

5V/12V eFuse with Over Voltage Protection and Blocking FET Control

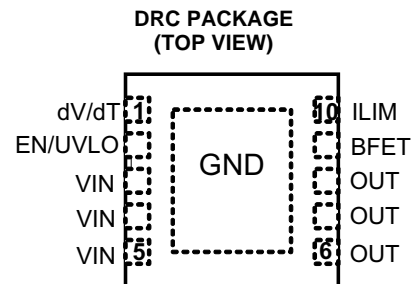
Check for Samples: [TPS2592Ax](#), [TPS2592Bx](#)

FEATURES

- 12 V Protection – TPS2592Ax
- 5 V Protection – TPS2592Bx
- Integrated 28mΩ Pass MOSFET
- Programmable Current Limit ($\pm 15\%$ Accuracy)
- Blocking FET Driver
- Fixed Over Voltage Setting
- Programmable Vout Slew Rate, UVLO
- Built-in Thermal Shutdown
 - Latched-off TPS2592AL, TPS2592BL
 - Auto-Retry TPS2592AA, TPS2592BA
- Applied for UL Recognition
- Small Foot Print—10L (3mm x 3mm) VSON Package

APPLICATIONS

- HDD and SSD Drives
- Set Top Boxes
- Servers / AUX Supplies
- Fan Control
- PCI/PCIe Cards
- Switches/Routers



DESCRIPTION

The TPS2592x family of eFuses are highly integrated circuit protection and power management solutions in a tiny package. With few external components and multiple protection modes they are a robust defense against overloads, shorts circuits, voltages surges, excessive inrush current, and reverse current. Only one external resistor is required for setting the current limit level, which has a typical accuracy of $\pm 15\%$. Over voltage events are limited by internal clamping circuits to a safe fixed maximum with no external components required. TPS2592Ax devices provide over voltage protection (OVP) for 12 V systems and TPS2592Bx devices for 5 V systems. In cases with particular voltage ramp requirements, a dV/dT pin is provided that can be programmed with a single capacitor to ensure proper output ramp rates. Many systems, such as SSDs, must not allow holdup capacitance energy to dump back through the FET body diode onto a drooping or shorted bus. The BFET pin is for just such systems. An external NFET can be connected “back to back” with the TPS2592xx output and the gate driven by BFET. When the TPS2592xx is disabled then current flow is stopped in both directions. TPS2592xL parts will latch off after a fault and TPS2592xA parts will attempt to restart after the thermal shutoff resets.

PRODUCT PREVIEW

PRODUCT INFORMATION

PART NUMBER	Vin OPERATING RANGE	UVLO, OVERVOLTAGE CLAMP (TYP)	THERMAL FAULT	STATUS
TPS2592AA	4.5 V – 13.8 V	4.3 V, 15.0 V	Auto	Active
TPS2592BA	4.5 V – 5.5 V	4.3 V, 6.1 V	Auto	Preview
TPS2592AL	4.5 V – 13.8 V	4.3V, 15.0 V	Latched	Preview
TPS2592BL	4.5 V – 5.5 V	4.3 V, 6.1 V	Latched	Active

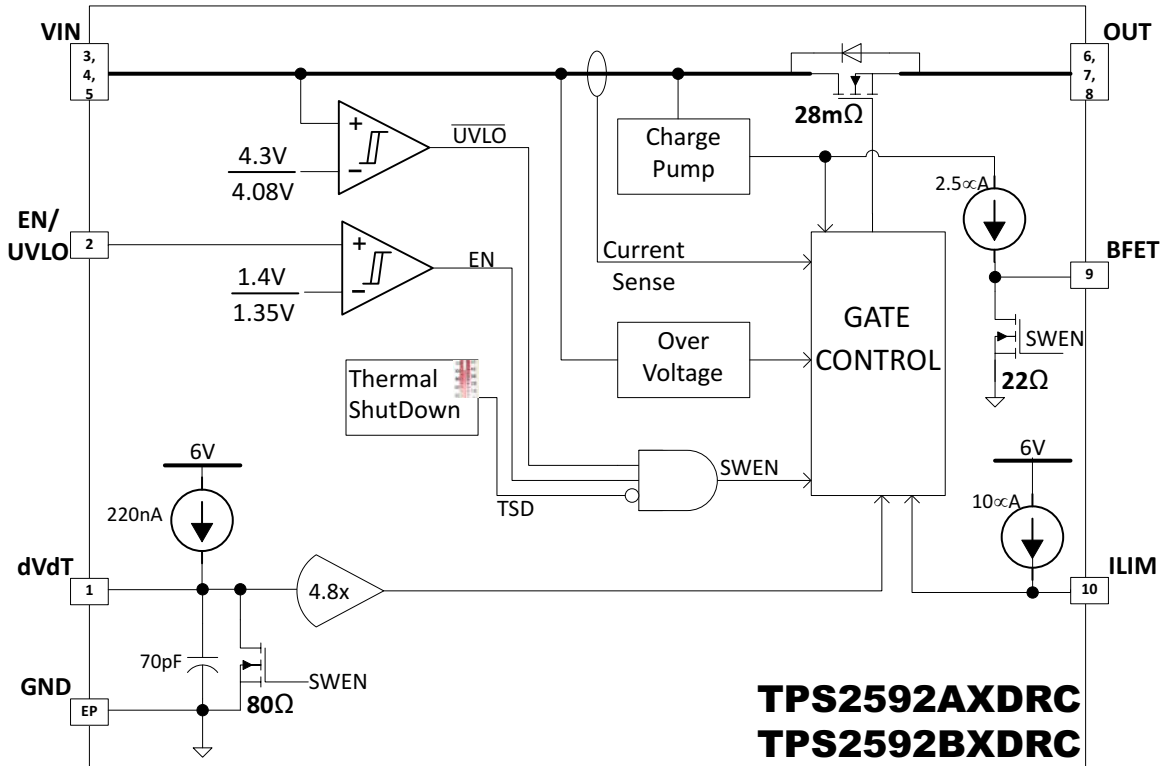


Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.



These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

BLOCK DIAGRAM



TPS2592AXDRC
TPS2592BXDRC

PIN DESCRIPTIONS

PIN		DESCRIPTION
NAME	NO.	
SUPPLY PINS		
VIN	3-5	Input Supply Voltage
GND	Power Pad	GND
CONTROL PINS		
dVdT	1	Tie a capacitor from this pin to GND to control the ramp of VOUT at device turn-on
EN/UVLO	2	This is a dual function control pin. When used as an ENABLE pin and pulled down, it shuts off the internal pass MOSFET and pulls BFET to GND. When pulled high, it enables the device and BFET. As an UVLO pin, it can be used to program different UVLO trip point via external resistor divider
BFET	9	Connect this pin to the gate of a blocking NFET. Charging and discharging of external FET is controlled by EN/UVLO pin and internal UVLO on VIN whichever triggers the threshold first. See detailed description and application note in this datasheet
ILIM	10	A resistor from this pin to GND will set the overload and short circuit limit
LOAD PINS		
OUT	6-8	Output of the device

PRODUCT PREVIEW

ORDERING INFORMATION⁽¹⁾

PART NUMBER	PART MARKING	PACKAGE
TPS2592ALDRC	2592AL	10-pin DRC
TPS2592AADRC	2592AA	10-pin DRC
TPS2592BLDRC	2592BL	10-pin DRC
TPS2592BADRC	2592BA	10-pin DRC

(1) For the most current package and ordering information, see the Package Option Addendum at the end of this document, or see the TI web site at www.ti.com.

ABSOLUTE MAXIMUM RATINGS

over operating free-air temperature range (unless otherwise noted) ⁽¹⁾

		VALUE ⁽²⁾		UNIT
		MIN	MAX	
Supply voltage range ⁽³⁾	VIN	-0.3	22	V
Output voltage	VOUT	-0.3	VIN + 0.3	V
ILIM		-0.3	7	V
EN/UVLO		-0.3	7	V
dVdT		-0.3	7	V
BFET		-0.3	30	V
Electrostatic discharge	Human body model ⁽⁴⁾		±2000	V
	Charged-device model ⁽⁵⁾		±500	V
Continuous power dissipation		See Dissipation Rating Table		

- (1) Stresses beyond those listed under absolute maximum ratings may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these or any conditions beyond those indicated under recommended operating conditions is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.
- (2) All voltage values, except differential voltages, are with respect to network ground terminal.
- (3) Tested in accordance with JEDEC Standard 22, Test Method A114-B
- (4) Tested in accordance with JEDEC Standard 22, Test Method C101-A
- (5) Tested in accordance with JEDEC Standard 22, Test Method A115-A

THERMAL CHARACTERISTICS⁽¹⁾

THERMAL METRIC		TPS2592xx	UNIT
		DRC (10) PINS	
θ_{JA}	Junction-to-ambient thermal resistance	45.9	°C/W
θ_{JcTop}	Junction-to-case (top) thermal resistance	53	
θ_{JB}	Junction-to-board thermal resistance	21.2	
ψ_{JT}	Junction-to-top characterization parameter	1.2	
ψ_{JB}	Junction-to-board characterization parameter	21.4	
θ_{JcBot}	Junction-to-case (bottom) thermal resistance	5.9	

(1) For more information about traditional and new thermal metrics, see the IC Package Thermal Metrics application report, [SPRA953](http://sprae953).

RECOMMENDED OPERATING CONDITIONS

over operating free-air temperature range (unless otherwise noted)

		MIN	TYP	MAX	UNIT
Input voltage range	VIN TPS2592Ax	4.5	12	13.8	V
	VIN TPS2592Bx	4.5	5	5.5	
	BFET	0		VIN+7	
	dVdT, EN/UVLO	0		6	
	ILIM	0		3.3	
Resistance	ILIM	40.2	100	162	kΩ
External capacitance	VOUT	0.1	1	1000	μF
	dVdT		1	1000	nF
Operating junction temperature range, T _J		-40	25	125	°C
Operating Ambient temperature range, T _A		-40	25	85	°C

ELECTRICAL CHARACTERISTICS

-40°C ≤ T_J ≤ 125°C, VIN = 12V for TPS2592A_, VIN = 5V for TPS2592B_, VEN/UVLO = 2V, RILIM = 100kΩ, CdVdT = OPEN. All voltages referenced to GND (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
VIN (INPUT SUPPLY)						
V _{UVR}	UVLO threshold, rising		4.15	4.3	4.45	V
V _{UVhyst}	UVLO hysteresis			5.4%		
I _{QON}	Supply current	Enabled: EN/UVLO = 2V, TPS2592A_	0.2	0.42	0.65	mA
		Enabled: EN/UVLO = 2V, TPS2592B_	0.4	0.62	0.85	mA
I _{QOFF}		EN/UVLO = 0V		0.1	0.25	mA
V _{OVC}	Over-voltage clamp	VIN > 16.5V, I _{VOUT} = 10mA, TPS2592A_	13.8	15	16.5	V
		TPS2592B_, VIN > 6.75V, I _{VOUT} = 10 mA, -40°C ≤ T _J ≤ 85°C	5.5	6.1	6.75	
		TPS2592B_, VIN > 6.75V, I _{VOUT} = 10 mA, -40°C ≤ T _J ≤ 125°C	5.25	6.1	6.75	
EN/UVLO (ENABLE/UVLO INPUT)						
V _{ENR}	EN Threshold voltage, rising		1.37	1.4	1.44	V
V _{ENF}	EN Threshold voltage, falling		1.32	1.35	1.39	V
I _{EN}	EN Input leakage current	0 V ≤ V _{EN} ≤ 5V	-100	0	100	nA
T _{OFFdly}	Turn Off delay	EN↓ to BFET↓, C _{BFET} = 0, t _{pf50-90} . See SLVSAL179		0.4		μs
dVdT (OUTPUT RAMP CONTROL)						
T _{dVdT}	Output ramp time	TPS2592A_, EN/UVLO → H to VOUT = 11.7V, C _{dVdT} = 0	0.7	1	1.3	ms
		TPS2592B_, EN/UVLO → H to VOUT = 4.9V, C _{dVdT} = 0	0.28	0.4	0.52	
		TPS2592A_, EN/UVLO → H to VOUT = 11.7V, C _{dVdT} = 1 nF		12		
		TPS2592B_, EN/UVLO → H to VOUT = 4.9V, C _{dVdT} = 1 nF		5		
I _{dVdT}	dVdT Charging current	V _{dVdT} = 0 V		220		nA
R _{dVdT_disch}	dVdT Discharging resistance	EN/UVLO = 0 V, I _{dVdT} = 10 mA sinking	50	73	100	Ω
V _{dVdTmax}	dVdT Max capacitor voltage			5.5		V
GAIN _{dVdT}	dVdT to VOUT gain	ΔV _{VOUT} /ΔV _{dVdT}		4.85		V/V

ELECTRICAL CHARACTERISTICS (continued)

–40°C ≤ T_J ≤ 125°C, V_{IN} = 12V for TPS2592A_, V_{IN} = 5V for TPS2592B_, V_{EN/UVLO} = 2V, R_{ILIM} = 100kΩ, C_{dVdT} = OPEN. All voltages referenced to GND (unless otherwise noted)

PARAMETER		TEST CONDITIONS	MIN	TYP	MAX	UNIT
ILIM (CURRENT LIMIT PROGRAMMING)						
I _{ILIM}	I _{LIM} Bias current			10		μA
I _{OL}	Overload current limit	R _{ILIM} = 45.3 kΩ, V _{VIN-VOUT} = 1 V	1.81	2.14	2.47	A
		R _{ILIM} = 100 kΩ, V _{VIN-VOUT} = 1 V	3.35	3.75	4.15	
		R _{ILIM} = 150 kΩ, V _{VIN-VOUT} = 1 V	4.4	5.2	6	
I _{OL-R-Short}		R _{ILIM} = 0 Ω, Shorted Resistor Current Limit (Single Point Failure Test: UL60950)		0.7		A
I _{OL-R-Open}	R _{ILIM} = OPEN, Open Resistor Current Limit (Single Point Failure Test: UL60950)		0.55		A	
I _{SC}	Short-circuit current limit	R _{ILIM} = 45.3 kΩ, V _{VIN-VOUT} = 5 V, TPS2592B_	1.76	2.1	2.42	A
		R _{ILIM} = 45.3 kΩ, V _{VIN-VOUT} = 12 V, TPS2592A_	1.7	2.03	2.37	
		R _{ILIM} = 100 kΩ, V _{VIN-VOUT} = 5 V, TPS2592B_	3.15	3.6	4.05	
		R _{ILIM} = 100 kΩ, V _{VIN-VOUT} = 12 V, TPS2592A_	3	3.37	3.73	
		R _{ILIM} = 150 kΩ, V _{VIN-VOUT} = 5 V, TPS2592B_	4.1	4.9	5.7	
		R _{ILIM} = 150 kΩ, V _{VIN-VOUT} = 12 V, TPS2592A_	3.81	4.51	5.21	
RATIO _{FASTTRIP}	Fast-Trip comparator level w.r.t. overload current limit	I _{FASTTRIP: IOL}		150%		
T _{FastOffDly}	Fast-Trip comparator delay	I _{VOUT} > I _{FASTTRIP}		3		μs
V _{OpenILIM}	ILIM Open resistor detect threshold	V _{ILIM} Rising, R _{ILIM} = OPEN		3.1		V
VOUT (PASS FET OUTPUT)						
T _{ON}	Turn-on delay	EN/UVLO → H to I _{VIN} = 100mA, 1A resistive load at V _{OUT}		220		μs
R _{DSon}	FET ON resistance	T _J = 25°C	21	28	33	mΩ
		T _J = 125°C ⁽¹⁾		39	46	
I _{VOUT-OFF-LKG}	V _{OUT} Bias current in off state	V _{EN/UVLO} = 0 V, V _{OUT} = 0 V (Sourcing)	–5	0	1	μA
I _{VOUT-OFF-SINK}		V _{EN/UVLO} = 0V, V _{OUT} = 300 mV (Sinking)	10	15	20	
BFET (BLOCKING FET GATE DRIVER)						
I _{BFET}	BFET Charging current	V _{BFET} = V _{VOUT}		2		μA
V _{BFETmax}	BFET Clamp voltage			V _{VIN+6.4}		V
R _{BFETdisch}	BFET Discharging resistance	V _{EN/UVLO} = 0 V, I _{BFET} = 100 A	15	26	36	Ω
T _{BFET-ON}	BFET Turn-on duration	EN/UVLO → H to V _{BFET} = 12 V, C _{BFET} = 1 nF		4.2		ms
		EN/UVLO → H to V _{BFET} = 12 V, C _{BFET} = 10 nF		42		
T _{BFET-OFF}	BFET Turn-off duration	EN/UVLO → L to V _{BFET} = 1 V, C _{BFET} = 1 nF		0.4		μs
		EN/UVLO → L to V _{BFET} = 1 V, C _{BFET} = 10 nF		1.4		
TSD (THERMAL SHUT DOWN)						
T _{SHDN}	TSD Threshold, rising			160		°C
T _{SHDNhyst}	TSD Hysteresis ⁽¹⁾			10		°C
	Thermal fault: latched or autoretry	TPS2592xL		LATCHED		
		TPS2592xA		AUTO-RETRY		

(1) The limits for these parameters are specified based on characterization data, and are not tested during production.

TYPICAL CHARACTERISTICS

$T_J = 25^\circ\text{C}$, $V_{VIN} = 12\text{ V}$ for TPS2592Ax, $V_{VIN} = 5\text{ V}$ for TPS2592Bx, $V_{EN/UVLO} = 2\text{ V}$, $R_{ILIM} = 100\text{ k}\Omega$, $C_{VIN} = 0.1\text{ }\mu\text{F}$, $C_{OUT} = 1\text{ }\mu\text{F}$, $C_{dVdT} = \text{OPEN}$ (unless stated otherwise)

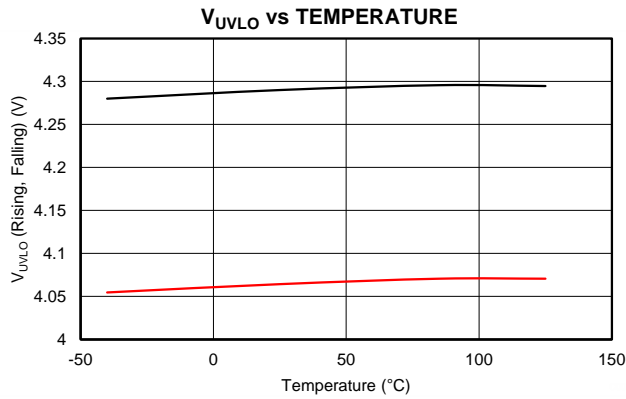


Figure 1.

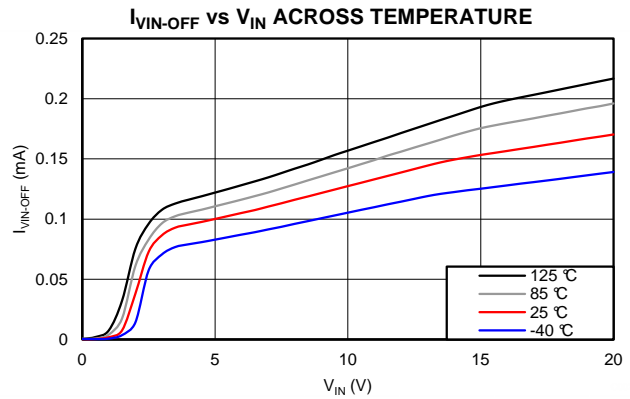


Figure 2.

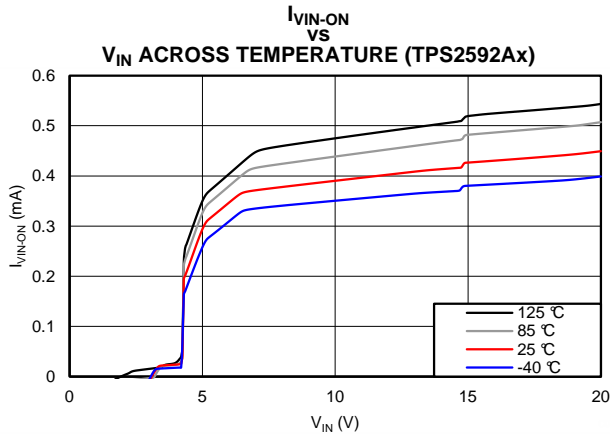


Figure 3.

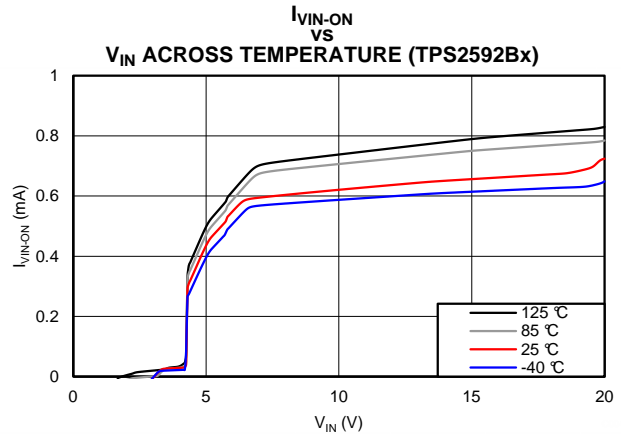


Figure 4.

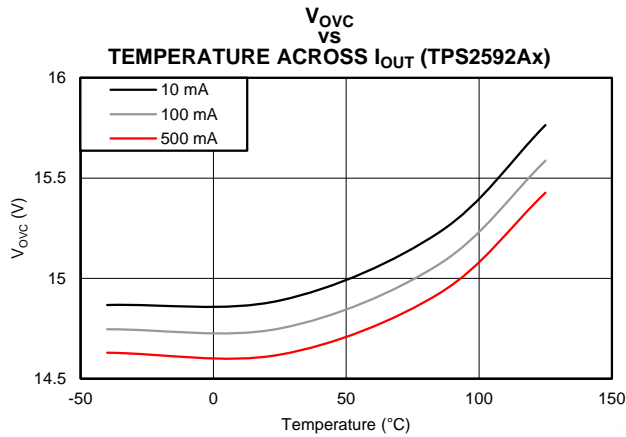


Figure 5.

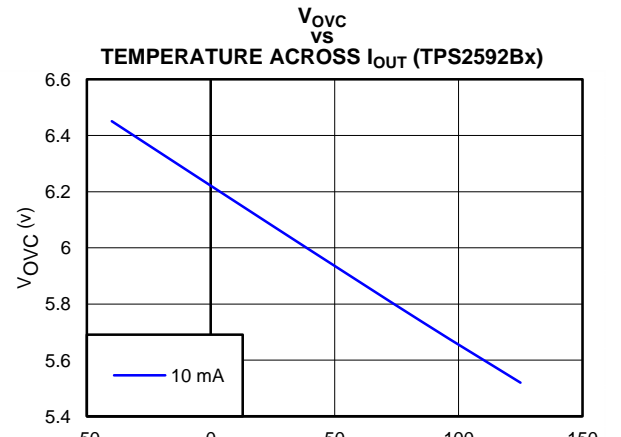


Figure 6.

PRODUCT PREVIEW

TYPICAL CHARACTERISTICS (continued)

$T_J = 25^\circ\text{C}$, $V_{VIN} = 12\text{ V}$ for TPS2592Ax, $V_{VIN} = 5\text{ V}$ for TPS2592Bx, $V_{ENUVLO} = 2\text{ V}$, $R_{LIM} = 100\text{ k}\Omega$, $C_{VIN} = 0.1\text{ }\mu\text{F}$, $C_{OUT} = 1\text{ }\mu\text{F}$, $C_{dVdT} = \text{OPEN}$ (unless stated otherwise)

TRANSIENT: OVER-VOLTAGE CLAMP: TPS2592Ax

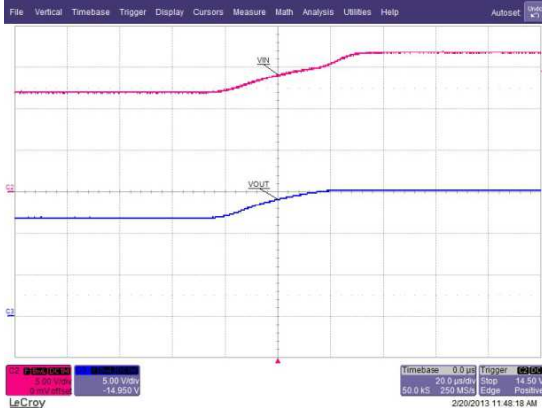


Figure 7.

TRANSIENT: OVER-VOLTAGE CLAMP: TPS2592Bx

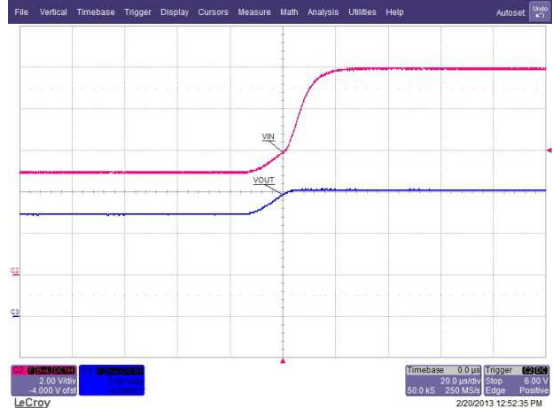


Figure 8.

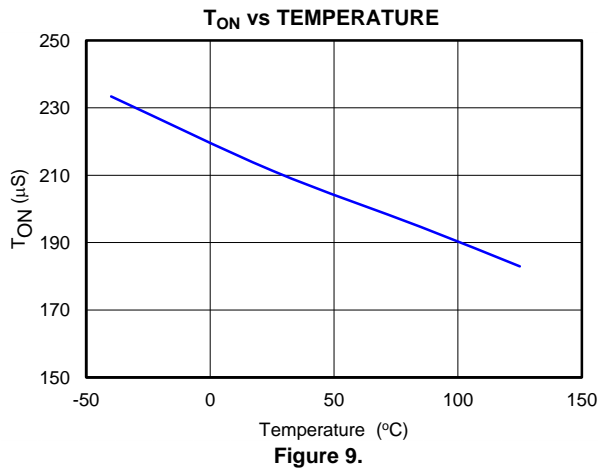


Figure 9.

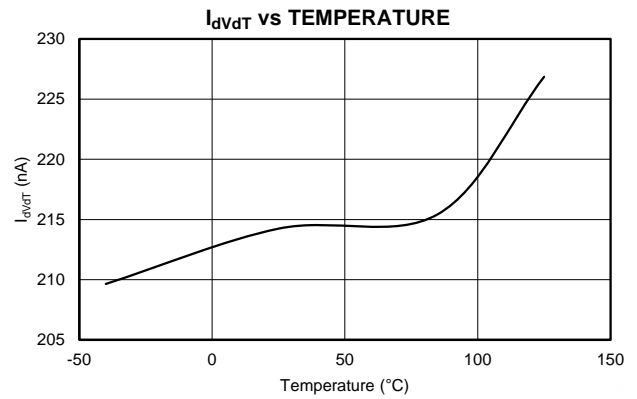


Figure 10.

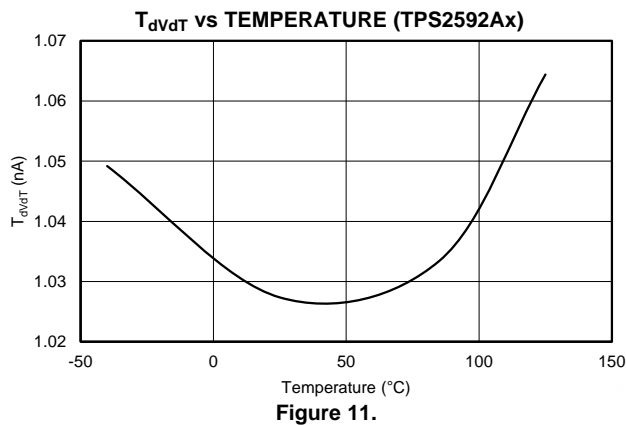


Figure 11.

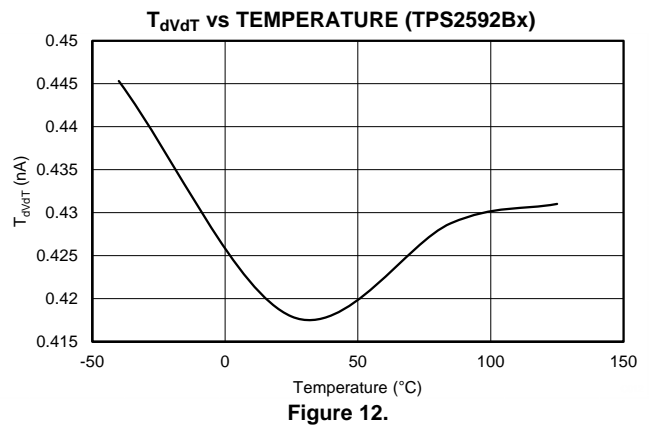


Figure 12.

PRODUCT PREVIEW

TYPICAL CHARACTERISTICS (continued)

$T_J = 25^\circ\text{C}$, $V_{VIN} = 12\text{ V}$ for TPS2592Ax, $V_{VIN} = 5\text{ V}$ for TPS2592Bx, $V_{EN/UVLO} = 2\text{ V}$, $R_{LIM} = 100\text{ k}\Omega$, $C_{VIN} = 0.1\text{ }\mu\text{F}$, $C_{OUT} = 1\text{ }\mu\text{F}$, $C_{dVdT} = \text{OPEN}$ (unless stated otherwise)

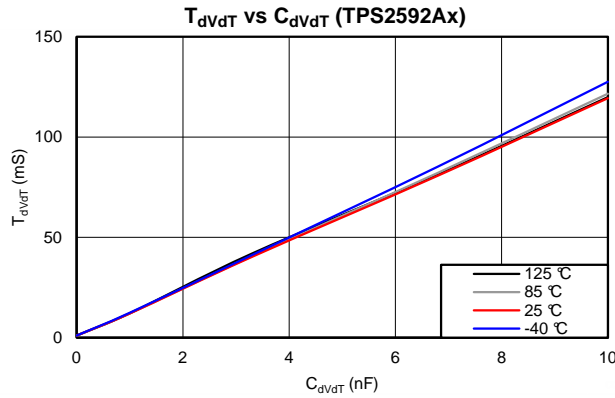


Figure 13.

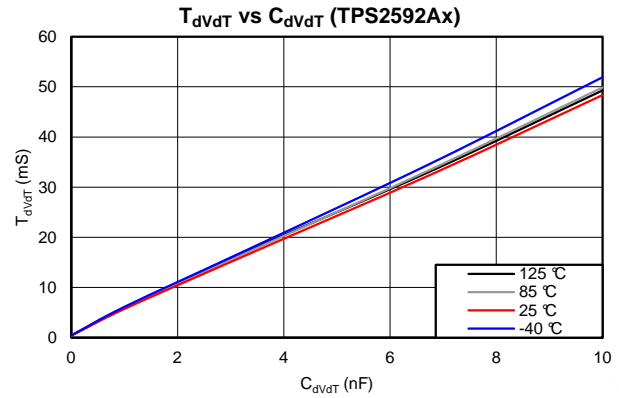


Figure 14.

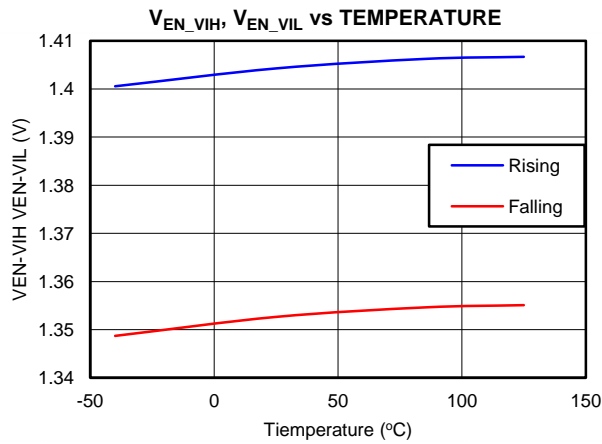


Figure 15.

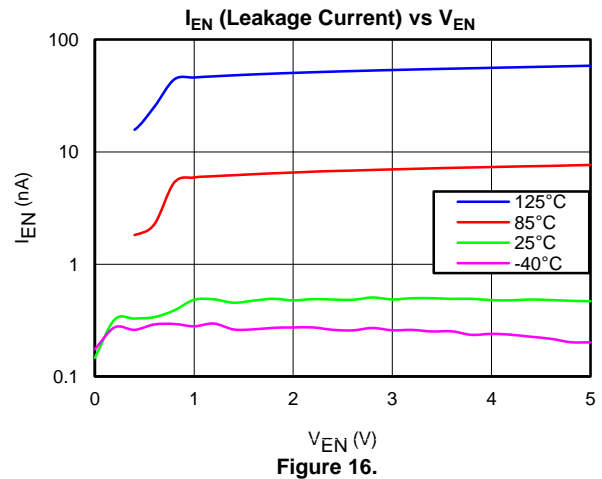


Figure 16.

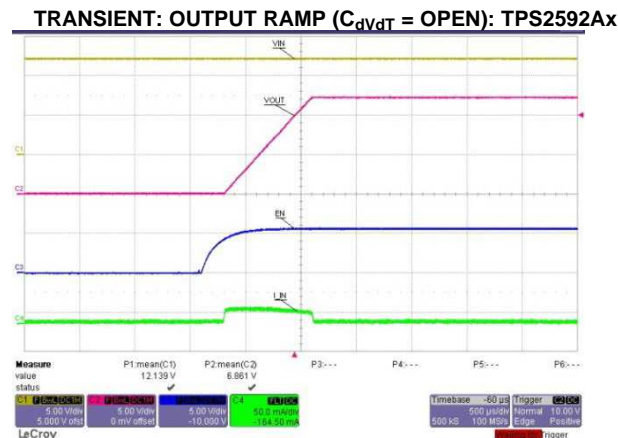


Figure 17.

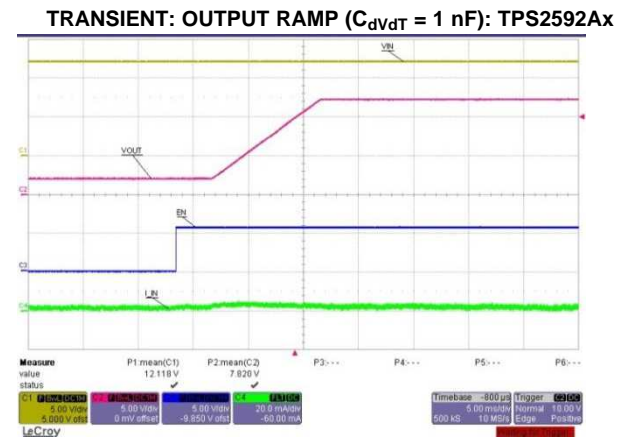


Figure 18.

PRODUCT PREVIEW

TYPICAL CHARACTERISTICS (continued)

$T_J = 25^\circ\text{C}$, $V_{VIN} = 12\text{ V}$ for TPS2592Ax, $V_{VIN} = 5\text{ V}$ for TPS2592Bx, $V_{EN/UVLO} = 2\text{ V}$, $R_{LIM} = 100\text{ k}\Omega$, $C_{VIN} = 0.1\text{ }\mu\text{F}$, $C_{OUT} = 1\text{ }\mu\text{F}$, $C_{dVdT} = \text{OPEN}$ (unless stated otherwise)

TRANSIENT: OUTPUT RAMP ($C_{dVdT} = \text{OPEN}$): TPS2592Bx

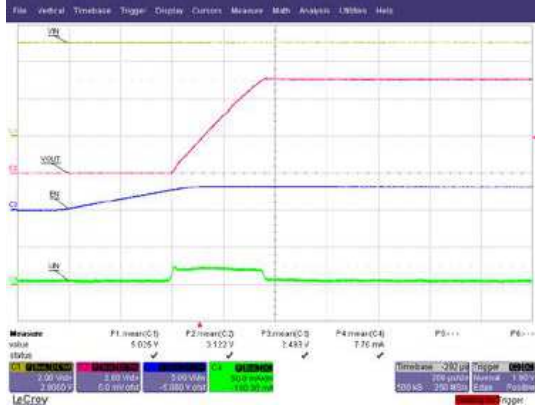


Figure 19.

Transient Output Ramp ($C_{dVdT} = 1\text{ nF}$, $C_{OUT} = 10\text{ }\mu\text{F}$, $R_{OUT} = 2.5\text{ }\Omega$): TPS2592Bx

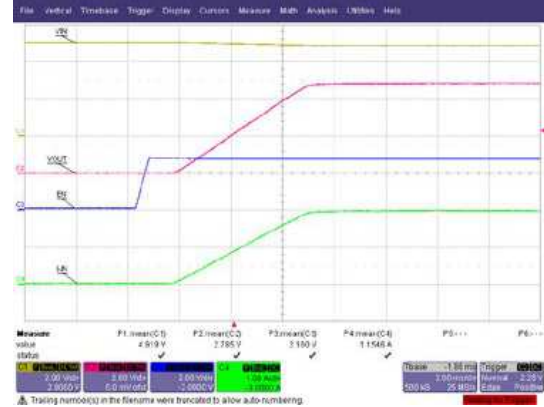


Figure 20.

TRANSIENT: TURN OFF DELAY ($EN\downarrow$)

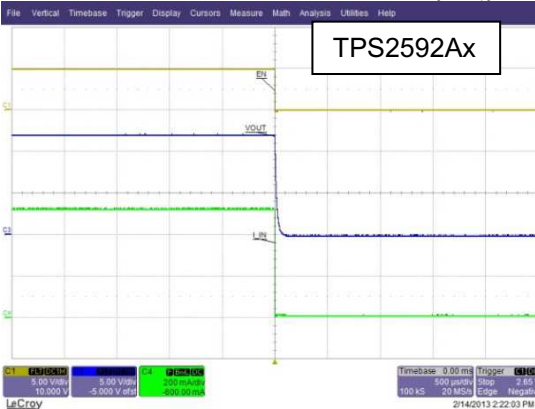


Figure 21.

TURN OFF DELAY TO BFET ($EN\downarrow$)

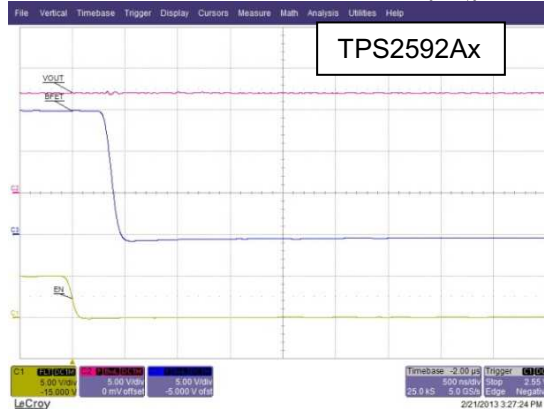


Figure 22.

TURN OFF DELAY TO BFET ($VIN\downarrow$)

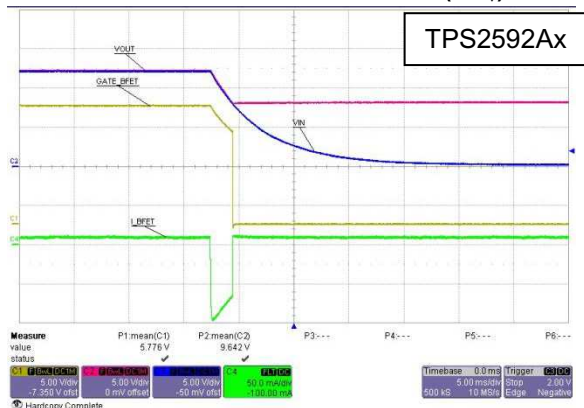


Figure 23.

TRANSIENT: TURN OFF DELAY TO BFET ($VIN\downarrow$)



Figure 24.

TYPICAL CHARACTERISTICS (continued)

$T_J = 25^\circ\text{C}$, $V_{VIN} = 12\text{ V}$ for TPS2592Ax, $V_{VIN} = 5\text{ V}$ for TPS2592Bx, $V_{ENUVLO} = 2\text{ V}$, $R_{ILIM} = 100\text{ k}\Omega$, $C_{VIN} = 0.1\text{ }\mu\text{F}$, $C_{OUT} = 1\text{ }\mu\text{F}$, $C_{dVdT} = \text{OPEN}$ (unless stated otherwise)

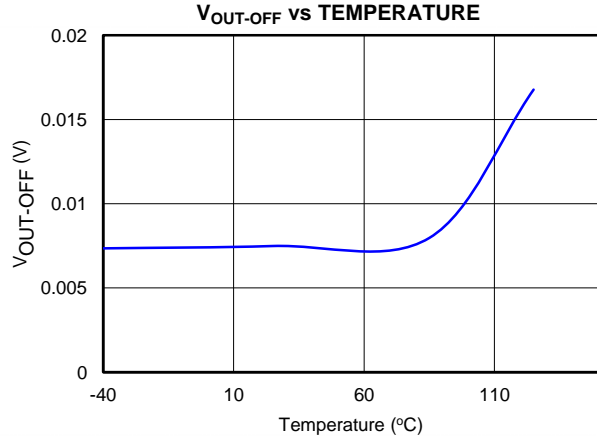


Figure 25.

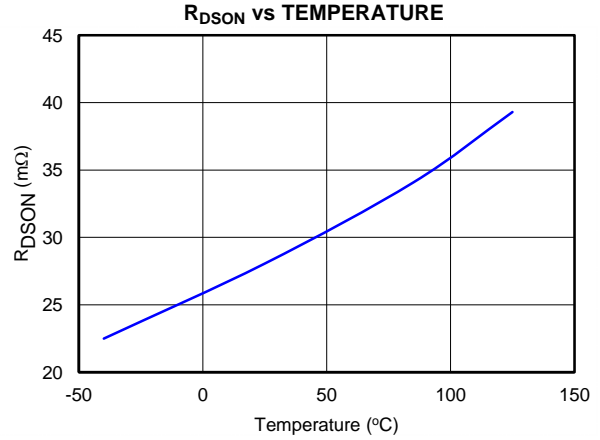


Figure 26.

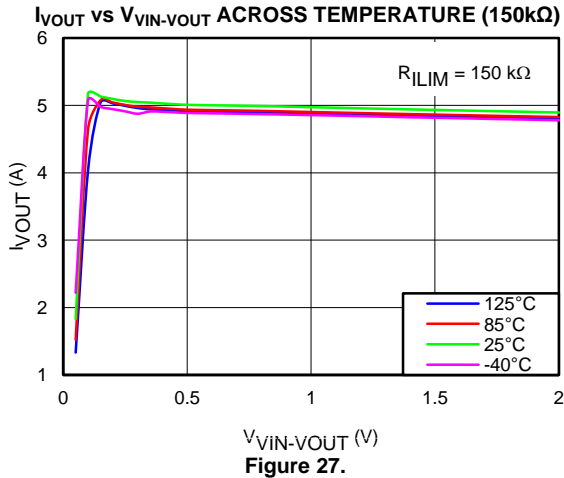


Figure 27.

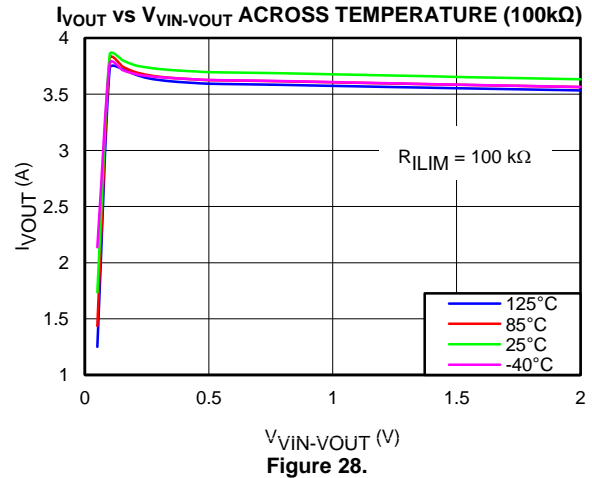


Figure 28.

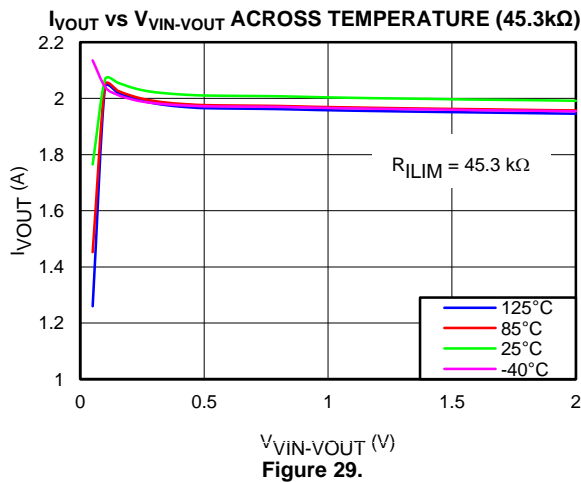


Figure 29.

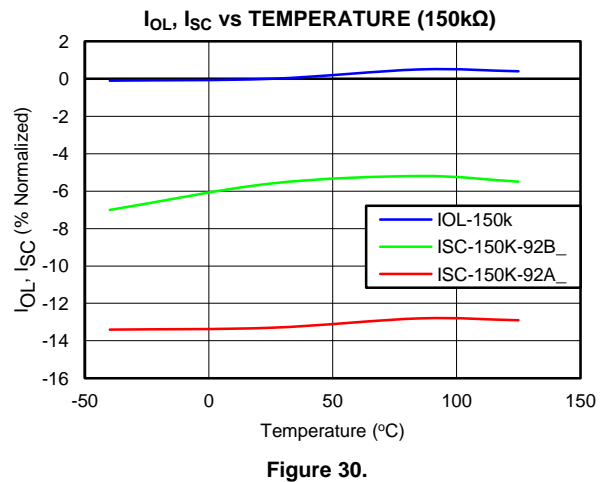


Figure 30.

PRODUCT PREVIEW

TYPICAL CHARACTERISTICS (continued)

$T_J = 25^\circ\text{C}$, $V_{VIN} = 12\text{ V}$ for TPS2592Ax, $V_{VIN} = 5\text{ V}$ for TPS2592Bx, $V_{ENUVLO} = 2\text{ V}$, $R_{ILIM} = 100\text{ k}\Omega$, $C_{VIN}=0.1\ \mu\text{F}$, $C_{OUT}=1\ \mu\text{F}$, $C_{dVdT} = \text{OPEN}$ (unless stated otherwise)

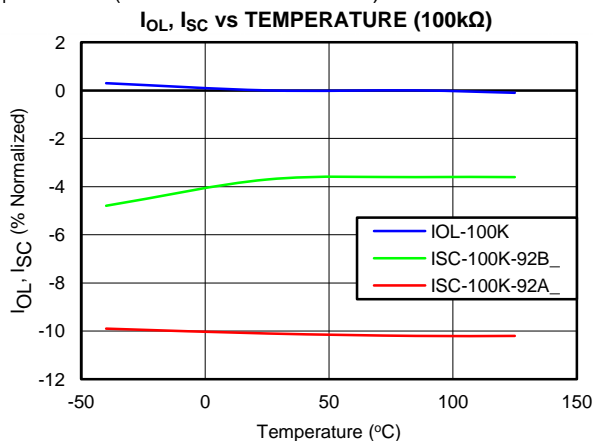


Figure 31.

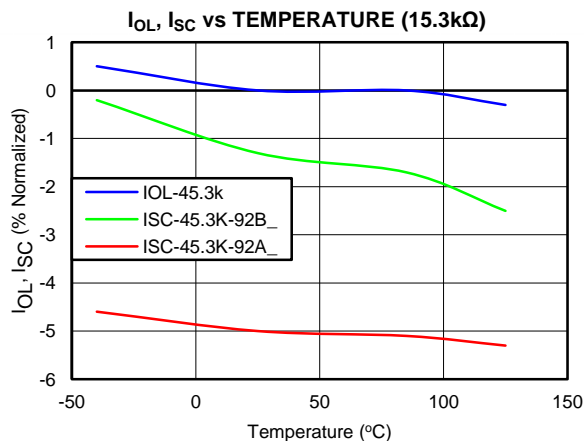


Figure 32.

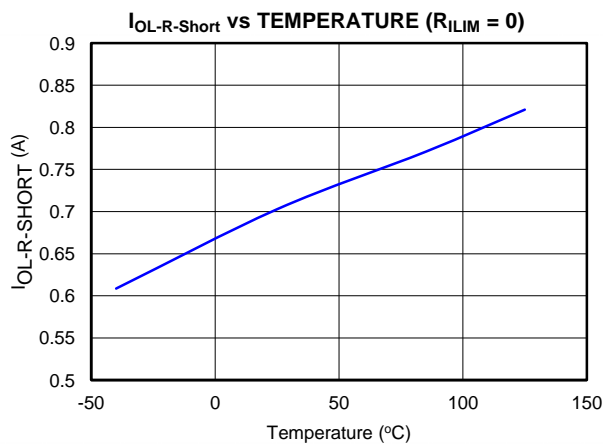


Figure 33.

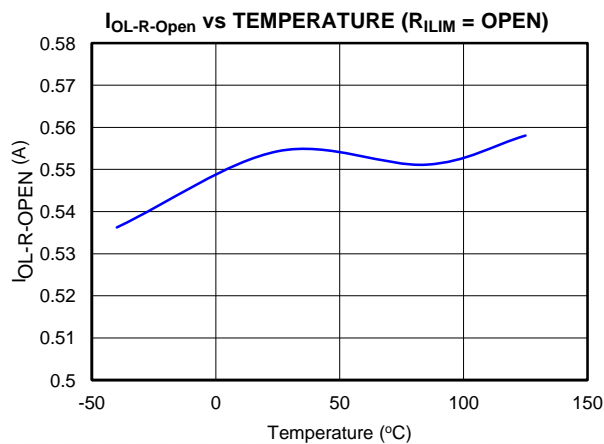


Figure 34.

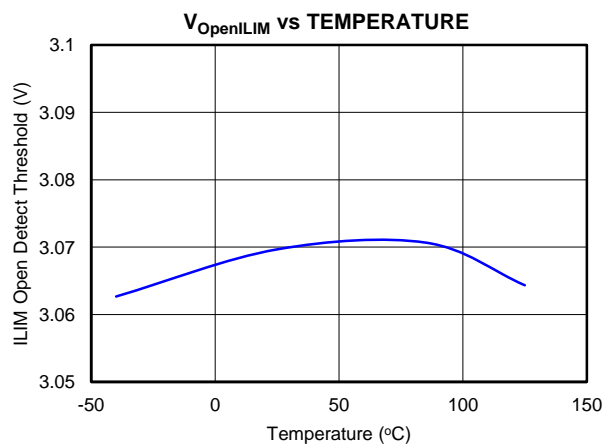


Figure 35.

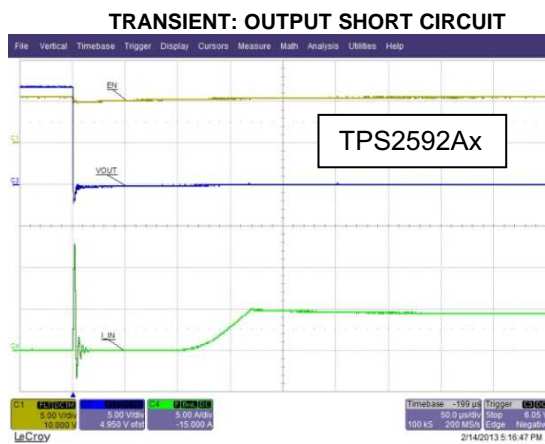


Figure 36.

PRODUCT PREVIEW

TYPICAL CHARACTERISTICS (continued)

$T_J = 25^\circ\text{C}$, $V_{VIN} = 12\text{ V}$ for TPS2592Ax, $V_{VIN} = 5\text{ V}$ for TPS2592Bx, $V_{ENUVLO} = 2\text{ V}$, $R_{LIM} = 100\text{ k}\Omega$, $C_{VIN} = 0.1\text{ }\mu\text{F}$, $C_{OUT} = 1\text{ }\mu\text{F}$, $C_{dVdT} = \text{OPEN}$ (unless stated otherwise)

SHORT CIRCUIT (Zoom): FAST-TRIP COMPARATOR

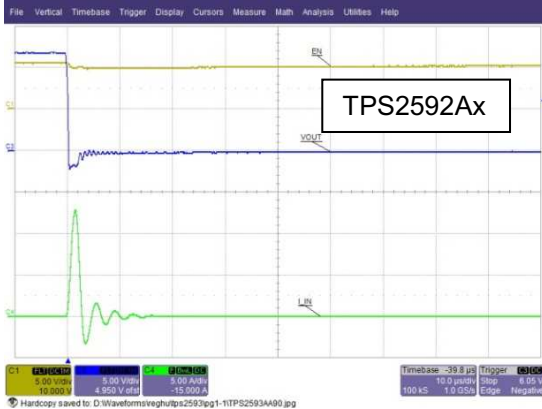


Figure 37.

TRANSIENT: RECOVERY FROM SHORT CIRCUIT

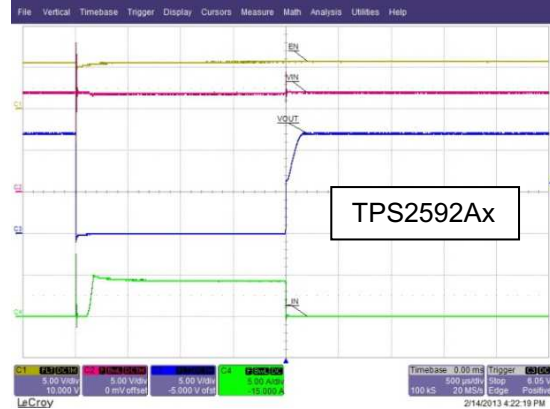


Figure 38.

TRANSIENT: WAKE UP TO SHORT CIRCUIT

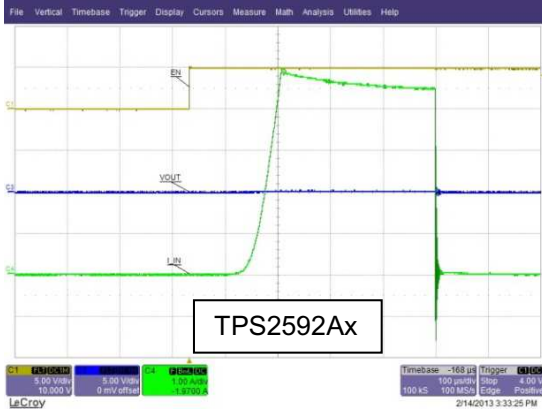


Figure 39.

**TRANSIENT: OVERLOAD CURRENT LIMIT:
(LOAD stepped from 50% to 120%, back to 50%)**

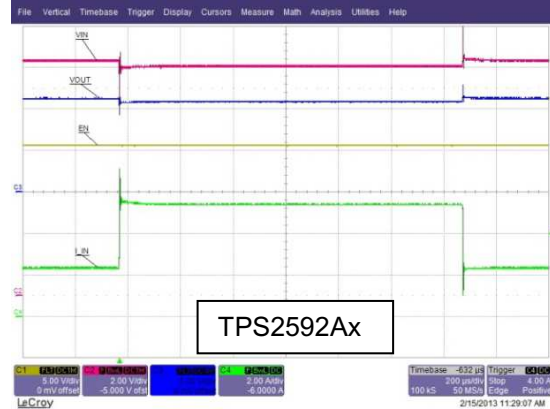


Figure 40.

TRANSIENT: THERMAL FAULT AUTO-RETRY (TPS2592xA)

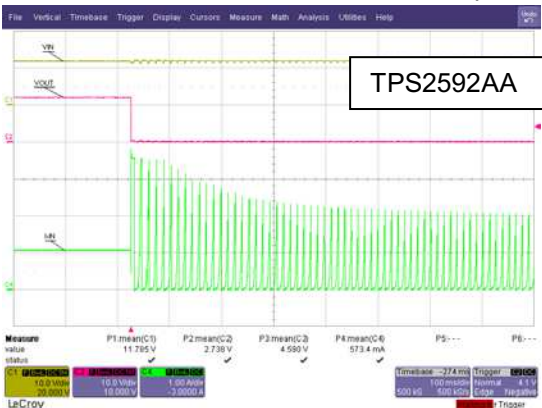


Figure 41.

TRANSIENT: THERMAL FAULT LATCHED (TPS2592xL)

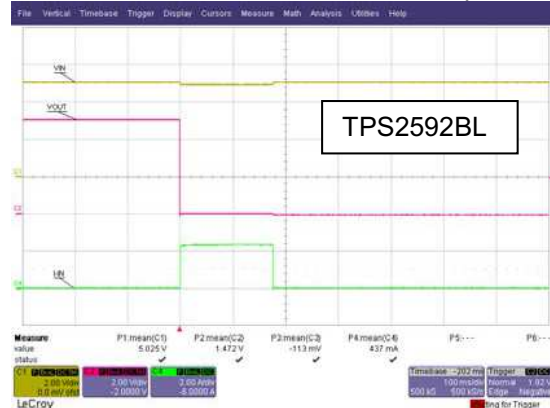


Figure 42.

DEVICE OPERATION

The TPS2592xx is a hot-swap controller with integrated power switch that is used to manage current/voltage/start-up voltage ramp to a connected load within pre-determined level. The device starts its operation by monitoring the VIN bus, when VIN exceeds the undervoltage threshold (VUVLO), the device samples the EN/UVLO pin. A high level on this pin will enable the internal MOSFET and also start charging the external blocking FET (if connected) via the BFET pin. As VIN rises the internal MOSFET of the device and external FET (if connected) will start conducting and allow current to flow from VIN to OUT. If EN/UVLO is held low, i.e. below V_{EN_LOW} the internal MOSFET remains turned off while BFET pin will be held low via the internal discharge resistor hence blocking the flow of current from VIN to OUT. User also has the ability to modify the output voltage ramp time by connecting a capacitor between GATE pin and GND.

Having successfully completed its start-up sequence, the device now actively monitors its load current and input voltage ensuring that the adjustable overload current limit I_{OL} is not exceeded and input voltage spikes are safely clamped to V_{OVC} level at the output to keep the connected device safe from harmful voltage and current transients. The device also has built-in thermal sensor. In the event device temperature (T_J) exceeds T_{SHDN} , typically 160°C, the thermal shutdown circuitry will shut down the internal MOSFET thereby disconnecting the load from the supply. In the TPS2592xL the output will remain disconnected (MOSFET open) until power to device is recycled or EN/UVLO is toggled (pulled low and then high or left open). In TPS2592xA device will remain off during a cooling period until device temperature falls below $T_{SHDN} - 10^\circ\text{C}$ and device will attempt to restart. This ON and OFF cycle will continue until fault is cleared.

DETAILED PIN DESCRIPTION

GND: This is the most negative voltage in the circuit and is used as a reference for all voltage measurements unless otherwise specified.

VIN: Input voltage to the TPS2592xx. The recommended operating voltage range is 4.5V – 13.8V for TPS2592Ax and 4.5V – 5.5V for TPS2592Bx

dVdT: Connect capacitor to control the slew rate of the output voltage at power-on. If no control of the start-up voltage ramp is needed leave this pin floating. Equation governing slew rate at start-up is shown below:

$$I_{dVdT} = (C_{EXT} + C_{INT}) \times \frac{dV_{OUT}}{dT \times GAIN_{dVdT}} \quad (1)$$

Where:

$$I_{dVdT} = 220 \text{ nA (TYP)}$$

$$C_{INT} = 70 \text{ pF (TYP)}$$

$$\frac{dV_{OUT}}{dT} = \text{Desired output slew rate}$$

The total ramp time (T_{dVdT}) for 0 to VIN can be calculated using the following equation:

$$T_{dVdT} = 10^6 \times VIN \times (C_{EXT} + 70 \text{ pF}) \quad (2)$$

BFET: Connect this pin to an external NFET that can be used to disconnect input supply from rest of the system in the event of power failure at VIN. BFET pin is controlled by either UVLO event or EN/UVLO (see table below). BFET can source charging current of 2µA (TYP) and sink (discharge) current from the gate of the external FET via a 5Ω internal discharge resistor to initiate fast turn-off, typically <1 µs.

EN/UVLO	UVLO Event	BFET Mode
H	not-triggered	Charge
L	triggered	Discharge

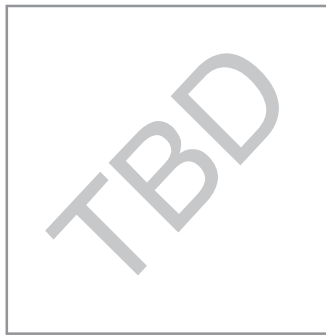
EN/UVLO: As an input pin it controls both the ON/OFF state of the internal MOSFET and that of the external blocking FET. In its high state the internal MOSFET is enabled and charging begins for the external FET. A low on this pin will turn off the internal MOSFET and pull the gate of the external FET to GND via the built-in discharge resistor. High and Low levels are specified in the parametric table of the datasheet. The EN/UVLO pin is also used to clear a thermal shutdown latch in the TPS2592xL by toggling this pin (L→H).

The internal de-glitch delay on EN/UVLO falling edge is intentionally kept low (1us typical) for quick detection of power failure. When used with a resistor divider from supply to EN/UVLO to GND, power-fail detection on EN/UVLO helps in quick turn-off of the BFET driver, thereby stopping the flow of reverse current (see typical application diagram, Figure 44). For applications where a higher de-glitch delay on EN/UVLO is desired, or when the supply is particularly noisy, it is recommended to use an external bypass capacitor from EN/UVLO to GND

ILIM: The device monitors current at both start-up and during normal operation. Current through the MOSFET is limited to value programmed by RILIM resistor that is connected between ILIM and GND. After start-up event and during normal operation, current limit is set to I_{OL} (over-load current limit).

$$I_{OL} = (0.7 + 3 \times 10^{-5} \times R_{ILIM}) \quad (3)$$

When power dissipation in the internal MOSFET [$P_D = V_{VIN-VOUT} \times I_{OUT}$] exceeds 10W, there is a 2% – 12% thermal foldback in the current limit value so that I_{OL} drops to I_{SC} . In each of the two modes, MOSFET gate voltage is increased or decreased to throttle short-circuit and overload current flowing to the load. Eventually, the device shuts down due to over temperature.



During a transient short circuit event, the current through the device increases very rapidly. The current-limit amplifier cannot respond very quickly to this event due to its limited bandwidth. Therefore, the TPS2592 incorporates a fast-trip comparator, which shuts down the pass device very quickly when $I_{OUT} > I_{FASTTRIP}$, and terminates the rapid short-circuit peak current. The trip threshold is set to 50% higher than the programmed over-load current limit ($I_{FASTTRIP} = 1.5 \times I_{OL}$). After the transient short-circuit peak current has been terminated by the fast-trip comparator, the current limit amplifier smoothly regulates the output current to I_{OL} (see figure below).



TPS2592xx TYPICAL APPLICATIONS

Application with Output Ramp-Rate Control

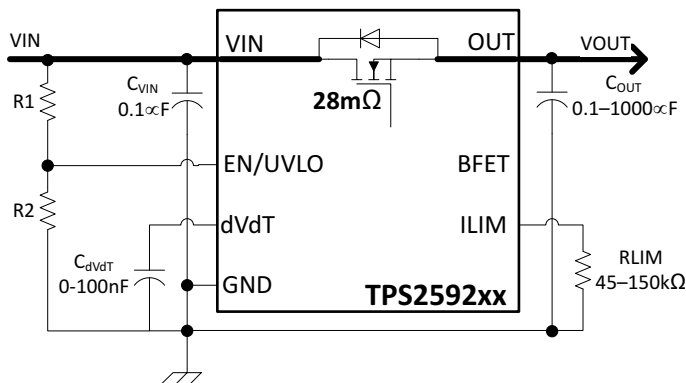


Figure 43. Simple e-Fuse (Current-Limiter)

TPS2592 UVLO is used as PowerFail detection to stop reverse drain of C_{HOLD-UP}.

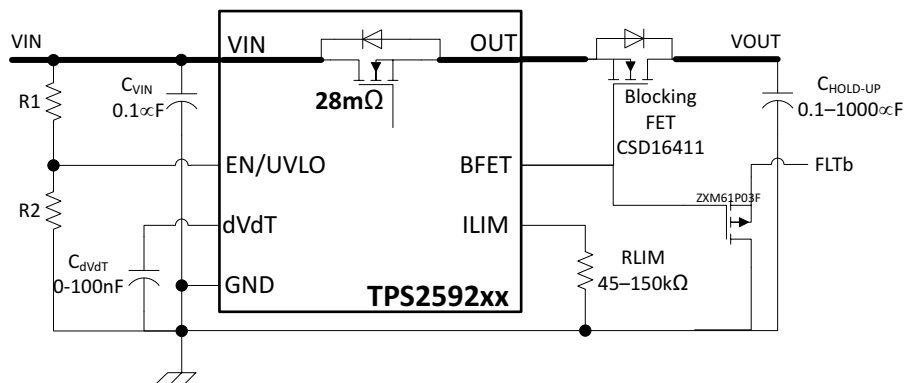


Figure 44. SSD Application with Blocking FET Before C_{HOLD-UP}

TPS2413 is used reverse current comparator.

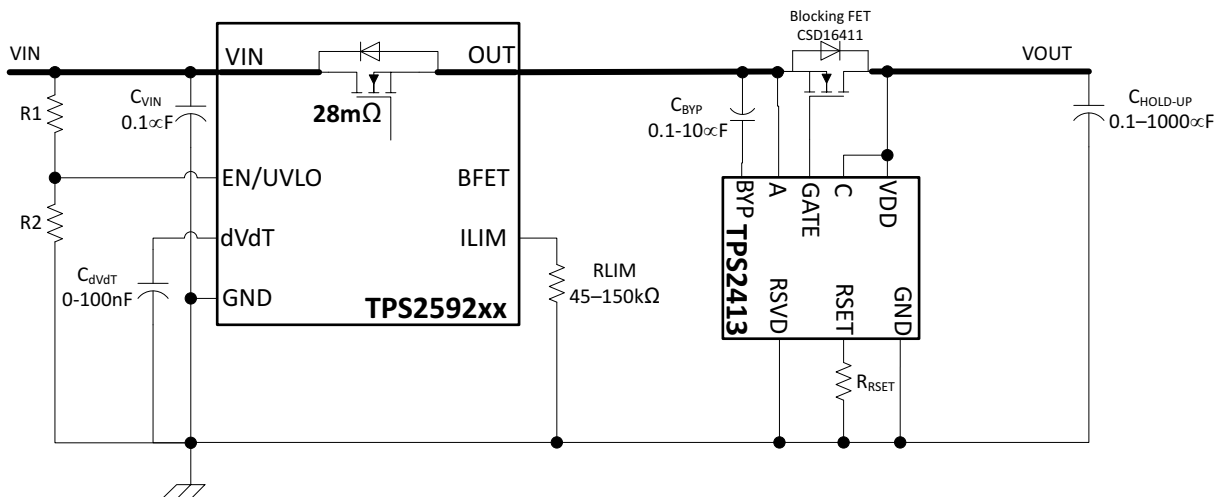


Figure 45. SSD Application with Blocking FET Before C_{HOLD-UP}

PRODUCT PREVIEW

PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead/Ball Finish	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
TPS2592AADRCR	PREVIEW	SON	DRC	10	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 125	2592AA	
TPS2592AADRCT	PREVIEW	SON	DRC	10	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 125	2592AA	
TPS2592BLDRCR	PREVIEW	SON	DRC	10		Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 125	2592BL	
TPS2592BLDRCT	PREVIEW	SON	DRC	10		Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 125	2592BL	

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSELETE: TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check <http://www.ti.com/productcontent> for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

Pb-Free (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

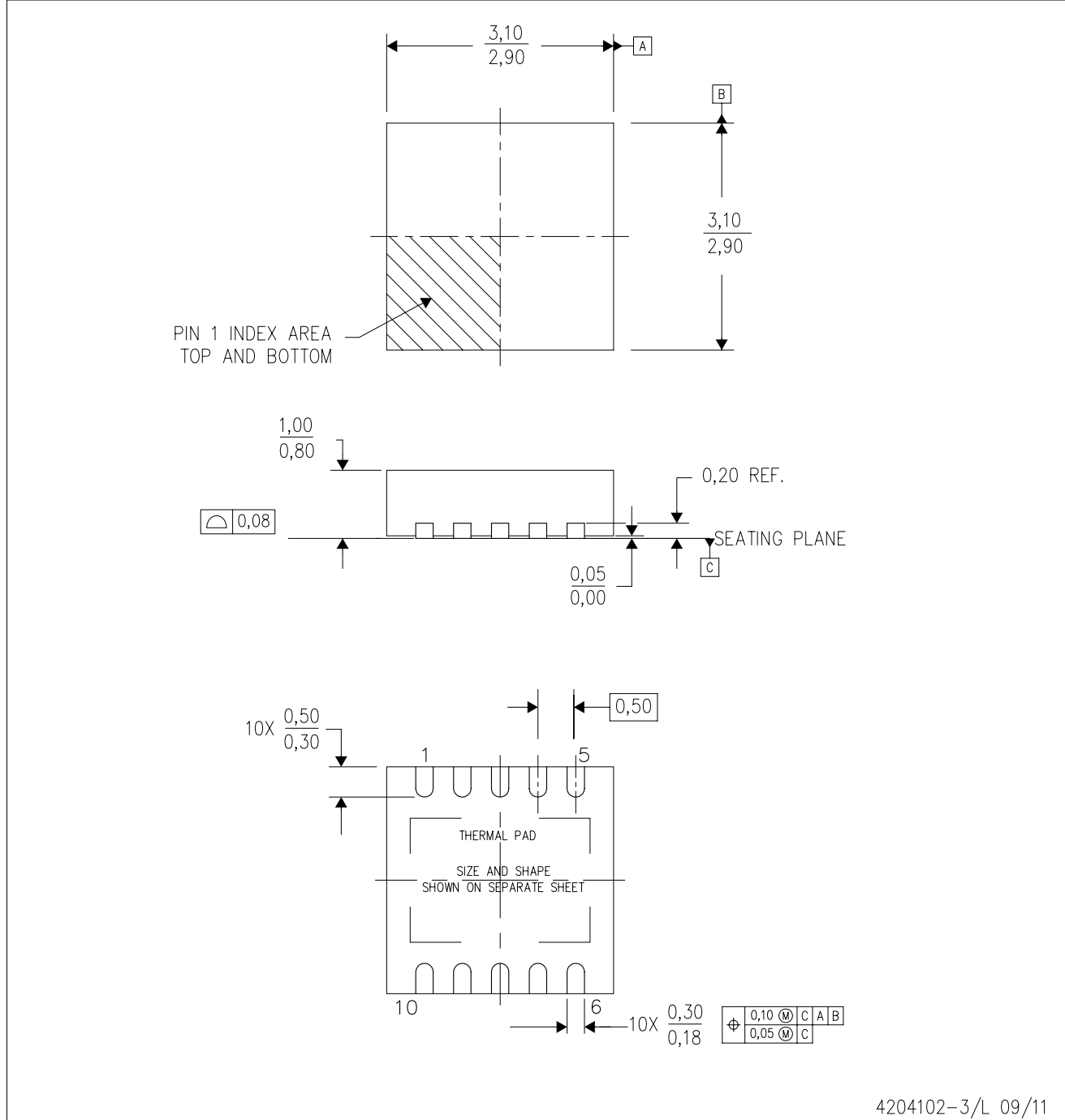
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DRC (S-PVSON-N10)

PLASTIC SMALL OUTLINE NO-LEAD



- NOTES:
- All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M-1994.
 - This drawing is subject to change without notice.
 - Small Outline No-Lead (SON) package configuration.
 - The package thermal pad must be soldered to the board for thermal and mechanical performance, if present.
 - See the additional figure in the Product Data Sheet for details regarding the exposed thermal pad features and dimensions, if present

THERMAL PAD MECHANICAL DATA

DRC (S-PVSON-N10)

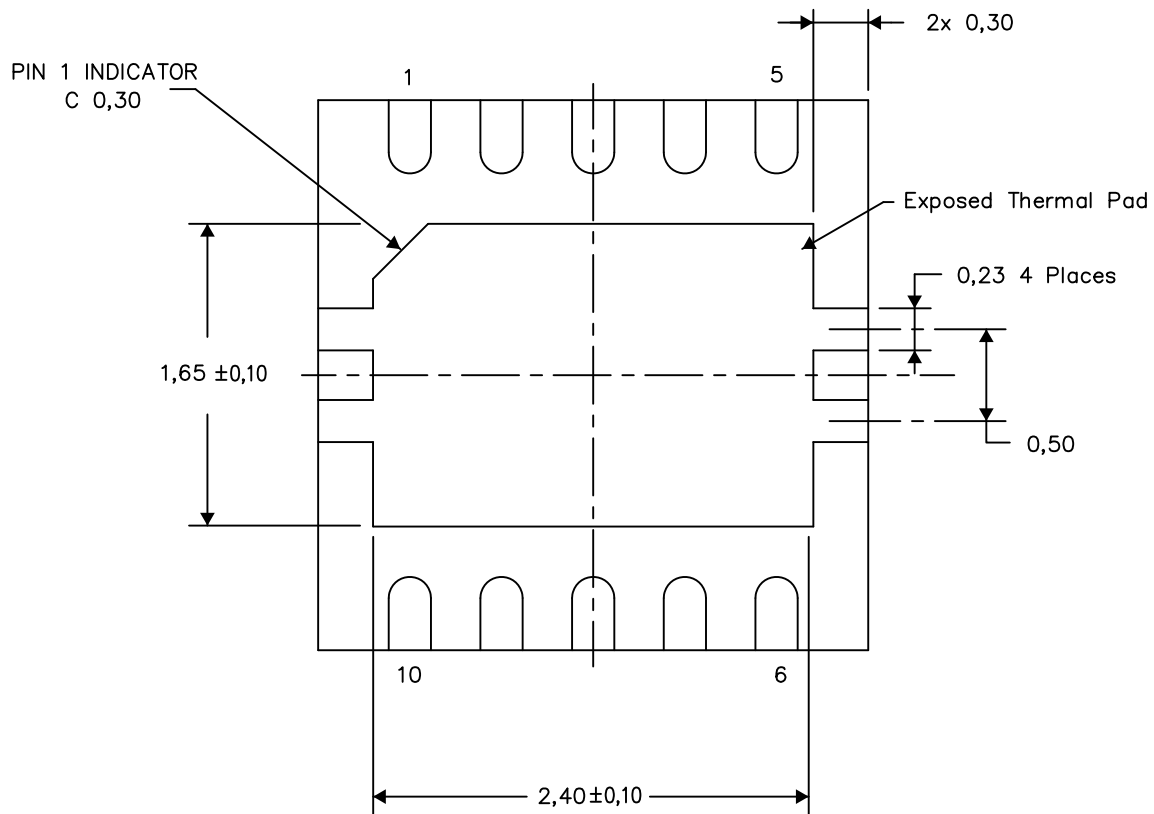
PLASTIC SMALL OUTLINE NO-LEAD

THERMAL INFORMATION

This package incorporates an exposed thermal pad that is designed to be attached directly to an external heatsink. The thermal pad must be soldered directly to the printed circuit board (PCB). After soldering, the PCB can be used as a heatsink. In addition, through the use of thermal vias, the thermal pad can be attached directly to the appropriate copper plane shown in the electrical schematic for the device, or alternatively, can be attached to a special heatsink structure designed into the PCB. This design optimizes the heat transfer from the integrated circuit (IC).

For information on the Quad Flatpack No-Lead (QFN) package and its advantages, refer to Application Report, QFN/SON PCB Attachment, Texas Instruments Literature No. SLUA271. This document is available at www.ti.com.

The exposed thermal pad dimensions for this package are shown in the following illustration.



Bottom View

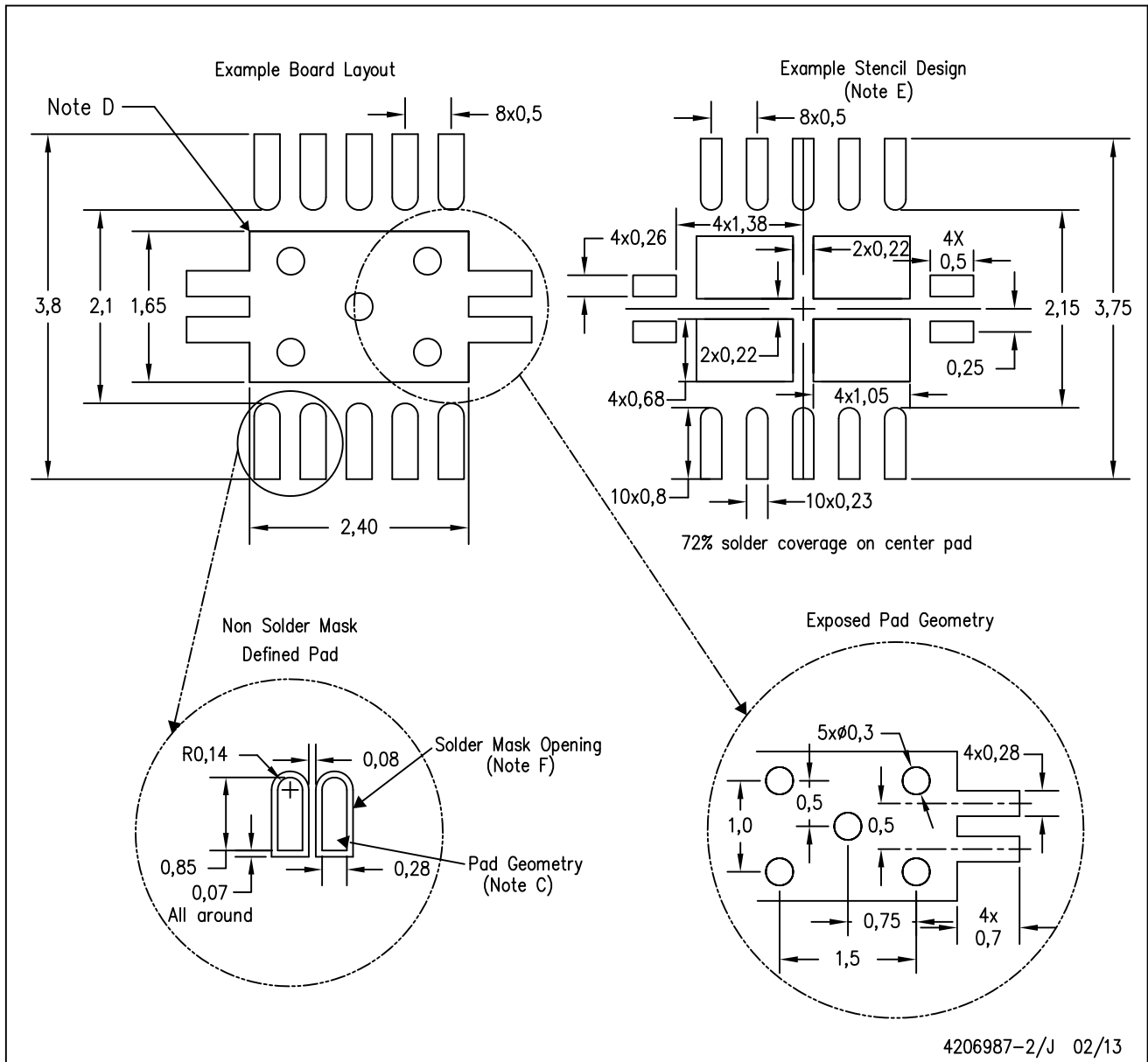
Exposed Thermal Pad Dimensions

4206565-3/R 03/13

NOTE: A. All linear dimensions are in millimeters

DRC (S-PVSON-N10)

PLASTIC SMALL OUTLINE NO-LEAD



- NOTES:
- All linear dimensions are in millimeters.
 - This drawing is subject to change without notice.
 - Publication IPC-7351 is recommended for alternate designs.
 - This package is designed to be soldered to a thermal pad on the board. Refer to Application Note, Quad Flat-Pack Packages, Texas Instruments Literature No. SLUA271, and also the Product Data Sheets for specific thermal information, via requirements, and recommended board layout. These documents are available at www.ti.com <<http://www.ti.com>>.
 - Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC 7525 for stencil design considerations.
 - Customers should contact their board fabrication site for minimum solder mask web tolerances between signal pads.

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