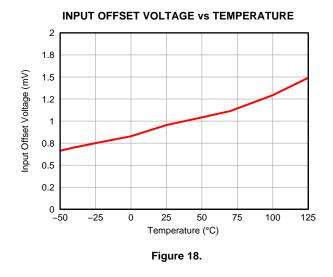
At T<sub>A</sub> = +25°C and Full Bias Mode, unless otherwise noted LARGE-SIGNAL ENABLE/DISABLE RESPONSE

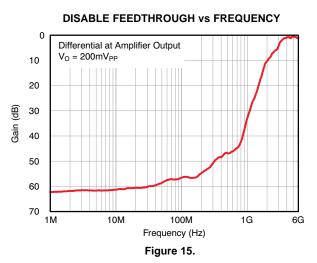




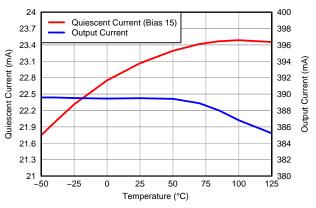




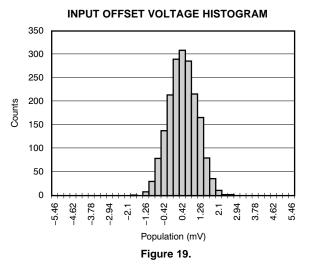




#### QUIESCENT CURRENT AND OUTPUT vs TEMPERATURE









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### **APPLICATION INFORMATION**

The THS6226 class H line driver provides exceptional ac performance in conjunction with wide output voltage swing. The class H operation allows voltage swings to exceed the power supply for short intervals limited only by the charge in the capacitor. In class AB mode, the THS6226 is capable of driving a  $60\Omega$  load from +1.9V to +10.1V. In class H mode, under the same conditions, the output voltage range becomes an impressive –5.5V to +17.5V, or  $46V_{PP}$  differentially with the capacitor fully charged.

Figure 20 shows a fully-differential, noninverting amplifier configuration with active impedance. In this configuration, the 10.2 $\Omega$  matching resistance appears through the transformer as 100 $\Omega$ , minimizing reflection on the line, while also minimizing transmission losses. The THS6226 gain is fixed and equal to 19V/V from input of the amplifier to the output of the amplifier (IN<sub>CD</sub> to OUT<sub>CD</sub>), not including the transformer-turn ratio.

To simplify the implementation as well as provide design flexibility, the THS6226 contains an integrated mid-supply buffer that provides the correct biasing to the amplifier core without requiring any external components. Also present is a two-pin serial interface that provides exceptional design flexibility and allows minimal power consumption for each xDSL profile.

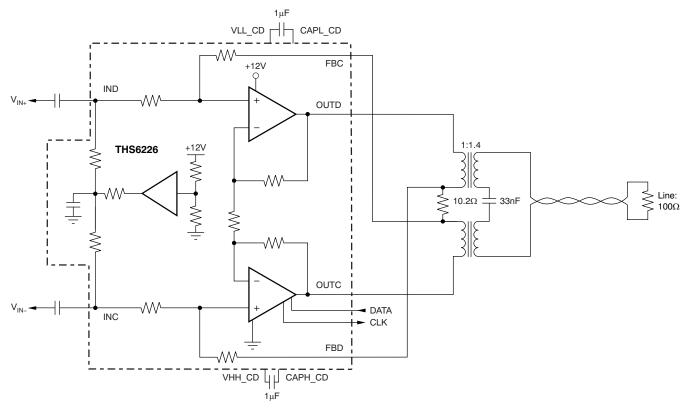


Figure 20. Multi-Tone Power Ratio (MTPR) Test Circuit



**THS6226** 

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### **PROGRAMMING THE THS6226**

Programming of the THS6226 is realized through a serial interface (pins 4 and 5) and proceeds in the following sequence.

Two start bits are required B0 = 0 followed by B1 = 1.

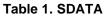
B2 through B9 are used to program the THS6226.

Refer to Table 1 for the bit descriptions.

B10 (refer to Table 2) is the parity bit that controls if the word is or is not loaded.

B11 is the stop bit and should be set to B11 = 1. Figure 21 shows the sequence to be adopted.

PARAMETER	DESCRIPTION			
B0, B1	Start bit			
B2, B3	Channel select			
B4, B5	Power-down features			
B6-B9	Quiescent current setting			
B10	Parity bit			
B11	Stop bit			



B10	ODD PARITY BIT			
0	If odd, number of high bits in B2 to B9			
1	If even, number of high bits in B2 to B9			
MCP				

Table 2. Parity Bit

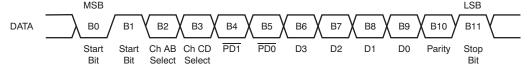


Figure 21. DATA Description



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### QUIESCENT CURRENT

The quiescent current of the THS6226 is dissipated in two main modules of the THS6226: the class AB and the charge pump. B4 and B5 select the mode of operation, class AB operating with or without the charge pump enabled, powering down the entire port, or operating in a line termination mode. Table 4 lists the details on each bit functionality and the approximate quiescent current.

The class AB quiescent current is set by bits B6 to B9, using B4 and B5 for the power-down function, and B2 and B3 for channel select. The approximate quiescent current for the amplifier core is shown in Table 3.

B6 (D3)	B7 (D2)	B8 (D1)	B9 (D0)	QUIESCENT CURRENT SETTING	APPROXIMATE I <sub>Q</sub> (mA/Port)
0	0	0	0	ADSL2+ mode	7.6
0	0	0	1		8.7
0	0	1	0	Profile 8b mode	9.8
0	0	1	1		10.9
0	1	0	0	Profile 17a mode	12
0	1	0	1		13
0	1	1	0		14
0	1	1	1		15
1	0	0	0		16
1	0	0	1		17
1	0	1	0	Profile 30a mode	18
1	0	1	1		19
1	1	0	0		20
1	1	0	1		21
1	1	1	0		22
1	1	1	1		23

#### Table 3. Class AB Quiescent Current

The various power modes are shown in Table 4. For all modes, when B6 through B9 are not defined, set B9 = B8 = B7 = B6 = 0 to achieve the lowest power dissipation possible.

#### Table 4. Power Modes

B4 (PD1)	B5 (PD0)	POWER-DOWN MODE	APPROXIMATE I <sub>Q</sub> (mA/Port)
0	0	Power-down (B9, B8, B7, B6 = 0)	0.85
0	1	Line termination mode (B9, B8, B7, B6 = 0)	4.4
1	0	Class AB driver $\rm I_Q$ set by B6 to B9, class H disabled	_
1	1	Class AB driver I <sub>Q</sub> set by B6 to B9, class H enabled	_



Channel selection is shown in Table 5. Each channel can be programmed independently, or together if both B2 and B3 are set to '1'.

B2 (Channel AB)	B3 (Channel CD)	CHANNEL SELECT
0	0	Bits B4 to B9 are ignored
0	1	Channel B programmed with B4 to B9
1	0	Channel A programmed with B4 to B9
1	1	Channels A and B programmed with B4 to B9

#### **Table 5. Channel Selection**

At startup, the internal register is set as shown in Table 6.

### Table 6. Internal Register

B2 (Channel AB)	B3 (Channel CD)	B4 (PD1)	B5 (PD0)	B6 (D3)	B7 (D2)	B8 (D1)	B9 (D0)
0	0	0	0	0	0	0	0

In this condition, the total quiescent power dissipation is 10.2mW/port on a +12V supply.



## **REVISION HISTORY**

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

CI	hanges from Revision B (February 2011) to Revision C	Page
•	Changed LOGIC, Logic pin input impedance typical specification and unit in Electrical Characteristics table	4
•	Changed Timing Characteristics section	6



11-Apr-2013

## **PACKAGING INFORMATION**

Orderable Device	Status	Package Type	Package	Pins	Package	Eco Plan	Lead/Ball Finish	MSL Peak Temp	Op Temp (°C)	Top-Side Markings	Samples
	(1)		Drawing		Qty	(2)		(3)		(4)	
THS6226IRHBR	ACTIVE	QFN	RHB	32	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR	-40 to 85	THS6226 IRHB	Samples
THS6226IRHBT	ACTIVE	QFN	RHB	32	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR	-40 to 85	THS6226 IRHB	Samples

<sup>(1)</sup> The marketing status values are defined as follows:

**ACTIVE:** Product device recommended for new designs.

**LIFEBUY:** TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

**NRND:** Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

**PREVIEW:** Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.

**TBD:** The Pb-Free/Green conversion plan has not been defined.

**Pb-Free (RoHS):** TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes. **Pb-Free (RoHS Exempt):** This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between

the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

<sup>(3)</sup> MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) Multiple Top-Side Markings will be inside parentheses. Only one Top-Side Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Top-Side Marking for that device.

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# PACKAGE MATERIALS INFORMATION

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### TAPE AND REEL INFORMATION





## QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



*All dimensions are nominal												
Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
THS6226IRHBR	QFN	RHB	32	3000	330.0	12.4	5.3	5.3	1.5	8.0	12.0	Q2
THS6226IRHBT	QFN	RHB	32	250	180.0	12.4	5.3	5.3	1.5	8.0	12.0	Q2

TEXAS INSTRUMENTS

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# PACKAGE MATERIALS INFORMATION

26-Jan-2013



\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
THS6226IRHBR	QFN	RHB	32	3000	367.0	367.0	35.0
THS6226IRHBT	QFN	RHB	32	250	210.0	185.0	35.0



NOTES: A. All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M-1994.

- B. This drawing is subject to change without notice.
- C. QFN (Quad Flatpack No-Lead) Package configuration.
- D. The package thermal pad must be soldered to the board for thermal and mechanical performance.
- E. See the additional figure in the Product Data Sheet for details regarding the exposed thermal pad features and dimensions.
- F. Falls within JEDEC MO-220.



## RHB (S-PVQFN-N32)

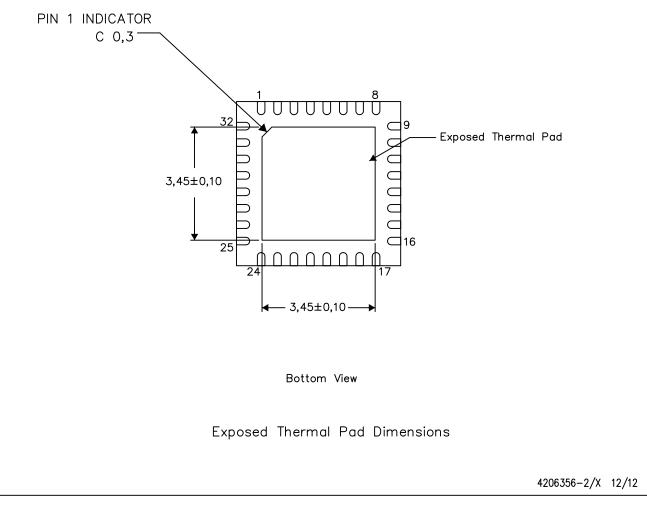
## PLASTIC QUAD FLATPACK NO-LEAD

### THERMAL INFORMATION

This package incorporates an exposed thermal pad that is designed to be attached directly to an external heatsink. The thermal pad must be soldered directly to the printed circuit board (PCB). After soldering, the PCB can be used as a heatsink. In addition, through the use of thermal vias, the thermal pad can be attached directly to the appropriate copper plane shown in the electrical schematic for the device, or alternatively, can be attached to a special heatsink structure designed into the PCB. This design optimizes the heat transfer from the integrated circuit (IC).

For information on the Quad Flatpack No-Lead (QFN) package and its advantages, refer to Application Report, QFN/SON PCB Attachment, Texas Instruments Literature No. SLUA271. This document is available at www.ti.com.

The exposed thermal pad dimensions for this package are shown in the following illustration.

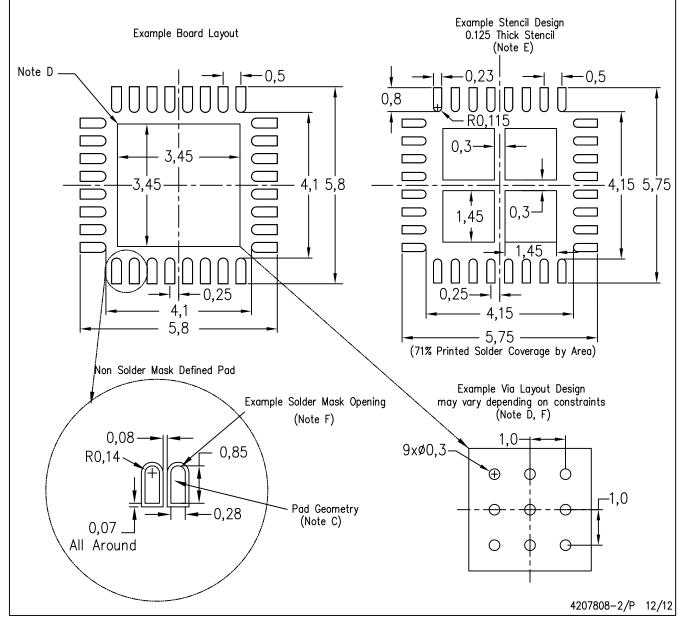


#### NOTE: A. All linear dimensions are in millimeters



# RHB (S-PVQFN-N32)

# PLASTIC QUAD FLATPACK NO-LEAD



NOTES: A. All linear dimensions are in millimeters.

- B. This drawing is subject to change without notice.
- C. Publication IPC-7351 is recommended for alternate designs.
- D. This package is designed to be soldered to a thermal pad on the board. Refer to Application Note, Quad Flat-Pack Packages, Texas Instruments Literature No. SLUA271, and also the Product Data Sheets for specific thermal information, via requirements, and recommended board layout. These documents are available at www.ti.com <a href="http://www.ti.com">http://www.ti.com</a>.
- E. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC 7525 for stencil design considerations.
- F. Customers should contact their board fabrication site for recommended solder mask tolerances and via tenting recommendations for vias placed in the thermal pad.



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