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LMH2180 75 MHz Dual Clock Buffer

Check for Samples: LMH2180

FEATURES

- (Typical Values are: V_{SUPPLY} = 2.7V and C_L = 10 pF, unless Otherwise Specified.)
- Small Signal Bandwidth 78 MHz
- Supply Voltage Range 2.4V to 5V
- Phase Noise (V_{IN} = 1 V_{PP}, f_C = 38.4 MHz, Δf = 1 kHz) -123 dBc/Hz
- Slew Rate 106 V/µs
- Total Supply Current 2.3 mA
- Shutdown Current 30 µA
- Rail-to-Rail Input and Output
- Individual Buffer Enable Pins
- Rapid T_{on} Technology
- Crosstalk Rejection Circuitry
- Packages:
 - 8-Pin WSON, Solder Bump and no Pullback
 - 8-Bump DSBGA
- Temperature Range -40°C to 85°C

APPLICATIONS

- 3G Mobile Applications
- WLAN–WiMAX Modules
- TD_SCDMA Multi-Mode MP3 and Camera
- GSM Modules
- Oscillator Modules

DESCRIPTION

The LMH2180 is a high speed dual clock buffer designed for portable communications and applications requiring multiple accurate multi-clock systems. The LMH2180 integrates two 75 MHz low noise buffers with independent shutdown pins into a small package. The LMH2180 ensures superb system operation between the baseband and the oscillator signal path by eliminating crosstalk between the multiple clock signals.

Unique technology and design provides the LMH2180 with the ability to accurately drive both large capacitive and resistive loads. Low supply current combined with shutdown pins for each channel means the LMH2180 is ideal for battery powered applications. This part does not use an internal ground reference, thus providing additional system flexibility.

The flexible buffers provide system designers the capacity to manage complex clock signals in the latest wireless applications. Each buffer delivers 106 V/ μ s internal slew rate with independent shutdown and duty cycle precision. Each input is internally biased to 1V, removing the need for external resistors. Both channels have rail-to-rail inputs and outputs, a gain of one, and are AC coupled with the use of one capacitor.

Replacing a discrete buffer solution with the LMH2180 provides many benefits: simplified board layout, minimized parasitic components, simplified BOM, design durability across multiple applications, simplification of clock paths, and the ability to reduce the number of clock signal generators in the system. The LMH2180 is produced in the tiny packages minimizing the required PCB space.

TYPICAL APPLICATION



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CONNECTION DIAGRAMS



Figure 1. 8-Pin WSON Package See Package Number NGW0008A or NGQ0008A



Figure 2. 8-Bump DSBGA Package See Package Number YFQ0008AAA

PIN DESCRIPTIONS

Pin No. WSON	Pin No. DSBGA	Pin Name	Description
1	A3	V _{DD}	Voltage supply connection
2	A2	IN 1	Input 1
3	A1	IN 2	Input 2
4	B1	ENABLE 2	Enable buffer 2
5	C1	V _{SS}	Ground connection
6	C2	OUT 2	Output 2
7	C3	OUT 1	Output 1
8	B3	ENABLE 1	Enable buffer 1



These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

ABSOLUTE MAXIMUM RATINGS (1)(2)

Supply Voltages (V ⁺ – V ⁻)	5.5V
ESD Tolerance	
Human Body ⁽³⁾	2000V
Machine Model ⁽⁴⁾	200V
Charged Device Model	1000V
Storage Temperature Range	−65°C to 150°C
Junction Temperature ⁽⁵⁾	150°C
Soldering Information	
Infrared or Convection (35 sec.)	235°C

(1) "Absolute Maximum Ratings" indicate limits beyond which damage to the device may occur, including inoperability and degradation of the device reliability and/or performance. Functional operation of the device and/or non-degradation at the Absolute Maximum Ratings or other conditions beyond those indicated in the Recommended Operating Conditions is not implied. The Recommended Operating Conditions indicate conditions at which the device is functional and the device should not be operated beyond such conditions. All voltages are measured with respect to the ground pin, unless otherwise specified.

- (2) If Military/Aerospace specified devices are required, please contact the Texas Instruments Sales Office/ Distributors for availability and specifications.
- (3) Human body model, applicable std. JESD22–A114C.
- (4) Machine model, applicable std. JESD22-A115-A.
- (5) The maximum power dissipation must be derated at elevated temperatures and is dictated by T_{J(MAX)}, θ_{JA}, and the ambient temperature T_A. The maximum allowable power dissipation is P_{DMAX} = (T_{JMAX} T_A) / θ_{JA} or the number given in the *Absolute Maximum Ratings*, whichever is lower.

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OPERATING RATINGS ⁽¹⁾

Supply Voltage $(V^+ - V^-)$	2.4V to 5.0V
Temperature Range (2)(3)	−40°C to 85°C
Package Thermal Resistance (2)(3)	
8-Pin WSON (θ _{JA})	217°C/W
8-Bump DSBGA (θ _{JA})	90°C/W

(1) "Absolute Maximum Ratings" indicate limits beyond which damage to the device may occur, including inoperability and degradation of the device reliability and/or performance. Functional operation of the device and/or non-degradation at the Absolute Maximum Ratings or other conditions beyond those indicated in the Recommended Operating Conditions is not implied. The Recommended Operating Conditions indicate conditions at which the device is functional and the device should not be operated beyond such conditions. All voltages are measured with respect to the ground pin, unless otherwise specified.

(2) The Electrical Characteristics tables show performance under the specified Recommended Operating Conditions except as otherwise modified by the Electrical Characteristics Conditions and/or Notes. Typical values represent typical performance as measured by production tests. Individual parts may vary.

(3) The maximum power dissipation must be derated at elevated temperatures and is dictated by T_{J(MAX)}, θ_{JA}, and the ambient temperature T_A. The maximum allowable power dissipation is P_{DMAX} = (T_{JMAX} - T_A) / θ_{JA} or the number given in the *Absolute Maximum Ratings*, whichever is lower.

2.7V ELECTRICAL CHARACTERISTICS

Unless otherwise specified, all limits are specified for $T_A = 25^{\circ}$ C, $V_{DD} = 2.7$ V, $V_{SS} = 0$ V, $V_{CM} = 1$ V, Enable_{1,2} = V_{DD} , $C_L = 10$ pF, $R_L = 30 \text{ k}\Omega$, Load is connected to V_{SS} , $C_{COUPLING} = 10$ nF. **Boldface** limits apply at temperature range extremes of operating condition.⁽¹⁾

Parameter		Test Conditions	Min ⁽²⁾	Тур ⁽³⁾	Max ⁽²⁾	Units
Frequency	Domain Response					
SSBW	Small Signal Bandwidth	V _{IN} = 100 mV _{PP} ; −3 dB		78		MHz
LSBW	Large Signal Bandwidth	V _{IN} = 1.0 V _{PP} ; −3 dB		60		MHz
GFN	Gain Flatness < 0.1 dB	f > 100 kHz		4.9		MHz
Distortion	and Noise Performance					
φ _n	Phase Noise	V_{IN} = 1 V_{PP} , f_C = 38.4 MHz, Δf = 1 kHz		-123		dBc/Hz
		V_{IN} = 1 $V_{\text{PP}},f_{\text{C}}$ = 38.4 MHz, Δf = 10 kHz		-132		dBc/Hz
e _n	Input-Referred Voltage Noise	$f = 1 \text{ MHz}, \text{ R}_{\text{SOURCE}} = 50\Omega$		13		nV/√Hz
I _{SOLATION}	Output to Input	$f = 1 \text{ MHz}, \text{ R}_{\text{SOURCE}} = 50\Omega$		84		dB
СТ	Crosstalk Rejection	f = 38.4 MHz, V _{IN} = 1 V _{PP}		41		dB
Time Doma	ain Response					
t _r	Rise Time	0.1 V _{PP} Step (10-90%)		6		ns
t _f	Fall Time			5		ns
t _s	Settling Time to 0.1%	1 V _{PP} Step		120		ns
OS	Overshoot	0.1 V _{PP} Step		37		%
SR	Slew Rate ⁽⁴⁾	$V_{IN} = 2 V_{PP}$		106		V/µs
Static DC F	Performance					
I _S	Supply Current	$Enable_{1,2} = V_{DD}$; No Load		2.3	2.7 2.9	mA
		$Enable_1 = V_{DD}$, $Enable_2 = V_{SS}$, No Load		1.3	1.5 1.6	mA
		$Enable_{1,2} = V_{SS}$; No Load		30	41 46	μΑ
PSRR	Power Supply Rejection Ratio	DC (3.0V to 5.0V)	65 64	68		dB

(1) The Electrical Characteristics tables show performance under the specified Recommended Operating Conditions except as otherwise modified by the Electrical Characteristics Conditions and/or Notes. Typical values represent typical performance as measured by production tests. Individual parts may vary.

(2) Datasheet min/max limits are specified by test or statistical analysis.

(3) Typical values represent the most likely parametric norms at $T_A = +25$ °C, and at the Recommended Operation Conditions at the time of product characterization.

(4) Slew rate is the average of the rising and falling slew rates.

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2.7V ELECTRICAL CHARACTERISTICS (continued)

Unless otherwise specified, all limits are specified for $T_A = 25^{\circ}C$, $V_{DD} = 2.7V$, $V_{SS} = 0V$, $V_{CM} = 1V$, Enable_{1,2} = V_{DD} , $C_L = 10$ pF, $R_L = 30 \text{ k}\Omega$, Load is connected to V_{SS} , $C_{COUPLING} = 10$ nF. **Boldface** limits apply at temperature range extremes of operating condition.⁽¹⁾

	Parameter	Test Conditions	Min ⁽²⁾	Тур ⁽³⁾	Max ⁽²⁾	Units	
A _{CL}	Small Signal Voltage Gain	$V_{IN} = 0.2 V_{PP}$	0.95	1.0	1.05	V/V	
V _{OS}	Output Offset Voltage			-0.5	17 18	mV	
TC V _{OS}	Temperature Coefficient Output Offset Voltage ⁽⁵⁾			2.8		µV/°C	
R _{OUT}	Output Resistance	f = 100 kHz		0.6			
		f = 38.4 MHz		166		Ω	
		disabled	Hi	gh Impedar	ice		
Miscellan	eous Performance						
R _{IN}	Input Resistance per Buffer	Enable = V _{DD}		137			
		Enable = V _{SS}		137		KΩ	
C _{IN}	Input Capacitance per Buffer	Enable = V _{DD}		1.3		. E	
		Enable = V _{SS}		1.3		р⊢	
Z _{IN}	Input Impedance	f = 38.4 MHz, Enable = V_{DD}		4.5		- 0	
		f = 38.4 MHz, Enable = V _{SS}		4.2		KΩ	
Vo	Output Swing Positive	V _{IN} = V _{DD}	2.66 2.65	2.69		V	
	Output Swing Negative	V _{IN} = V _{SS}		19	35 37	mV	
I _{SC}	Output Short-Circuit Current ⁽⁶⁾⁽⁷⁾	Sourcing, $V_{IN} = V_{DD}$, $V_{OUT} = V_{SS}$	-21 - 18	-25		•	
		Sinking, $V_{IN} = V_{SS}$, $V_{OUT} = V_{DD}$	23 15	25		mA	
V _{en_hmin}	Enable High Active Minimum Voltage			1.2		N/	
V _{en_Imax}	Enable Low Inactive Maximum Voltage			0.6		v	

(5) Average Temperature Coefficient is determined by dividing the changing in a parameter at temperature extremes by the total temperature change.

(6) Short-Circuit test is a momentary test. Continuous short circuit operation at elevated ambient temperature can result in exceeding the maximum allowed junction temperature of 150°C.

(7) Positive current corresponds to current flowing into the device.

5V ELECTRICAL CHARACTERISTICS

Unless otherwise specified, all limits are specified for $T_A = 25^{\circ}C$, $V_{DD} = 5V$, $V_{SS} = 0V$, $V_{CM} = 1V$, $Enable_{1,2} = V_{DD}$, $C_L = 10 \text{ pF}$, $R_L = 30 \text{ k}\Omega$, Load is connected to V_{SS} , $C_{COUPLING} = 10 \text{ nF}$. **Boldface** limits apply at temperature range extremes of operating condition.⁽¹⁾

	Parameter	Test Conditions	Min ⁽²⁾	Тур ⁽³⁾	Max ⁽²⁾	Units
Frequency	Domain Response					
SSBW	Small Signal Bandwidth	V_{IN} = 100 m V_{PP} ; -3 dB		87		MHz
LSBW	Large Signal Bandwidth	V _{IN} = 1.0 V _{PP} ; −3 dB		68		MHz
GFN	Gain Flatness < 0.1 dB	f > 100 kHz		25		MHz
Distortion	and Noise Performance					
φ _n	Phase Noise	V_{IN} = 1 V_{PP} , f _C = 38.4 MHz, Δ f = 1 kHz		-123		dBc/Hz
		V_{IN} = 1 $V_{\text{PP}},f_{\text{C}}$ = 38.4 MHz, Δf = 10 kHz		-132		dBc/Hz

(1) The Electrical Characteristics tables show performance under the specified Recommended Operating Conditions except as otherwise modified by the Electrical Characteristics Conditions and/or Notes. Typical values represent typical performance as measured by production tests. Individual parts may vary.

(2) Datasheet min/max limits are specified by test or statistical analysis.

(3) Typical values represent the most likely parametric norms at $T_A = +25$ °C, and at the Recommended Operation Conditions at the time of product characterization.





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5V ELECTRICAL CHARACTERISTICS (continued)

Unless otherwise specified, all limits are specified for $T_A = 25^{\circ}C$, $V_{DD} = 5V$, $V_{SS} = 0V$, $V_{CM} = 1V$, Enable_{1,2} = V_{DD} , $C_L = 10$ pF, $R_L = 30$ kΩ, Load is connected to V_{SS} , $C_{COUPLING} = 10$ nF. **Boldface** limits apply at temperature range extremes of operating condition.⁽¹⁾

	Parameter	Test Conditions	Min ⁽²⁾	Тур ⁽³⁾	Max ⁽²⁾	Units
e _n	Input-Referred Voltage Noise	$f = 1 MHz, R_{SOURCE} = 50\Omega$		12		nV/√Hz
I SOLATION	Output to Input	$f = 1 MHz, R_{SOURCE} = 50\Omega$		84		dB
СТ	Crosstalk Rejection	f = 38.4 MHz, P _{IN} = 0 dBm		59		dB
Time Doma	ain Response	1	-			
t _r	Rise Time	0.1 V _{PP} Step (10-90%)		6		ns
t _f	Fall Time			6		ns
t _s	Settling Time to 0.1%	1 V _{PP} Step		70		ns
OS	Overshoot	0.1V _{PP} Step		13		%
SR	Slew Rate ⁽⁴⁾	V _{IN} = 2 V _{PP}		124		V/µs
Static DC F	Performance	-	1			1
I _S	Supply Current	$Enable_{1,2} = V_{DD}$; No Load		3.4	4.0 4.1	mA
		$\begin{array}{l} {\sf Enable_1} = {\sf V}_{{\sf DD}}, {\sf Enable_2} = {\sf V}_{{\sf SS}} ; {\sf No} \\ {\sf Load} \end{array}$		1.8	2.2 2.3	mA
		$Enable_{1,2} = V_{SS}$; No Load		32	43 49	μΑ
PSRR	Power Supply Rejection Ratio	DC (3.0V to 5.0V)	65 64	68		dB
A _{CL}	Small Signal Voltage Gain	V _{IN} = 0.2 V _{PP}	0.95	1.0	1.05	V/V
V _{OS}	Output Offset Voltage			-1.4	21 22	mV
TC V _{OS}	Temperature Coefficient Output Offset Voltage ⁽⁵⁾			2.4		µV/°C
R _{OUT}	Output Resistance	f = 100 kHz		0.5		
		f = 38.4 MHz		126		Ω
		disabled	Hi	gh Impedar	nce	
Miscellane	ous Performance					
R _{IN}	Input Resistance per Buffer	Enable = V_{DD}		138		k0
		Enable = V_{SS}		138		K12
C _{IN}	Input Capacitance per Buffer	Enable = V_{DD}		1.3		~ F
		Enable = V_{SS}		1.3		рг
Z _{IN}	Input Impedance	f = 38.4 MHz, Enable = V_{DD}		4.3		ĿО
		f = 38.4 MHz, Enable = V _{SS}		4.2		K12
Vo	Output Swing Positive	V _{IN} = V _{DD}	4.96 4.95	4.99		V
	Output Swing Negative	V _{IN} = V _{SS}		10	35 50	mV
I _{SC}	Output Short-Circuit Current ⁽⁶⁾⁽⁷⁾	Sourcing, $V_{IN} = V_{DD}$, $V_{OUT} = V_{SS}$	-80 -62	-90		~ ^
		Sinking, $V_{IN} = V_{SS}$, $V_{OUT} = V_{DD}$	60 43	65		mA
V _{en_hmin}	Enable High Active Minimum Voltage			1.2		V
V _{en_Imax}	Enable Low Inactive Maximum Voltage			0.6		v

(4) Slew rate is the average of the rising and falling slew rates.

Average Temperature Coefficient is determined by dividing the changing in a parameter at temperature extremes by the total (5) temperature change.

Short-Circuit test is a momentary test. Continuous short circuit operation at elevated ambient temperature can result in exceeding the (6) maximum allowed junction temperature of 150°C.

Positive current corresponds to current flowing into the device. (7)

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TYPICAL PERFORMANCE CHARACTERISTICS

 $T_A = 25^{\circ}C$, $V_{DD} = 2.7V$, $V_{SS} = 0V$, $Enable_{1,2} = V_{DD}$, $C_L = 10 \text{ pF}$, $R_L = 30 \text{ k}\Omega$ and $C_{COUPLING} = 10 \text{ nF}$, unless otherwise specified.



Frequency Response Over Temperature











Frequency Response Over Temperature





EXAS NSTRUMENTS

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TYPICAL PERFORMANCE CHARACTERISTICS (continued) $T_A = 25^{\circ}C$, $V_{DD} = 2.7V$, $V_{SS} = 0V$, $Enable_{1,2} = V_{DD}$, $C_L = 10 \text{ pF}$, $R_L = 30 \text{ k}\Omega$ and $C_{COUPLING} = 10 \text{ nF}$, unless otherwise specified. Input Impedance vs. Frequency 150 _l 2.7V 120 IMPEDANCE (k0) 5V 90 60 30 0 10k 100k 1M 10M 100M FREQUENCY (Hz) Figure 27. VOUT vs. I_{OUT} (Sourcing) 1.5 -40℃ 4 1.0 0.5 Vout (V) 0.0 85°C -0.5 Open Input $\dot{V}_{S} = 5\dot{V}$ -1.0 -30 -80 -70 -60 -50 -40 -30 -20 -10 0 IOUT (mA) Figure 29. V_{OUT} vs. I_{OUT} (Sinking) 3.0 Open Input $V_{\rm S} = 5V_{\rm S}$ 2.5 2.0 85°C Vout (V) 1.5 1.0 40°C 0.5 0.0 L 0 10 20 30 40 50 60 IOUT (mA)













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TYPICAL PERFORMANCE CHARACTERISTICS (continued)

 $T_{A} = 25^{\circ}C, V_{DD} = 2.7V, V_{SS} = 0V, Enable_{1,2} = V_{DD}, C_{L} = 10 \text{ pF}, R_{L} = 30 \text{ k}\Omega \text{ and } C_{COUPLING} = 10 \text{ nF}, \text{ unless otherwise specified}.$









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APPLICATION INFORMATION

GENERAL

The LMH2180 is designed to minimize the effects of spurious signals from the base chip to the oscillator. Also the influence of varying load resistance and capacitance to the oscillator is minimized, while the drive capability is increased.

The inputs of the LMH2180 are internally biased at 1V, making AC coupling possible without external bias resistors.

To optimize current consumption, a buffer that is not in use can be disabled by connecting it's enable pin to V_{SS}.

The LMH2180 has no internal ground reference; therefore, either single or split supply configurations can be used.

The LMH2180 is an easy replacement for discrete circuitry. It simplifies board layout and minimizes the effect of layout related parasitic components.

INPUT CONFIGURATION

The internal 1V input biasing allows AC coupling of the input signal. This biasing avoids the use of external resistors, as depicted in Figure 36. The biasing prevents a large DC load at the oscillators output that creates a load impedance and may affect it's oscillating frequency. As a result of this biasing, the maximum amplitude of the AC signal is $2V_{PP}$.

The coupling capacitance C1 should be large enough to let the AC signal pass. This is a unity gain buffer with rail-to-rail inputs and outputs.



Figure 36. Input Configuration

FREQUENCY PULLING

Frequency pulling is the frequency variation of an oscillator caused by a varying load. In the typical application, the load of the oscillator is a fixed capacitor (C1) in series with the input impedance of the buffer.

To keep the input impedance as constant as possible, the input is biased at 1V, even when the part is disabled. A simplified schematic of the input configuration is shown in Figure 36.

ISOLATION AND CROSSTALK

Output to input isolation prevents the clock signal of the oscillator from being affected by spurious signals generated by the digital blocks behind the output buffer. See Figure 13.

A block diagram of the isolation is shown in Figure 37. Crosstalk rejection between buffers prevents signals from affecting each other. Figure 37 shows a Baseband IC and a Bluetooth module as an example. See Figure 14 for more information.

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Figure 37. Isolation Block Diagram

DRIVING CAPACITIVE LOADS

Each buffer can drive a capacitive load. Be aware that every capacitor directly connected to the output becomes part of the loop of the buffer. In most applications the load consists of the capacitance of copper tracks and the input capacitance of the application blocks. Capacitance reduces the gain/phase margin and decreases the stability. This leads to peaking in the frequency response and in extreme situations oscillations can occur. To drive a large capacitive load it is recommended to include a series resistor between the buffer and the load capacitor. The best value for this isolation resistance can be found by experimentation.

The LMH2180 datasheet reflects measurements with capacitive loads of 10 pF at the output of the buffers. Most common applications will probably use a lower capacitive load, which will result in lower peaking and significantly greater bandwidth, see Figure 38.



Figure 38. Bandwidth and Peaking

PHASE NOISE

A clock buffer adds noise to the clock signal. This noise causes uncertainty in the phase of the clock signal. This uncertainty is described by jitter (time domain) or phase noise (frequency domain). Communication systems, such as Wireless LAN, require a low jitter/phase noise clock signal to obtain a low Bit Error Rate. Figure 39 shows the frequency domain representation of a clock signal with frequency f_c . Without Phase Noise the entire signal power would only be located at the frequency f_c . Phase Noise spreads some of the power to adjacent frequencies. Phase Noise is usually specified in dBc/Hz at a given frequency offset Δf from the carrier, where dBc is the power level in dB relative to the carrier. The noise power is measured within a 1 Hz bandwidth.





Figure 39. Phase Noise

Figure 40 shows the setup used to measure the LMH2180 phase noise. The clock driving the LMH2180 is a state of the art 38.4MHz TCXO. Both the TCXO phase noise and the phase noise at the LMH2180 output were measured. At offset frequencies of 1 kHz and higher from the carrier, the TCXO phase noise is sufficiently low to accurately calculate the LMH2180 contribution to the phase noise at the output. The LMH6559, whose phase noise contribution can be neglected, is used to drive the 50 Ω input impedance of the Signal Source Analyzer.

LAYOUT DESIGN RECOMMENDATION

Careful consideration during circuit design and PCB layout will eliminate problems and will optimize the performance of the LMH2180. It is best to have the same ground plane on the PCB for all decoupling and other ground connections.

To ensure a clean supply voltage it is best to place decoupling capacitors close to the LMH2180, between V_{DD} and V_{SS} .

Another important issue is the value of the components, because this also determines the sensitivity to disturbances. Resistor values have to be low enough to avoid a significant noise contribution and large enough to avoid a significant increase in power consumption while loading inputs or outputs to heavily.



Figure 40. Measurement Setup

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REVISION HISTORY

Cł	nanges from Revision C (March 2013) to Revision D	age
•	Changed layout of National Data Sheet to TI format	15

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PACKAGING INFORMATION

Orderable Device	Status	Package Type	Package	Pins	Package	Eco Plan	Lead/Ball Finish	MSL Peak Temp	Op Temp (°C)	Top-Side Markings	Samples
	(1)		Drawing		Qty	(2)		(3)		(4)	
LMH2180SD/NOPB	ACTIVE	WSON	NGQ	8	1000	Green (RoHS & no Sb/Br)	CU SN	Level-1-260C-UNLIM	-40 to 85	2180S	Samples
LMH2180SDE/NOPB	ACTIVE	WSON	NGQ	8	250	Green (RoHS & no Sb/Br)	CU SN	Level-1-260C-UNLIM	-40 to 85	2180S	Samples
LMH2180SDX/NOPB	ACTIVE	WSON	NGQ	8	4500	Green (RoHS & no Sb/Br)	CU SN	Level-1-260C-UNLIM	-40 to 85	2180S	Samples
LMH2180TM/NOPB	ACTIVE	DSBGA	YFQ	8	250	Green (RoHS & no Sb/Br)	SNAGCU	Level-1-260C-UNLIM		A	Samples
LMH2180TMX/NOPB	ACTIVE	DSBGA	YFQ	8	3000	Green (RoHS & no Sb/Br)	SNAGCU	Level-1-260C-UNLIM		A	Samples
LMH2180YD/NOPB	ACTIVE	SON	NGW	8	1000	Green (RoHS & no Sb/Br)	CU SNAGCU	Level-1-260C-UNLIM	-40 to 85	2180Y	Samples
LMH2180YDX/NOPB	ACTIVE	SON	NGW	8	4500	Green (RoHS & no Sb/Br)	CU SNAGCU	Level-1-260C-UNLIM	-40 to 85	2180Y	Samples

⁽¹⁾ The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

Pb-Free (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

⁽³⁾ MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

⁽⁴⁾ Multiple Top-Side Markings will be inside parentheses. Only one Top-Side Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Top-Side Marking for that device.



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PACKAGE MATERIALS INFORMATION

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TAPE AND REEL INFORMATION





QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



*All dimensions are nominal												
Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
LMH2180SD/NOPB	WSON	NGQ	8	1000	178.0	12.4	3.3	3.3	1.0	8.0	12.0	Q1
LMH2180SDE/NOPB	WSON	NGQ	8	250	178.0	12.4	3.3	3.3	1.0	8.0	12.0	Q1
LMH2180SDX/NOPB	WSON	NGQ	8	4500	330.0	12.4	3.3	3.3	1.0	8.0	12.0	Q1
LMH2180TM/NOPB	DSBGA	YFQ	8	250	178.0	8.4	1.35	1.35	0.76	4.0	8.0	Q1
LMH2180TMX/NOPB	DSBGA	YFQ	8	3000	178.0	8.4	1.35	1.35	0.76	4.0	8.0	Q1
LMH2180YD/NOPB	SON	NGW	8	1000	178.0	12.4	3.3	3.3	1.0	8.0	12.0	Q1
LMH2180YDX/NOPB	SON	NGW	8	4500	330.0	12.4	3.3	3.3	1.0	8.0	12.0	Q1

TEXAS INSTRUMENTS

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PACKAGE MATERIALS INFORMATION

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*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
LMH2180SD/NOPB	WSON	NGQ	8	1000	210.0	185.0	35.0
LMH2180SDE/NOPB	WSON	NGQ	8	250	210.0	185.0	35.0
LMH2180SDX/NOPB	WSON	NGQ	8	4500	367.0	367.0	35.0
LMH2180TM/NOPB	DSBGA	YFQ	8	250	210.0	185.0	35.0
LMH2180TMX/NOPB	DSBGA	YFQ	8	3000	210.0	185.0	35.0
LMH2180YD/NOPB	SON	NGW	8	1000	210.0	185.0	35.0
LMH2180YDX/NOPB	SON	NGW	8	4500	367.0	367.0	35.0

MECHANICAL DATA

NGQ0008A





MECHANICAL DATA

NGW0008A





YFQ0008



B. This drawing is subject to change without notice.



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