

SN65LVCP114

SLLSEA8 -JANUARY 2012

# 14.2-Gbps Quad 1:2-2:1 MUX, Linear-Redriver With Signal Conditioning

Check for Samples: SN65LVCP114

## **FEATURES**

- Quad 2:1 Mux / 1:2 Demux
- Multi-Rate Operation up to 14.2 Gbps Serial Data Rate
- Linear Receiver Equalization Which Increases Margin at System Level of Decision Feedback Equalizer
- Bandwidth: 18 GHz, Typical
- Per-Lane P/N Pair Inversion
- Port or Single Lane Switching
- Low Power: 150 mW/Channel, Typical
- Loopback Mode on All Three Ports
- I<sup>2</sup>C Control in Addition to GPIO
- DIAG Mode That Outputs Data of Line Side Port to Both Fabric Side Ports
- 2.5-V/3.3-V Single Power Supply
- PBGA Package 12-mm × 12-mm × 1-mm, 0.8-mm Terminal Pitch
- Excellent Impedance Matching to 100- $\Omega$  PCB Transmission Lines

## DESCRIPTION

- Small Package Size Provides Board Real Estate Saving
- Adjustable Output Swing Provides Flexible
   EMI and Crosstalk Control
- Low Power
- Supports 10GBASE-KR Applications With Ability to Transparency for Link Training

## **APPLICATIONS**

- High-Speed Redundancy Switch in Telecom and Data Communication
- Backplane Interconnect for 10G-KR, 16GFC

The SN65LVCP114 is an asynchronous, protocol-agnostic, low-latency QUAD mux, linear-redriver optimized for use in systems operating at up to 14.2 Gbps. The device linearly compensates for channel loss in backplane and active-cable applications. The architecture of SN65LVCP114 linear-redriver is designed to work effectively with ASIC or FPGA products implementing digital equalization using decision feedback equalizer (DFE) technology. The SN65LVCP114 mux, linear-redriver preserves the integrity (composition) of the received signal, ensuring optimum DFE and system performance. The SN65LVCP114 provides a low-power mux-demux, linear-redriver solution while at the same time extending the effectiveness of DFE.

SN65LVCP114 is configurable via GPIO or an I<sup>2</sup>C interface.

A single 2.5-V or 3.3-V power supply supports the operation of the SN65LVCP114.

The SN65LVCP114 is packaged in a 12-mm × 12-mm × 1-mm PBGA package with 0.8-mm pitch.

The SN65LVCP114 has three ports; each port is a quad lane. The switch logic of SN65LVCP114 can be implemented to support a 2:1 MUX per lane, 1:2 DEMUX per lane, and independent lane switching. The receive equalization can be independently programmed for each of the ports. The SN65LVCP114 supports loopback on all three ports.



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These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

## **TYPICAL IMPLEMENTATION**

SN65LVCP114 can be implemented on the transmit side or the receive side of a backplane channel as shown in Figure 1.

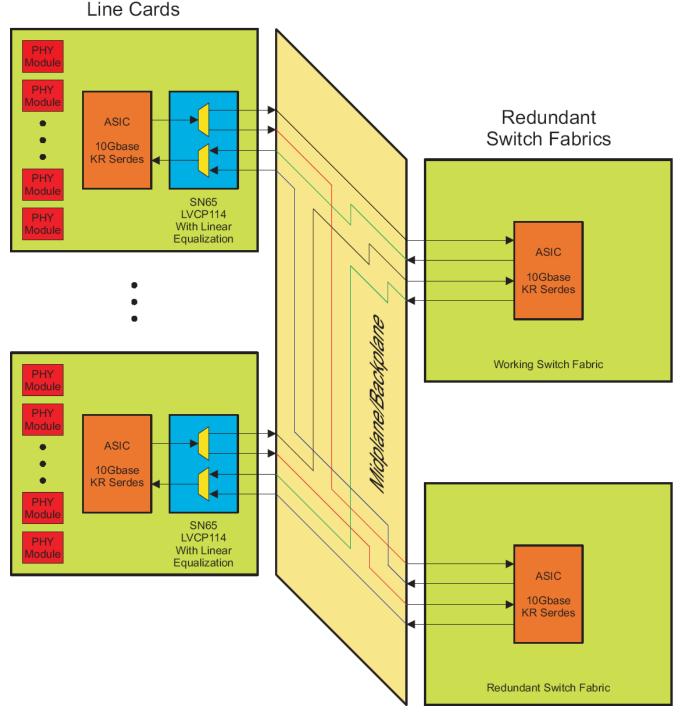


Figure 1. SN65LVCP114 Typical Implementation



### **BLOCK DIAGRAM**

A simplified block diagram of the SN65LVCP114 is shown in Figure 2 for input quad channels AIN and BIN through the 2:1 MUX and output quad channel COUT, together with the input quad channel CIN through the 1:2 DEMUX through output quad channels AOUT and BOUT. The MUX and DEMUX channels contain a linear receive equalizer and an output linear driver.

The SN65LVCP114 provides both GPIO and  $I^2C$  interfaces to control the configuration of the device. A detailed description of the SN65LVCP114 pin functions is provided in .

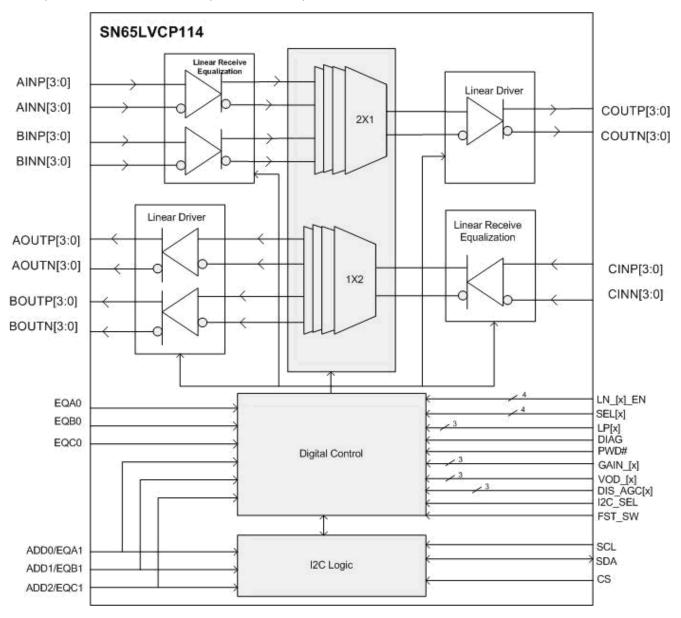


Figure 2. Simplified Block Diagram of SN65LVCP114



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## PACKAGE

The package for the SN65LVCP114 is a 12-mm × 12-mm × 1-mm, 167 pin PBGA with 0.8-mm pitch. The top view with the pin names is shown in Figure 3.

							тор	VIEW							
	٨	в	с	D	Е	F	G	н	J	к	L	м	N	Р	
14	ADD0/ EQA1	ADD1/ EQB1	EQCO	I2C_SEL	P₩D#	AINPO	AINNO	GND	AINP1	AINN1	LN_2_EN	LPC	LPB	LPA	14
13	CINNO	VCC	CS	FST_SW	VCC	GND	VCC	VCC	GND	LN_1_EN	LN_3_EN	AOUTP2	VCC	REXT	13
12	CINPO	GND	COUTNO	GND	AOUTPO	AOUTNO	GND	AOUTP1	AOUTNI	LN_O_EN	DIAG	AOUTN2	GND	AINP2	12
11	GND	VCC	COUTPO									GND	VCC	AINN2	11
10	CINN1	VCC	GND		GND	GND	GND	GND	GND	GND		AOUTP3	VCC	GND	10
э	CINP1	GND	COUTNI		GND	GND	GND	GND	GND	GND		AOUTN3	GND	AINP3	э
8	GND	VCC	COUTP1		GND	GND	GND	GND	GND	GND		GND	GND	AINN3	8
7	CINN2	VCC	GND		GND	GND	GND	GND	GND	GND		BOUTPO	VCC	GND	7
6	CINP2	GND	COUTN2		GND	GND	GND	GND	GND	GND		BOUTNO	GND	BINPO	6
5	GND	VCC	COUTP2			GND	GND	GND	GND	GND		GND	VCC	BINNO	5
4	CINN3	VCC	GND									BOUTP1	VCC	GND	4
3	CINP3	GND	COUTN3	EQAO	GAIN_C	BOUTN3	BOUTP3	GND	BOUTN2	BOUTP2	GND	BOUTNI	GND	BINP1	3
2	SDA	SCL	COUTP3	GAIN_B	GAIN_A	GND	VCC	VCC	GND	VCC	VOD_C	VOD_B	VOD_A	BINN1	2
1	ADD2/ EQC1	DIS_AGC_A	DIS_AGC_B	DIS_AGC_C	BINN3	BINP3	GND	BINN2	BINP2	EQ_B0	SEL3	SEL2	SEL1	SELO	1
	A	в	с	D	E	F	G	н	J	к	L	м	N	Р	

### Figure 3. Package Pinout

### PIN DESCRIPTIONS

	PIN	DIRECTION TYPE SUPPLY	DESCRIPTION
SIGNAL	BALLS	DIRECTION TYPE SUPPLY	DESCRIPTION
LINE-SIDE HIGH	I-SPEED I/O		
CINP0 CINN0	A12 A13	Input (with 50- $\Omega$ termination to input common mode)	Differential input, lane 0 line side.
CINP1 CINN1	A9 A10	Input (with 50-Ω termination to input common mode)	Differential input, lane 1 line side
CINP2 CINN2	A6 A7	Input (with 50- $\Omega$ termination to input common mode)	Differential input, lane 2 line side
CINP3 CINN3	A3 A4	Input (with 50- $\Omega$ termination to input common mode)	Differential input, lane 3 line side
COUTP0 COUTN0	C11 C12	Output	Differential output, lane 0 line side
COUTP1 COUTN1	C8 C9	Output	Differential output, lane 1 line side
COUTP2 COUTN2	C5 C6	Output	Differential output, lane 2 line side
COUTP3 COUTN3	C2 C3	Output	Differential output, lane 3 line side

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# PIN DESCRIPTIONS (continued)

	PIN					DESCRIPTION		
SIGNAL	BALLS	DIRECTION TYPE SUPPLY				DESCRIPTION	4	
SWITCH-SIDE HI	GH-SPEED I/O							
AINP0 AINN0	F14G14	Input (with 50- $\Omega$ termination to input common mode)	Different	tial input, lane	e 0, fabric switcl	h_A_side		
AINP1 AINN1	J14 K14	Input (with 50- $\Omega$ termination to input common mode)	Different	tial input, lane	e 1, fabric switcl	h_A_side		
AINP2 AINN2	P12 P11	Input (with 50- $\Omega$ termination to input common mode)	Different	tial input, lane	e 2, fabric switcl	h_A_side		
AINP3 AINN3	P9 P8	Input (with 50- $\Omega$ termination to input common mode)	Different	tial input, lane	e 3, fabric switcl	h_A_side		
BINP0 BINN0	P6 P5	Input (with 50- $\Omega$ termination to input common mode)	Different	tial input, lane	e 0, fabric switcl	h_B_side		
BINP1 BINN1	P3 P2	Input (with 50- $\Omega$ termination to input common mode)	Differential input, lane 1, fabric switch_B_side					
BINP2 BINN2	J1 H1	Input (with 50-Ω termination to input common mode)	Differential input, lane 2, fabric switch_B_side					
BINP3 BINN3	F1 E1	Input (with 50- $\Omega$ termination to input common mode)	Differential input, lane 3, fabric switch_B_side					
AOUTP0 AOUTN0	E12 F12	Output	Differential output, lane 0, fabric switch_A_side					
AOUTP1 AOUTN1	H12 J12	Output	Differential output, lane 1, fabric switch_A_side					
AOUTP2 AOUTN2	M13 M12	Output	Different	tial output, lan	ne 2, fabric swit	ch_A_side		
AOUTP3 AOUTN3	M10 M9	Output	Different	tial output, lan	ne 3, fabric swit	ch_A_side		
BOUTP0 BOUTN0	M7 M6	Output	Different	tial output, lan	ne 0, fabric swit	ch_B_side		
BOUTP1 BOUTN1	M4 M3	Output	Different	tial output, lan	ne 1, fabric swit	ch_B_side		
BOUTP2 BOUTN2	К3 J3	Output	Different	tial output, lan	ne 2, fabric swit	ch_B_side		
BOUTP3 BOUTN3	G3 F3	Output	Different	tial output, lan	ne 3, fabric swit	ch_B_side		
CONTROL SIGN	ALS							
ADD0/EQA1 ADD1/EQB1 ADD2/EQC1	A14 B14 A1	Input, 2.5-V/3.3-V CMOS - 3-state	EQ gain EQ cont EQ gain EQ cont	rol pins. EQA of port A. rol pins. EQB of port B.	1 and EQB0 pi	ns are 3-state and cor ns are 3-state and cor ns are 3-state and cor	ntrol the	I <sup>2</sup> C mode ADD0 along with pins ADD1 and ADD2 comprise the three bits of the I <sup>2</sup> C slave address.
				EQ[x]0	EQ[x]0	Peaking in dB		
				0	0	1.3		
				0	HiZ	2		
				0	1	3.6		
				HiZ	0	5		
			HiZ HiZ 6.5					
				HiZ	1	8.3		
				1	0	10		
				1	HiZ	11.9		
				1	1	13.9		

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## **PIN DESCRIPTIONS (continued)**

SIGNAL	PIN BALLS	DIRECTION TYPE SUPPLY				DESCRIPTION	
EQA0 EQB0 EQC0	D3 K1 C14	Input, 2.5-V/3.3-V CMOS - 3-state	theEQ ga EQ contr theEQ ga EQ contr	rol pins. EQA ain of port A. rol pins. EQB <sup>-</sup> ain of port B.	1 and EQB0 p	ns are 3-state and control ns are 3-state and control ins are 3-state and control	I <sup>2</sup> C mode No action needed
			) I	E0[v]0	EO[v10	Booking in dP	
				EQ[x]0 0	EQ[x]0	Peaking in dB	
				0	HiZ	1.3	
				0	1	3.6	
				HiZ	0	5	
				HiZ	HiZ	6.5	
				HiZ	1	8.3	
				1	0	10	
				1	HiZ	11.9	
				1	1	13.9	
LPA LPB LPC	P14 N14 M14	Input (with 48-kΩ pulldown) 2.5-V/3.3-V CMOS	HIGH: Lo LOW: Lo	bles loopback oopback enab oopback disab	oled	e logic	I <sup>2</sup> C mode No action needed
SEL0	P1	Input (with 48-kΩ pulldown)	GPIO m				I <sup>2</sup> C mode
SEL1 SEL2	N1 M1	2.5-V/3.3-V CMOS		B switch cont ort B is select			No action needed
SEL3	L1			ort A is selecte			
			See Tab	le 2 and Figu	re 12 for devic	e logic	
REXT	P13	Input, analog	External	bias resistor,	1,200 $\Omega$ to gr	bund	I
CS	C13	Input (with 48-kΩ pulldown) 2.5-V/3.3-V CMOS	GPIO me No action	ode n needed			<b>I<sup>2</sup>C mode</b> HIGH: acts as chip select LOW: disables I <sup>2</sup> C interface
PWD	E14	Input (with 48-kΩ pullup) 2.5-V/3.3-V CMOS		owers down th ormal operation		ts off and outputs disabled, r	esets I <sup>2</sup> C
DIAG	L12	Input (with 48-kΩ pulldown) 2.5-V/3.3-V CMOS	output or LOW: No	nables the sa n both fabric s ormal operatio	side ports (Por	e line side (Port C) to be t A and Port B). e logic.	I <sup>2</sup> C mode No action needed
LN_0_EN	K12	Input (with 48-kΩ pullup)	GPIO m	ode			I <sup>2</sup> C mode
LN_1_EN LN_2_EN LN_3_EN	K13 L14 L13	2.5-V/3.3-V CMOS		0		ports A, B, and C ports A, B, and C	No action needed
DIS_AGC_A DIS_AGC_B DIS_AGC_C	B1 C1 D1	Input (with 48-kΩ pulldown) 2.5-V/3.3-V CMOS	DIS_AG	the AGC loop C = High, disa	p internal to th ables the AGC bles the AGC		I <sup>2</sup> C mode No action needed
VOD_A VOD_B VOD_C	N2 M2 L2	Input, 2.5-V/3.3-V CMOS - 3-state	2.2	elects VOD ou		2 V maximum and a gain of	I <sup>2</sup> C mode No action needed
			1.1			0 mV maximum and a gain o ximum and a gain of 2.2.	
Gain_A	E2	Input, 2.5-V/3.3-V CMOS - 3-state	GPIO m	•			I <sup>2</sup> C mode
Gain_B	D2		HIGH: R	eceiver gain =			No action needed
Gain_C	E3			eceiver gain = ating, it defaul			
SDA	A2	Input / output, open-drain output	GPIO m	0.	10 0.0		J <sup>2</sup> C mode
				n needed			I <sup>2</sup> C mode I <sup>2</sup> C data. Connect a 10-kΩ pullu resistor externally
SCL	B2	Input, open-drain input	GPIO mo No action	ode n needed			I <sup>2</sup> C mode I <sup>2</sup> C clock. Connect a 10-kΩ pullt resistor externally



### **PIN DESCRIPTIONS (continued)**

	PIN		DESCRIPTION					
SIGNAL	BALLS	DIRECTION TYPE SUPPLY	DESCRIPTION					
FST_SW	D13	Input (with 48-kΩ pullup) 2.5-V/3.3-V CMOS input	GPIO mode HIGH: Fast switching; the idle outputs are squelched (see tSM specification). LOW: Slow switching; the idle outputs are powered off (see tSM1 specification).	I <sup>2</sup> C mode No action needed				
I2C_SEL	D14	Input (with 48-kΩ pulldown) 2.5-V/3.3-V CMOS input	Configures the device in I <sup>2</sup> C or GPIO mode of operation HIGH: Enables I <sup>2</sup> C mode LOW: Enables GPIO mode					
POWER SUPPL	Y							
VCC	B4, B5, B7, B8, B10, B11, B13, E13, G2, G13, H2, H13, K2,N4, N5, N7, N10, N11, N13	Power, 2.5 V ±5% / 3.3 V ±5%	Power supply pins					
GROUND	- I							
GND	A5, A8, A11, B3, B6, B9, B12, C4, C7, C10,D12, F2, F13,G1, G12, G1, G12, H3, H14,J2, J13, L3, M5,M8, M11, N3, N6, N8, N9, N12, P4, P7, P10	Ground	Ground pins					
GND CenterPad	E6, E7, E8, E9, E10, F5, F6, F7, F8, F9, F10, G5, G6, G7, G8, G9, G10, H5, H6, H7, H8, H9, H10, J5, J6, J7, J8, J9, J10, K5, K6, K7,K8, K9, K10	Ground	These pins must be connected to the GND plane.					

### **ABSOLUTE MAXIMUM RATINGS**

over operating free-air temperature range (unless otherwise noted) <sup>(1)</sup>

				VALUE / UNIT
V <sub>CC</sub>	Supply volta	age range <sup>(2)</sup>		–0.3 V to 4 V
V <sub>IN,DIFF</sub>	Differential	voltage between xINx_P and xIN	x_N	±2.5 V
V <sub>IN+, IN</sub>				–0.5 V to VCC + 0.5 V
V <sub>IO</sub>	Voltage on	control I/O pins		–0.3 V to VCC + 0.5 V
I <sub>IN+</sub> I <sub>IN-</sub>	Continuous	current at high-speed differential	data inputs (differential)	-25 mA to 25 mA
I <sub>OUT+</sub> I <sub>OUT-</sub>	Continuous	current at high-speed differential	data outputs	-25 mA to 25 mA
ESD		Human-body model <sup>(3)</sup>	All pins	2 kV
E3D		Charged-device model <sup>(4)</sup>	All pins	500 V

(1) Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under Recommended Operating Conditions is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

(2) All voltage values, except differential I/O bus voltages, are with respect to network ground terminal.

(3) Tested in accordance with JEDEC Standard 22, Test Method A114-A.

(4) Tested in accordance with JEDEC Standard 22, Test Method C101.

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STRUMENTS

EXAS

### THERMAL INFORMATION

	THERMAL METRIC <sup>(1)</sup>	SN65LVCP114	
		ZJA (167 PINS)	UNITS
$\theta_{JA}$	Junction-to-ambient thermal resistance <sup>(2)</sup>	38.8	
θ <sub>JCtop</sub>	Junction-to-case (top) thermal resistance <sup>(3)</sup>	7.55	
$\theta_{JB}$	Junction-to-board thermal resistance <sup>(4)</sup>	17.8	°C11/
Ψ <sub>JT</sub>	Junction-to-top characterization parameter <sup>(5)</sup>	0.2	°C/W
Ψ <sub>JB</sub>	Junction-to-board characterization parameter <sup>(6)</sup>	17.5	
$\theta_{\text{JCbot}}$	Junction-to-case (bottom) thermal resistance <sup>(7)</sup>	N/A	

For more information about traditional and new thermal metrics, see the *IC Package Thermal Metrics* application report, SPRA953.
 The junction-to-ambient thermal resistance under natural convection is obtained in a simulation on a JEDEC-standard, high-K board, as

specified in JESD51-7, in an environment described in JESD51-2a.

(3) The junction-to-case (top) thermal resistance is obtained by simulating a cold plate test on the package top. No specific JEDEC-standard test exists, but a close description can be found in the ANSI SEMI standard G30-88.

(4) The junction-to-board thermal resistance is obtained by simulating in an environment with a ring cold plate fixture to control the PCB temperature, as described in JESD51-8.

(5) The junction-to-top characterization parameter,  $\psi_{JT}$ , estimates the junction temperature of a device in a real system and is extracted from the simulation data for obtaining  $\theta_{JA}$ , using a procedure described in JESD51-2a (sections 6 and 7).

(6) The junction-to-board characterization parameter,  $\psi_{JB}$ , estimates the junction temperature of a device in a real system and is extracted from the simulation data for obtaining  $\theta_{JA}$ , using a procedure described in JESD51-2a (sections 6 and 7).

(7) The junction-to-case (bottom) thermal resistance is obtained by simulating a cold plate test on the exposed (power) pad. No specific JEDEC standard test exists, but a close description can be found in the ANSI SEMI standard G30-88.

## **RECOMMENDED OPERATING CONDITIONS**

SPEC	CIFICATION	MIN	NOM	MAX	UNIT	
	Operating data rate, dR			14.2	Gbps	
	Supply voltage, V <sub>CC</sub> , 2.5-V nominal supply	2.375	2.5	2.625	V	
	Supply voltage, V <sub>CC</sub> , 3.3-V nominal supply	3.135	3.3	3.465	V	
	PSNR BG, bandgap circuitry PSNR, 10 Hz–10 GHz	20			dB	
CON	CONTROL INPUTS					
VIH	High-level input voltage	0.8 × V <sub>CC</sub>				
V <sub>IM</sub>	Mid-level input voltage	V <sub>CC</sub> /2 – 0.3		$V_{CC}/2 + 0.3$	V	
VIL	Low- level input voltage			$0.2 \times V_{CC}$	V	
T <sub>C</sub>	Junction temperature <sup>(1)</sup>	-10		125	°C	
	Maximum Board Temperature <sup>(1)</sup>			See Table 1	°C	

(1) Use of θJB and φJB are recommended for thermal calculations. For more information about traditional and new thermal metrics, see IC Package Thermal Metrics application report, SPRA953.



## ELECTRICAL CHARACTERISTICS (V<sub>cc</sub> 2.5 V ±5%)

over operating conditions range. All parameters are referenced to package pins (unless otherwise noted).

	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
POWER	CONSUMPTION					
$PD_L$	Device power dissipation, loopback mode	Ports A, B, and C in loopback mode with all 12 channels active. VOD = LOW		1800	2300	mW
$PD_N$	Device power dissipation, normal mode	Device configured in mux-demux mode with 8 channels active. VOD = LOW		1400	1800	mW
PD <sub>OFF</sub>	Device power dissipation, lanes disabled	All 4 lanes disabled. See the I <sup>2</sup> C section for device configuration.		50		mW
PD <sub>STB</sub>	Device power dissipation, standby	All 12 channels active, VOD = LOW, FAST_SW = HIGH. See the $I^2C$ section for device configuration.		1800	2300	mW

# ELECTRICAL CHARACTERISTICS (V<sub>cc</sub> 3.3 V ±5%)

over operating conditions range. All parameters are referenced to package pins (unless otherwise noted).

PARAMETER		TEST CONDITIONS		TYP	MAX	UNIT
POWER	CONSUMPTION					
$PD_L$	Device power dissipation, loopback mode	Ports A, B, and C in loopback mode with all 12 channels active. VOD = LOW		2500	3150	mW
PD <sub>N</sub>	Device power dissipation, normal mode	Device configured in mux-demux mode with 8 channels active. VOD = LOW		1800	2500	mW
PD <sub>OFF</sub>	Device power dissipation, lanes disabled	All 4 lanes disabled. See the I <sup>2</sup> C section for device configuration.		50		mW
PD <sub>STB</sub>	Device power dissipation, standby	All 12 channels active, VOD = LOW, FAST_SW = HIGH. See $I^2C$ section for device configuration.		2500	3150	mW

## ELECTRICAL CHARACTERISTICS (V<sub>cc</sub> 3.3 V $\pm$ 5%, 2.5 V $\pm$ 5%)

### over operating conditions range. All parameters are referenced to package pins (unless otherwise noted).

	PARAMETER	TEST CONDITIONS	MIN TYP <sup>(1)</sup>	MAX	UNIT
CMOS DC	SPECIFICATIONS	<u>.</u>	<u>.</u>	·	
I <sub>IH</sub>	High-level input current	$V_{IN} = 0.9 \times V_{CC}$		80	μA
I <sub>IL</sub>	Low-level input current	$V_{IN} = 0.1 \times V_{CC}$	-80		μA
CML INPU	TS (AINP[3:0], AINN[3:0], BINP[3:0], BINN[3:0], CINI	P[3:0], CINN[3:0])			
r <sub>in</sub>	Differential input resistance	INx_P to INx_N	100		Ω
VINPP	Input linear dynamic range	Gain = 0.5	1200		mVpp
VICM	Common-mode input voltage	Internally biased	V <sub>CC</sub> - 0.3		V
SCD11	Input differential to common-mode conversion	100 MHz to 7.1GHz	-25		dB
SDD11	Differential input return loss	100 MHz to 7.1GHz	-10		dB
CML OUTF	PUTS (AOUTP[3:0], AOUTN[3:0], BOUTP[3:0], BOUT	N[3:0], COUTP[3:0], COUTN[3:0])			
	Output linear dynamic range	$R_L = 100 \Omega$ , $V_{OD} = High$	1200		
V <sub>OD</sub>	Output linear dynamic range	$R_L = 100 \Omega$ , $V_{OD} = Low$	600		$mV_{PP}$
Vos	Output offset voltage	$R_L = 100 \Omega$ , 0 V applied at inputs		20	$\mathrm{mV}_{\mathrm{PP}}$
V <sub>CM,RIP</sub>	Common-mode output ripple	K28.5 pattern at 14.2 Gbps, no interconnect loss, V <sub>oD</sub> = HIGH	10	20	mV <sub>RMS</sub>
V <sub>OD,RIP</sub>	Differential path output ripple	K28.5 pattern at 14.2Gbps, no interconnect loss, V <sub>IN</sub> = 1200 mVpp. Outputs squelched.		20	$mV_{PP}$
V <sub>OCM</sub>	Output common mode voltage	See Figure 4	V <sub>CC</sub> - 0.35		V
V <sub>OC(SS)</sub>	Change in steady-state common-mode output voltage between logic states		±10		mV
t <sub>PLH</sub>	Low-to-high propagation delay		200		ps
t <sub>PHL</sub>	High-to-low propagation delay	See Figure 5	200		ps
t <sub>SK(O)</sub>	Inter-pair output skew (2)	All outputs terminated with 100 $\Omega$ . See Figure 7	50		ps

(1) All typical values are at 25°C and with 2.5-V and 3.3-V supply, unless otherwise noted.

(2) t<sub>SK(O)</sub> is the magnitude of the time difference between the channels within a Port. For more information, see SN65LVCP114 Guidelines for Skew Compensation, SLLA323.

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## ELECTRICAL CHARACTERISTICS (V<sub>cc</sub> 3.3 V $\pm$ 5%, 2.5 V $\pm$ 5%) (continued)

over operating conditions range. All parameters are referenced to package pins (unless otherwise noted).

	PARAMETER	TEST CONDITIONS	MIN	TYP <sup>(1)</sup>	MAX	UNIT
t <sub>SK(PP)</sub>	Part-to-part skew <sup>(3)</sup>				100	ps
t <sub>R</sub>	Rise time	Input signal with 30ps rise time, 20% to 80%. See Figure 6		31		ps
t <sub>F</sub>	Fall time	Input signal with 30ps fall time, 20% to 80%. See Figure 6		31		ps
SDD22	Differential output return loss	100 MHz to 7.1GHz		-10		dB
SCC22	Common-mode output return loss	100 MHz to 7.1GHz		-5		dB
t <sub>sM</sub>		Mux to valid output (idle outputs are squelched)		100		ns
t <sub>SM1</sub>	Multiplexer switch time	Mux to valid output (idle outputs are turned off)		10		μs
Ch	Channel-to-channel isolation <sup>(4)</sup>	Frequency at 5.1625 GHz		52.2		٦٢
Ch <sub>iso</sub>	Channel-to-channel isolation ??	Frequency at 7.1GHz	43.5			dB
OUT		10 MHz to 7.1 GHz. No other noise source present. $V_{\text{OD}}$ = LOW			1500	
OUT <sub>NOISE</sub>	Output referred noise	10 MHz to 7.1 GHz. No other noise source present. $V_{\text{OD}}$ = HIGH		30	3000	μV <sub>RN</sub>
Vpre	Output pre-cursor pre-emphasis	Input signal with 3.75 pre-cursor and measured on the output signal. See Figure 8. Vpre = 20 log(V3/V2)		5		dB
Vpst	Output post-cursor pre-emphasis	Input signal with 12 dB post-cursor and measure it on the output signal. See Figure 8. Vpst = 20 log(V1/V2)		14		dB
r <sub>ot</sub>	Single-ended output resistance	Single-ended on-chip terminations to VCC, outputs are ac-coupled		50		Ω
r <sub>om</sub>	Output termination mismatch at 1 MHz	$\Delta rom = 2 \times \frac{rp - rn}{rp + rn} \times 100$			5	%
EQUALIZAT	ION					
EQ <sub>Gain</sub>	At 7.1 GHz input signal	Equalization gain, EQ = MAX	10	15		dB
DJ1	TX residual deterministic jitter at 10.3125 Gbps	Tx launch amplitude = $0.6$ Vpp, EQ= $1.3$ dB, VOD and GAIN are High. Test Channel = $0$ ". See Figure 10.		0.08		Ulp-p
DJ2	TX residual deterministic jitter at 14.2 Gbps	Tx launch Amplitude = 0.6 Vpp, EQ=1.3dB, VOD and GAIN are High. Test Channel = 0". See Figure 10.		0.06		Ulp-
DJ3	RX residual deterministic jitter at 10.3125 Gbps	Tx Launch Amplitude = 0.6 Vpp, test channel = $12^{\circ}$ (9dB loss at 5GHz), EQ=13.9dB, VOD and GAIN are High. See Figure 9.		0.04		Ulp-
DJ4	RX residual deterministic Jitter at 14.2 Gbps	Tx Launch Amplitude = 0.6 Vpp, test channel = 8" (9dB loss at 7GHz), EQ=13.9dB, VOD=LOW and GAIN=HIGH. See Figure 9.		0.08		Ulp-j

(3) t<sub>SK(PP)</sub> is the magnitude of the difference in propagation delay times between any specified terminals of two devices when both devices operate with the same supply voltages, at the same temperature, and have identical packages and test circuits.

(4) All noise sources added.

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					Maximun Bo	oard Temperatu	re <sup>(1)</sup>	
LOOP_A	LOOP_B	LOOP_C	DIAG	V <sub>cc</sub> =	= 2.5V	$V_{CC} = 3.3V$		
				V <sub>OD</sub> = LOW	V <sub>OD</sub> = HIGH	V <sub>OD</sub> = LOW	V <sub>OD</sub> = HIGH	
LOW	LOW	LOW	LOW	85°C	85°C	85°C	75°C	
LOW	LOW	LOW	HIGH	85°C	85°C	75°C	System Specific (2)	
LOW	LOW	HIGH	LOW	85°C	85°C	85°C	75°C	
LOW	LOW	HIGH	HIGH	85°C	85°C	85°C	75°C	
LOW	HIGH	LOW	LOW	85°C	85°C	75°C	System Specific <sup>(2)</sup>	
LOW	HIGH	LOW	HIGH	85°C	85°C	75°C	System Specific <sup>(2)</sup>	
LOW	HIGH	HIGH	LOW	85°C	85°C	75°C	System Specific <sup>(2)</sup>	
LOW	HIGH	HIGH	HIGH	85°C	85°C	75°C	System Specific <sup>(2)</sup>	
HIGH	LOW	LOW	LOW	85°C	85°C	85°C	75°C	
HIGH	LOW	LOW	HIGH	85°C	85°C	75°C	System Specific <sup>(2)</sup>	
HIGH	LOW	HIGH	LOW	85°C	85°C	85°C	75°C	
HIGH	LOW	HIGH	HIGH	85°C	85°C	75°C	System Specific <sup>(2)</sup>	
HIGH	HIGH	LOW	LOW	85°C	85°C	75°C	System Specific <sup>(2)</sup>	
HIGH	HIGH	LOW	HIGH	85°C	85°C	75°C	System Specific <sup>(2)</sup>	
HIGH	HIGH	HIGH	LOW	85°C	85°C	75°C	System Specific <sup>(2)</sup>	
HIGH	HIGH	HIGH	HIGH	85°C	85°C	75°C	System Specific <sup>(2)</sup>	

### Table 1. RECOMMENDED MAXIMUM BOARD TEMPERATURE

(1)

Maximum board temperature is allowed as long as the device maximum junction temperature is not exceeded. Texas Instruments recommends a system thermal and device use case power analysis to decide possible use of a heat sink. (2)

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### PARAMETER MEASUREMENT INFORMATION

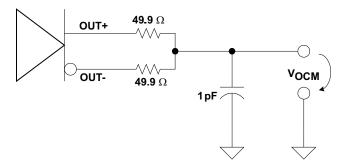


Figure 4. Common-Mode Output-Voltage Test Circuit

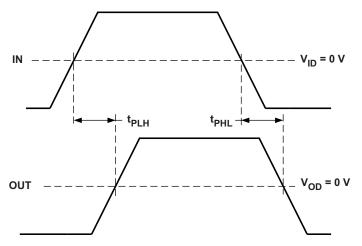


Figure 5. Propagation Delay, Input to Output

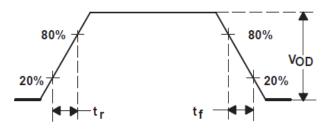


Figure 6. Output Rise and Fall Times

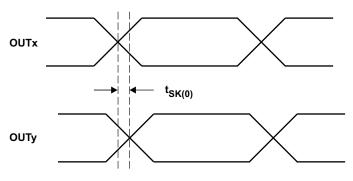


Figure 7. Output Inter-Pair Skew





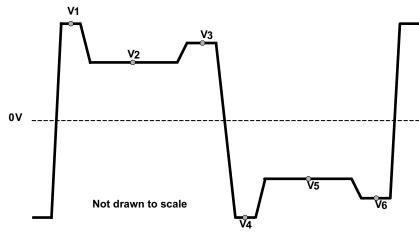


Figure 8. Vpre and Vpost [The Test Pattern is 111111100000000 (Eight 1s, Eight 0s)]

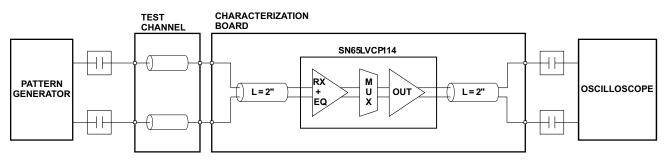


Figure 9. Receive-Side Performance Test Circuit

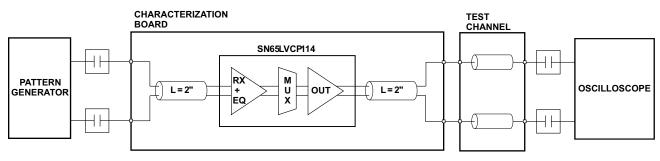


Figure 10. Transmit-Side Performance Test Circuit

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vcc

48 kΩ

48 kΩ

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## PARAMETER MEASUREMENT INFORMATION (continued) EQUIVALENT INPUT AND OUTPUT SCHEMATIC DIAGRAMS

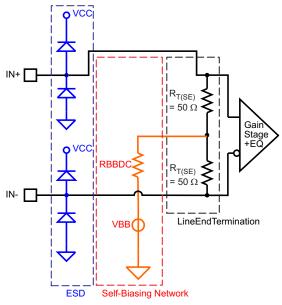




Figure 12. 3-Level Input Biasing Network

vcc

ESD

ESD

## FUNCTIONAL DEFINITIONS

	Table 2. Loopback, DIAG, and SEL Controls									
Loop_A	Loop_B	Loop_C	DIAG	SEL[3:0]	Output Port A	Output Port B	Output Port C			
0	0	0	0	0	In_Port_C[3:0]	idle	In_Port_A[3:0]			
0	0	0	0	1	idle	In_Port_C[3:0]	In_Port_B[3:0]			
0	0	0	1	0	In_Port_C[3:0]	In_Port_C[3:0]	In_Port_A[3:0]			
0	0	0	1	1	In_Port_C[3:0]	In_Port_C[3:0]	In_Port_B[3:0]			
0	0	1	0	0	In_Port_C[3:0]	Idle	In_Port_C[3:0]			
0	0	1	0	1	Idle	In_Port_C[3:0]	In_Port_C[3:0]			
0	0	1	1	0	In_Port_C[3:0]	In_Port_C[3:0]	In_Port_C[3:0]			
0	0	1	1	1	In_Port_C[3:0]	In_Port_C[3:0]	In_Port_C[3:0]			
0	1	0	0	0	In_Port_C[3:0]	In_Port_B[3:0]	In_Port_A[3:0]			
0	1	0	0	1	Idle	In_Port_B[3:0]	In_Port_B[3:0]			
0	1	0	1	0	In_Port_C[3:0]	In_Port_B[3:0]	In_Port_A[3:0]			
0	1	0	1	1	In_Port_C[3:0]	In_Port_B[3:0]	In_Port_B[3:0]			
0	1	1	0	0	In_Port_C[3:0]	In_Port_B[3:0]	In_Port_C[3:0]			
0	1	1	0	1	ldle	In_Port_B[3:0]	In_Port_C[3:0]			
0	1	1	1	0	In_Port_C[3:0]	In_Port_B[3:0]	In_Port_C[3:0]			
0	1	1	1	1	In_Port_C[3:0]	In_Port_B[3:0]	In_Port_C[3:0]			
1	0	0	0	0	In_Port_A[3:0]	Idle	In_Port_A[3:0]			
1	0	0	0	1	In_Port_A[3:0]	In_Port_C[3:0]	In_Port_B[3:0]			
1	0	0	1	0	In_Port_A[3:0]	In_Port_C[3:0]	In_Port_A[3:0]			
1	0	0	1	1	In_Port_A[3:0]	In_Port_C[3:0]	In_Port_B[3:0]			
1	0	1	0	0	In_Port_A[3:0]	Idle	In_Port_C[3:0]			
1	0	1	0	1	In_Port_A[3:0]	In_Port_C[3:0]	In_Port_C[3:0]			
1	0	1	1	0	In_Port_A[3:0]	In_Port_C[3:0]	In_Port_C[3:0]			
1	0	1	1	1	In_Port_A[3:0]	In_Port_C[3:0]	In_Port_C[3:0]			

### Table 2. Loopback, DIAG, and SEL Controls

IN



## PARAMETER MEASUREMENT INFORMATION (continued) Table 2. Loopback, DIAG, and SEL Controls (continued)

-			-				
Loop_A	Loop_B	Loop_C	DIAG	SEL[3:0]	Output Port A	Output Port B	Output Port C
1	1	0	0	0	In_Port_A[3:0]	In_Port_B[3:0]	In_Port_A[3:0]
1	1	0	0	1	In_Port_A[3:0]	In_Port_B[3:0]	In_Port_B[3:0]
1	1	0	1	0	In_Port_A[3:0]	In_Port_B[3:0]	In_Port_A[3:0]
1	1	0	1	1	In_Port_A[3:0]	In_Port_B[3:0]	In_Port_B[3:0]
1	1	1	0	0	In_Port_A[3:0]	In_Port_B[3:0]	In_Port_C[3:0]

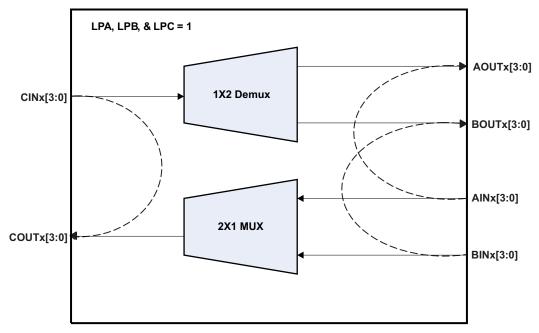


Figure 13. Loopback Mode

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SELx = 0, DIAG = 1 AOUTx[3:0] = CINx[3:0] AOUTx[3:0] 1X2 Demux CINx[3:0] BOUTx[3:0] BOUTx[3:0] = CINx[3:0] COUTx[3:0] = AINx[3:0] ------2X1 MUX SELx=1, DIAG = 1 AOUTx[3:0] = CINx[3:0] 1X2 Demux CINx[3:0] BOUTx[3:0] = CINx[3:0]

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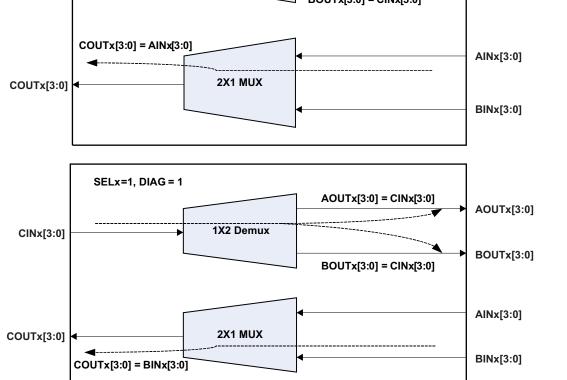


Figure 14. Diagnostic Mode



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## Table 3. Overall Gain Settings

Gain_x	Input Gain [dB]	VOD_x	VOD Gain [dB]	EQ[x]0	EQ[x]1	Total DC gain [dB]	Total EQ gain (7 GHz) [dB]
LOW	6	LOW	1	LOW	LOW	-5	1.3
LOW	6	LOW	1	LOW	HiZ	-5	2
LOW	6	LOW	1	LOW	HIGH	-5	3.6
LOW	6	LOW	1	HiZ	LOW	-8	5
LOW	6	LOW	1	HiZ	HiZ	-8	6.5
LOW	6	LOW	1	HiZ	HIGH	-8	8.3
LOW	6	LOW	1	HIGH	LOW	-11	10
LOW	6	LOW	1	HIGH	HiZ	-11	11.9
LOW	-6	LOW	1	HIGH	HIGH	-11	13.9
LOW	-6	HIGH	6.8	LOW	LOW	0.8	1.3
LOW	-6	HIGH	6.8	LOW	HiZ	0.8	2
LOW	6	HIGH	6.8	LOW	HIGH	0.8	3.6
LOW	6	HIGH	6.8	HiZ	LOW	-2.2	5
LOW	6	HIGH	6.8	HiZ	HiZ	-2.2	6.5
LOW	6	HIGH	6.8	HiZ	HIGH	-2.2	8.3
LOW	6	HIGH	6.8	HIGH	LOW	-5.2	10
LOW	6	HIGH	6.8	HIGH	HiZ	-5.2	11.9
LOW	-6	HIGH	6.8	HIGH	HIGH	-5.2	13.9
HIGH	0	LOW	1	LOW	LOW	1	1.3
HIGH	0	LOW	1	LOW	HiZ	1	2
HIGH	0	LOW	1	LOW	HIGH	1	3.6
HIGH	0	LOW	1	HiZ	LOW	-2	5
HIGH	0	LOW	1	HiZ	HiZ	-2	6.5
HIGH	0	LOW	1	HiZ	HIGH	-2	8.3
HIGH	0	LOW	1	HIGH	LOW	-5	10
HIGH	0	LOW	1	HIGH	HiZ	-5	11.9
HIGH	0	LOW	1	HIGH	HIGH	-5	13.9
HIGH	0	HIGH	6.8	LOW	LOW	6.8	1.3
HIGH	0	HIGH	6.8	LOW	HiZ	6.8	2
HIGH	0	HIGH	6.8	LOW	HIGH	6.8	3.6
HIGH	0	HIGH	6.8	HiZ	LOW	3.8	5
HIGH	0	HIGH	6.8	HiZ	HiZ	3.8	6.5
HIGH	0	HIGH	6.8	HiZ	HIGH	3.8	8.3
HIGH	0	HIGH	6.8	HIGH	LOW	0.8	10
HIGH	0	HIGH	6.8	HIGH	HiZ	0.8	11.9
HIGH	0	HIGH	6.8	HIGH	HIGH	0.8	13.9

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## **TYPICAL CHARACTERISTICS**

Typical operating condition is at  $V_{CC}$  = 2.5 V and  $T_A$  = 25°C, no interconnect line at the output, and with default device settings (unless otherwise noted).

EQ = 1.3 dB,

EQ = 13.9 dB, Gain = 1, VOD = 1.2

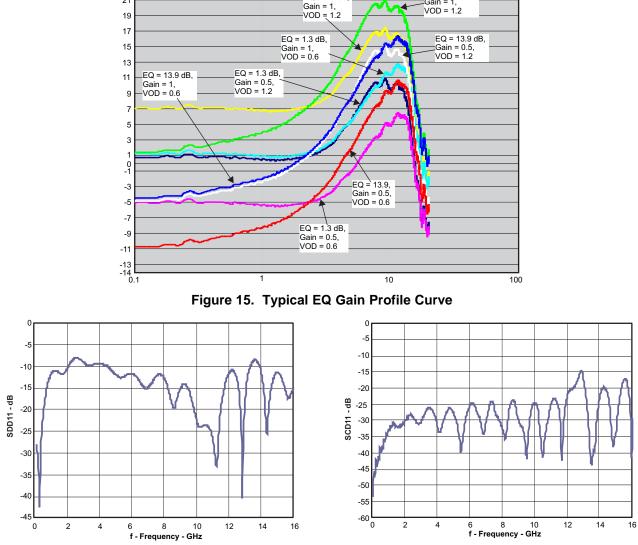


Figure 16. Differential Input Return Loss

Figure 17. Differential to Common-Mode Conversion

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### **TYPICAL CHARACTERISTICS (continued)**

Typical operating condition is at  $V_{CC} = 2.5$  V and  $T_A = 25$ °C, no interconnect line at the output, and with default device settings (unless otherwise noted).

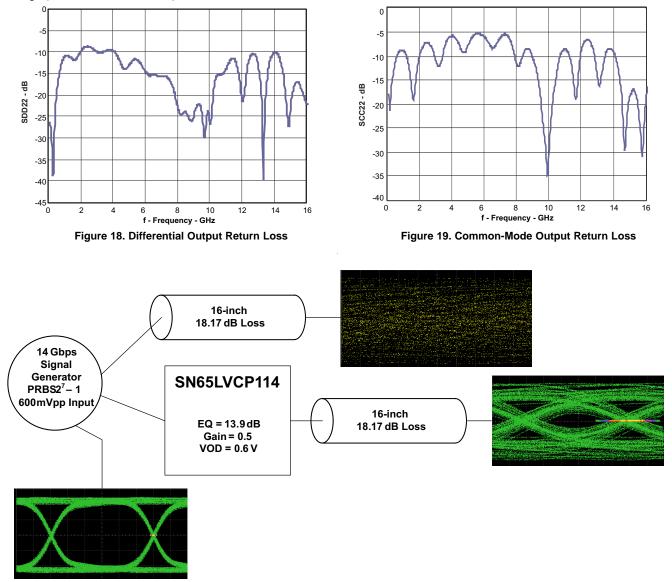


Figure 20. Transmit-Side Typical Application

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### **TYPICAL CHARACTERISTICS (continued)**

Typical operating condition is at  $V_{CC} = 2.5$  V and  $T_A = 25^{\circ}$ C, no interconnect line at the output, and with default device settings (unless otherwise noted).

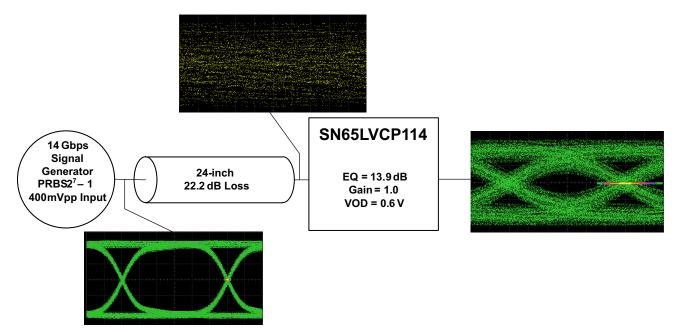


Figure 21. Receive-Side Typical Application



### APPLICATION INFORMATION

### TWO-WIRE SERIAL INTERFACE AND CONTROL LOGIC

The SN65LVCP114 uses a 2-wire serial interface for digital control. The two circuit inputs, SDA and SCL, are driven, respectively, by the serial data and serial clock from a microprocessor, for example. The SDA and SCL pins require external 10-k $\Omega$  pullups to V<sub>CC</sub>.

The 2-wire interface allows write access to the internal memory map to modify control registers and read access to read out the control signals. The SN65LVCP114 is a slave device only, which means that it cannot initiate a transmission itself; it always relies on the availability of the SCL signal for the duration of the transmission. The master device provides the clock signal as well as the START and STOP commands. The protocol for a data transmission is as follows:

- 1. START command
- 7-bit slave address (0000ADD[2:0]) followed by an 8th bit which is the data direction bit (R/W). A zero indicates a WRITE and a 1 indicates a READ. The ADD[2:0] address bits change with the status of the ADD2, ADD1, and ADD0 device pins, respectively. If the pins are left floating or pulled down, the 7-bit slave address is 0000000.
- 3. 8-bit register address
- 4. 8-bit register data word
- 5. STOP command

Regarding timing, the SN65LVCP114 is I<sup>2</sup>C compatible. The typical timing is shown in Figure 22, and a complete data transfer is shown in Figure 23. Parameters for Figure 22 are defined in Table 4.

Bus Idle: Both SDA and SCL lines remain HIGH

Start data transfer: A change in the state of the SDA line, from HIGH to LOW, while the SCL line is HIGH, defines a START condition (S). Each data transfer is initiated with a START condition.

**Stop Data Transfer:** A change in the state of the SDA line from LOW to HIGH while the SCL line is HIGH defines a STOP condition (P). Each data transfer is terminated with a STOP condition; however, if the master still must communicate on the bus, it can generate a repeated START condition and address another slave without first generating a STOP condition.

**Data Transfer:** The number of data bytes transferred between a START and a STOP condition is not limited and is determined by the master device. The receiver acknowledges the transfer of data.

**Acknowledge:** Each receiving device, when addressed, is obliged to generate an acknowledge bit. The transmitter releases the SDA line, and a device that acknowledges must pull down the SDA line during the acknowledge clock pulse in such a way that the SDA line is stable LOW during the HIGH period of the acknowledge clock pulse. Setup and hold times must be taken into account. When a slave-receiver does not acknowledge the slave address, the data line must be left HIGH by the slave. The master can then generate a STOP condition to abort the transfer. If the slave-receiver does acknowledge the slave address but some time later in the transfer cannot receive any more data bytes, the master must abort the transfer. This is indicated by the slave generating the not acknowledge on the first byte to follow. The slave leaves the data line HIGH and the master generates the STOP condition.

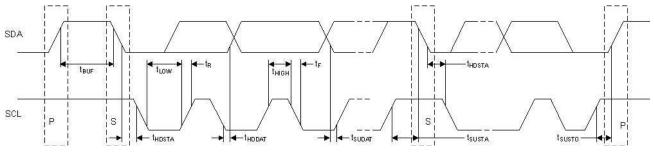


Figure 22. Two-Wire Serial Interface Timing Diagram

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SYMBOL	PARAMETER	MIN	MAX	UNIT
f <sub>SCL</sub>	SCL clock frequency		400	kHz
t <sub>BUF</sub>	Bus free time between START and STOP conditions	1.3		μs
t <sub>HDSTA</sub>	Hold time after repeated START condition. After this period, the first clock pulse is generated.	0.6		μs
t <sub>LOW</sub>	Low period of the SCL clock	1.3		μs
t <sub>HIGH</sub>	High period of the SCL clock	0.6		μs
t <sub>SUSTA</sub>	Setup time for a repeated START condition	0.6		μs
t <sub>HDDAT</sub>	Data hold time	0		μs
t <sub>SUDAT</sub>	Data setup time	100		ns
t <sub>R</sub>	Rise time of both SDA and SCL signals		300	ns
t <sub>F</sub>	Fall time of both SDA and SCL signals		300	ns
t <sub>susto</sub>	Setup time for STOP condition	0.6		μs



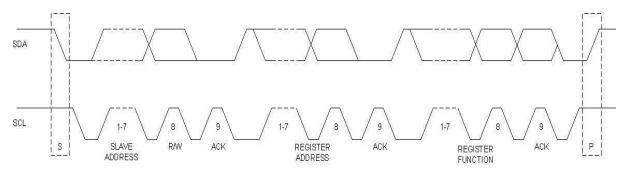


Figure 23. Two-wire Serial Interface Data Transfer

## SN65LVCP114 Register Mapping Information

Register 0x00 (	Register 0x00 (General Device Settings) R/W											
Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0					
SW_GPIO	PWRDOWN						RSVD					

Register 0x01 (Device Control Settings) R/W											
Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0				
DIAG	LOOP[C]	LOOP[B]	LOOP[A]	SEL[3]	SEL[2]	SEL[1]	SEL[0]				

Register 0x02 (Port A Control Settings) R/W										
Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0			
OUT_DIS_0	OUT_DIS_1	OUT_DIS_2	OUT_DIS_3	FAST_SW	RSVD		DIS_AGC			

Register 0x03 (Port A Input Settings) R/W									
Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0		
INOFF	RSVD	RSVD	RSVD	EQ3	EQ2	EQ1	EQ0		

Register 0x04 (Port A Output Settings) R/W										
Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0			
		VOD1	VOD0			GAIN1	GAIN0			

Register 0x06 (Port B Control Settings) R/W									
Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0		
OUT_DIS_0	OUT_DIS_1	OUT_DIS_2	OUT_DIS_3	FAST_SW	RSVD		DIS_AGC		



Register 0x07 (Port B Input Settings) R/W Bit 7 Bit 6 Bit 5 Bit 4 Bit 3 Bit 2 Bit 1 Bit 0 INOFF RSVD RSVD RSVD EQ0 EQ3 EQ2 EQ1 Register 0x08 (Port B Output Settings) R/W Bit 4 Bit 0 Bit 7 Bit 6 Bit 5 Bit 3 Bit 2 Bit 1 VOD1 VOD0 GAIN1 GAIN0 Register 0x0A (Port C Control Settings) R/W Bit 7 Bit 6 Bit 5 Bit 4 Bit 3 Bit 2 Bit 1 Bit 0 OUT\_DIS\_0 OUT\_DIS\_1 OUT\_DIS\_2 OUT\_DIS\_3 FAST\_SW DIS\_AGC RSVD Register 0x0B (Port C Input Settings) R/W Bit 7 Bit 6 Bit 5 Bit 4 Bit 3 Bit 2 Bit 1 Bit 0 RSVD INOFF RSVD RSVD EQ3 EQ1 EQ0 EQ2 \_ ~ ~ ..... 0v0C /D 2

Register 0x0C (	Port C Output Se	ettings) R/W					
Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
		VOD1	VOD0			GAIN1	GAIN0

Register 0x0D (Reserved Settings) R/W										
Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0			
RSVD	RSVD	RSVD	RSVD	RSVD	RSVD	RSVD	RSVD			

Register 0x0F (Reserved Settings) Read Only											
Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0				
RSVD	RSVD	RSVD	RSVD	RSVD	RSVD	RSVD	RSVD				

Register 0x10 (Polarity Control Settings for Port A and B) R/W									
Bit 7         Bit 6         Bit 5         Bit 4         Bit 3         Bit 2         Bit 1         Bit 0									
POL_B[3]	POL_B[2]	POL_B[1]	POL_B[0]	POL_A[3]	POL_A[2]	POL_A[1]	POL_A[0]		

Register 0x11 (Polarity Control Settings for Port C) R/W										
Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0			
				POL_C[3]	POL_C[2]	POL_C[1]	POL_C[0]			

Register 0x12 (Lane Enable) R/W										
Bit 7         Bit 6         Bit 5         Bit 4         Bit 3         Bit 2         Bit 1         Bit 0										
				LN_EN[3]	LN_EN[2]	LN_EN[1]	LN_EN[0]			

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REGISTER	BIT	SYMBOL	FUNCTION	DEFAULT
	7	SW_GPIO	Switching logic is controlled by GPIO or $I^2C$ : 1 = GPIO control 0 = $I^2C$ control	
	6	PWRDOWN	Power down the device: 0 = Normal operation 1 = Power down	
0x00	5			00000000
	4			
	3			
	2			
	1			
	0	RSVD	For TI use only	
	7	DIAG	Enables Diag Mode: 0 = Disable 1 = Enable	
	6	LOOP[C]	Enables port C loopback: 0 = Disable 1 = Enable	
	5	LOOP[B]	Enables port B loopback: 0 = Disable 1 = Enable	
0.01	4	LOOP[A]	Enables port A loopback: 0 = Disable 1 = Enable	
0x01	3	SEL[3]	Lane 3, port A/port B switch control: 0 = Port A selected 1 = Port B selected	
	2	SEL[2]	Lane 2, port A/port B switch control: 0 = Port A selected 1 = Port B selected	
	1	SEL[1]	Lane 1, port A/port B switch control: 0 = Port A selected 1 = Port B selected	
	0	SEL[0]	Lane 0, port A/port B switch control: 0 = Port A selected 1 = Port B selected	



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Table 5. SN65LVCP114 Register Descriptions (continued
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REGISTER	BIT	SYMBOL		FUNCTION										
	7	OUT_DIS0	Disables 0 = Enal 1 = Disa		0:									
	6	OUT_DIS1	Disables 0 = Enal 1 = Disa		1:					_				
	5	OUT_DIS2	Disables 0 = Enal 1 = Disa											
0x02 0x06 0x0A	4	OUT_DIS3	0 = Enal	Disables output lane 3: 0 = Enable 1 = Disable										
	3	FAST_SW	0 = Idle	Fast switch: 0 = Idle outputs are disabled (save power) 1 = Idle outputs are squelched (fast switch time)										
	2	RSVD	For TI u	For TI use only										
	1													
	0	DIS_AGC	0 = Enal	GC loop: = Enable = Disable										
-	7	IN_OFF	0 = Norr	ower down input stages: = Normal = Power down										
	6	RSVD	For TI u											
	5	RSVD	For TI u											
	4	RSVD	For TI u	e only										
	3	EQ3		EQ3	EQ2	EQ1	EQ0	Peaking in dB						
0x03	2	EQ2		0	х	х	x	1.3						
0x07	1	EQ1		1	0	0	0	2	000000	000				
0x0B				1	0	0	1	3.6						
				1	0	1	0	5						
				1	0	1	1	6.5						
			-	1	1	0	0	8.3						
				1	1	0	1	10						
				1	1	1	0	11.9						
	0	EQ0		1	1	1	1	13.9						
	7		1		Ļ	-	+	I						
	6													
	5	VOD1	VOD co	ntrol [VOD1:\	/OD0]:									
0x04	4	VOD0	00 = 120 01 = 600 10 = 120	00 mV maxim 0 mV maximu 00 mV maxim 00 mV maxim	ium im ium									
0x08 0x0C	3								000000	000				
	2													
	1	GAIN1	GAIN co	ontrol [GAIN1	:GAIN01:									
	0	GAINO	00 = 0.5 $01 = 1$ $10 = 0.5$ $11 = 1$		-1-									

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REGISTER	BIT	SYMBOL	FUNCTION	DEFAULT
	7			
	6			
	5			
EGISTER	4			0000000
	3			
	2			
	1			
	0			
	7	RSVD	For TI use only	
	6	RSVD	For TI use only	
	5	RSVD	For TI use only	
0x0D	4	RSVD	For TI use only	0000000
	3	RSVD	For TI use only	
	2	RSVD	For TI use only	
	1	RSVD	For TI use only	
	0	RSVD	For TI use only	
	7	RSVD	For TI use only	
	6	RSVD	For TI use only	
0v0F	5	RSVD	For TI use only	
	4	RSVD	For TI use only	00010001
0,01	3	RSVD	For TI use only	
	2	RSVD	For TI use only	
	1	RSVD	For TI use only	
	0	RSVD	For TI use only	
	7	POL_B[3]	Polarity switch of output lane 3 of port B: 0 = Normal 1 = Switched	
	6	POL_B[2]	Polarity switch of output lane 2 of port B: 0 = Normal 1 = Switched	
	5	POL_B[1]	Polarity switch of output lane 1 of port B: 0 = Normal 1 = Switched	
010	4	POL_B[0]	Polarity switch of output lane 0 of port B 0 = Normal 1 = Switched	0000000
UXTU	3	POL_A[3]	Polarity switch of output lane 3 of port A: 0 = Normal 1 = Switched	
	2	POL_A[2]	Polarity switch of output lane 2 of port A: 0 = Normal 1 = Switched	
	1	POL_A[1]	Polarity switch of output lane 1 of port A: 0 = Normal 1 = Switched	
	0	POL_A[0]	Polarity switch of output lane 0 of port A: 0 = Normal 1 = Switched	

### Table 5. SN65LVCP114 Register Descriptions (continued)

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REGISTER	BIT	SYMBOL	FUNCTION	DEFAULT
	7			
	6			
	5			
	4			
	3	POL_C[3]	Polarity switch of output lane 3 of port C: 0 = Normal 1 = Switched	
0x11	2	POL_C[2]	Polarity switch of output lane 2 of port C: 0 = Normal 1 = Switched	0000000
	1	POL_C[1]	Polarity switch of output lane 1 of port C: 0 = Normal 1 = Switched	
	0	POL_C[0]	Polarity switch of output lane 0 of port C: 0 = Normal 1 = Switched	
	7			
	6			
	5			
	4			
	3	LN_EN_3	Lane 3 of ports A, B, and C: 0 = Disable 1 = Enable	
0x12	2 LN_EN_2 Lane 2 of ports A, B, and C: 0 = Disable 1 = Enable		0 = Disable	00001111
	1	LN_EN_1	Lane 1 of ports A, B, and C: 0 = Disable 1 = Enable	
	0	LN_EN_0	Lane 0 of ports A, B, and C: 0 = Disable 1 = Enable	

## Table 5. SN65LVCP114 Register Descriptions (continued)



11-Apr-2013

## PACKAGING INFORMATION

Orderable Device	Status	Package Type	Package	Pins	Package	Eco Plan	Lead/Ball Finish	MSL Peak Temp	Op Temp (°C)	Top-Side Markings	Samples
	(1)		Drawing		Qty	(2)		(3)		(4)	
SN65LVCP114ZJA	ACTIVE	NFBGA	ZJA	167	189	Green (RoHS & no Sb/Br)	SNAGCU	Level-3-260C-168 HR	-40 to 85	SN65LVCP114	Samples

<sup>(1)</sup> The marketing status values are defined as follows:

**ACTIVE:** Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

**PREVIEW:** Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

<sup>(2)</sup> Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.

**TBD:** The Pb-Free/Green conversion plan has not been defined.

**Pb-Free (RoHS):** TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes. **Pb-Free (RoHS Exempt):** This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

<sup>(3)</sup> MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

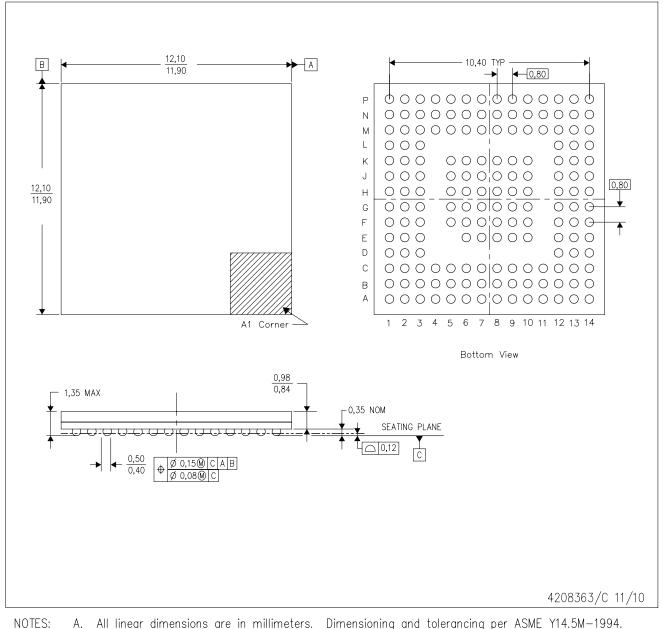
(4) Multiple Top-Side Markings will be inside parentheses. Only one Top-Side Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Top-Side Marking for that device.

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ZJA (S-PBGA-N167)

PLASTIC BALL GRID ARRAY



A. All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M-1994.
 B. This drawing is subject to change without notice.

C. This is a Pb-free solder ball design.



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