

LME49723 Dual High Fidelity Audio Operational Amplifier

Check for Samples: [LME49723](#)

FEATURES

- Easily Drives 600Ω Loads
- Optimized for Superior Audio Signal Fidelity
- Output Short Circuit Protection
- PSRR and CMRR Exceed 100dB (typ)
- SOIC Package

APPLICATIONS

- High Quality Audio Amplification
- High Fidelity Preamplifiers
- High Fidelity Multimedia
- Phono Pre Amps
- High Performance Professional Audio
- High Fidelity Equalization and Crossover Networks
- High Performance Line Drivers
- High Performance Line Receivers
- High Fidelity Active Filters

KEY SPECIFICATIONS

- Power Supply Voltage Range: ± 2.5 to ± 17 V
- THD+N ($A_V = 1$, $V_{OUT} = 3V_{RMS}$, $f_{IN} = 1$ kHz)
 - $R_L = 2k\Omega$: 0.0002 % (typ)
 - $R_L = 600\Omega$: 0.0002 % (typ)
- Input Noise Density: 3.6 nV/ \sqrt{Hz} (typ)
- Slew Rate: ± 8 V/ μs (typ)
- Gain Bandwidth Product: 17 MHz (typ)
- Open Loop Gain ($R_L = 600\Omega$): 105 dB (typ)
- Input Bias Current: 200 nA (typ)
- Input Offset Voltage: 0.3 mV (typ)

DESCRIPTION

The LME49723 is part of the ultra-low distortion, low noise, high slew rate operational amplifier series optimized and fully specified for high performance, high fidelity applications. Combining advanced leading-edge process technology with state-of-the-art circuit design, the LME49723 audio operational amplifiers deliver superior audio signal amplification for outstanding audio performance. The LME49723 combines extremely low voltage noise density (3.6nV/ \sqrt{Hz}) with vanishingly low THD+N (0.0002%) to easily satisfy the most demanding audio applications. To ensure that the most challenging loads are driven without compromise, the LME49723 has a high slew rate of $\pm 20V/\mu s$ and an output current capability of $\pm 26mA$. Further, dynamic range is maximized by an output stage that drives 2kΩ loads to within 1V of either power supply voltage and to within 1.4V when driving 600Ω loads.

The LME49723's outstanding CMRR (100dB), PSRR (100dB), and V_{OS} (0.3mV) give the amplifier excellent operational amplifier DC performance.

The LME49723 has a wide supply range of $\pm 2.5V$ to $\pm 17V$. Over this supply range the LME49723's input circuitry maintains excellent common-mode and power supply rejection, as well as maintaining its low input bias current. The LME49723 is unity gain stable.

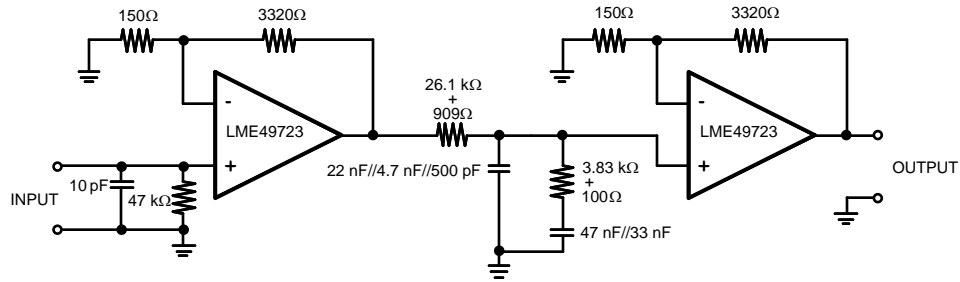
The LME49723 is available in an 8-lead narrow body SOIC package. Demonstration boards are available for each package.



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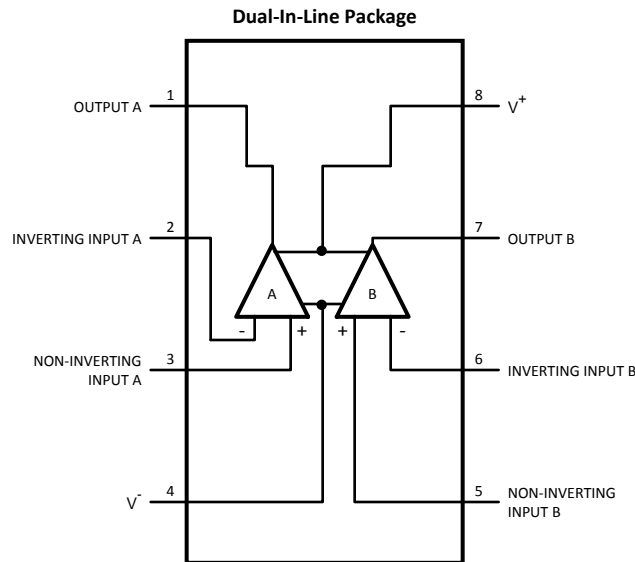
TYPICAL APPLICATION



Note: 1% metal film resistors, 5% polypropylene capacitors

Figure 1. Passively Equalized RIAA Phono Preamp

CONNECTION DIAGRAM



**Figure 2. SOIC Package
See Package Number D0008A**



These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

ABSOLUTE MAXIMUM RATINGS⁽¹⁾⁽²⁾⁽³⁾

Power Supply Voltage ($V_S = V^+ - V^-$)	36V
Storage Temperature	-65°C to 150°C
Input Voltage	(V-) - 0.7V to (V+) + 0.7V
Output Short Circuit ⁽⁴⁾	Continuous
Power Dissipation	Internally Limited
ESD Susceptibility ⁽⁵⁾	800V
ESD Susceptibility ⁽⁶⁾	180V
Junction Temperature	150°C
Thermal Resistance θ_{JA} (SO)	145°C/W
Temperature Range $T_{MIN} \leq T_A \leq T_{MAX}$	-40°C $\leq T_A \leq$ 85°C
Supply Voltage Range	$\pm 2.5V \leq V_S \leq \pm 17V$

- (1) Absolute Maximum Ratings indicate limits beyond which damage to the device may occur.
- (2) Operating Ratings indicate conditions for which the device is functional, but do not ensure specific performance limits. For ensured specifications and test conditions, see the Electrical Characteristics. The ensured specifications apply only for the test conditions listed. Some performance characteristics may degrade when the device is not operated under the listed test conditions.
- (3) If Military/Aerospace specified devices are required, please contact the Texas Instruments Sales Office/ Distributors for availability and specifications.
- (4) Amplifier output connected to GND, any number of amplifiers within a package.
- (5) Human body model, 100pF discharged through a 1.5k Ω resistor.
- (6) Machine Model ESD test is covered by specification EIAJ IC-121-1981. A 200pF cap is charged to the specified voltage and then discharged directly into the IC with no external series resistor (resistance of discharge path must be under 50 Ω).

ELECTRICAL CHARACTERISTICS FOR THE LME49723⁽¹⁾⁽²⁾

The specifications apply for $V_S = \pm 15V$, $R_L = 2k\Omega$, $f_{IN} = 1kHz$, $T_A = 25^\circ C$, unless otherwise specified.

Symbol	Parameter	Conditions	LME49723		Units (Limits)
			Typical ⁽³⁾	Limit ⁽⁴⁾	
THD+N	Total Harmonic Distortion + Noise	$A_V = 1$, $V_{OUT} = 3V_{RMS}$	0.0002		% (max)
		$R_L = 2k\Omega$ $R_L = 600\Omega$	0.0002	0.0004	
IMD	Intermodulation Distortion	$A_V = 1$, $V_{OUT} = 3V_{RMS}$ Two-tone, 60Hz & 7kHz 4:1	0.0005		%
GBWP	Gain Bandwidth Product		19	15	MHz (min)
SR	Slew Rate		± 8	± 6	V/ μs (min)
FPBW	Full Power Bandwidth	$V_{OUT} = 1V_{P-P}$, -3dB referenced to output magnitude at $f = 1kHz$	4		MHz
e_n	Equivalent Input Noise Voltage	$f_{BW} = 20Hz$ to 20kHz	0.45	0.65	μV_{RMS} (max)
	Equivalent Input Noise Density	$f = 1kHz$ $f = 10Hz$	3.2 8.5	5	nV/ \sqrt{Hz} (max)
i_n	Current Noise Density	$f = 1kHz$ $f = 10Hz$	0.7 1.3		pA/ \sqrt{Hz}
V_{OS}	Offset Voltage		± 0.3	1	mV (max)
$\Delta V_{OS}/\Delta T_{mp}$	Average Input Offset Voltage Drift vs Temperature	-40°C $\leq T_A \leq$ 85°C	0.2		$\mu V/^\circ C$
PSRR	Average Input Offset Voltage Shift vs Power Supply Voltage	$\Delta V_S = 20V^{(5)}$	100	95	dB (min)
ISO _{CH-CH}	Channel-to-Channel Isolation	$f_{IN} = 1kHz$ $f_{IN} = 20kHz$	118 112		dB
I_B	Input Bias Current	$V_{CM} = 0V$	200	300	nA (max)

- (1) Absolute Maximum Ratings indicate limits beyond which damage to the device may occur.
- (2) Operating Ratings indicate conditions for which the device is functional, but do not ensure specific performance limits. For ensured specifications and test conditions, see the Electrical Characteristics. The ensured specifications apply only for the test conditions listed. Some performance characteristics may degrade when the device is not operated under the listed test conditions.
- (3) Typical specifications are specified at +25°C and represent the most likely parametric norm.
- (4) Tested limits are specified to AOQL (Average Outgoing Quality Level).
- (5) PSRR is measured as follows: V_{OS} is measured at two supply voltages, $\pm 5V$ and $\pm 15V$. $PSRR = |20\log(\Delta V_{OS}/\Delta V_S)|$.

ELECTRICAL CHARACTERISTICS FOR THE LME49723⁽¹⁾⁽²⁾ (continued)

The specifications apply for $V_S = \pm 15V$, $R_L = 2k\Omega$, $f_{IN} = 1kHz$, $T_A = 25^\circ C$, unless otherwise specified.

Symbol	Parameter	Conditions	LME49723		Units (Limits)
			Typical ⁽³⁾	Limit ⁽⁴⁾	
$\Delta I_{OS}/\Delta T_{e\ mp}$	Input Bias Current Drift vs Temperature	$-40^\circ C \leq T_A \leq 85^\circ C$	0.1		nA/°C
I_{OS}	Input Offset Current	$V_{CM} = 0V$	7	100	nA (max)
V_{IN-CM}	Common-Mode Input Voltage Range		± 14	(V+) – 2.0 (V-) + 2.0	V (min)
CMRR	Common-Mode Rejection	$-10V < V_{cm} < 10V$	100	90	dB (min)
Z_{IN}	Differential Input Impedance		30		k Ω
	Common Mode Input Impedance	$-10V < V_{cm} < 10V$	1000		M Ω
A_{VOL}	Open Loop Voltage Gain	$-10V < V_{out} < 10V$, $R_L = 600\Omega$	100	98	dB (min)
		$-10V < V_{out} < 10V$, $R_L = 2k\Omega$	105		
		$-10V < V_{out} < 10V$, $R_L = 10k\Omega$	105		
V_{OUTMAX}	Maximum Output Voltage Swing	$R_L = 600\Omega$	± 13.5	± 12.5	V (min)
		$R_L = 2k\Omega$	± 14.0		
		$R_L = 10k\Omega$	± 14.1		
I_{OUT}	Output Current	$R_L = 600\Omega$, $V_S = \pm 17V$	± 25	± 21	mA (min)
I_{OUT-CC}	Instantaneous Short Circuit Current		+53 –42		mA
R_{OUT}	Output Impedance	$f_{IN} = 10kHz$ Closed-Loop Open-Loop	0.01 13		Ω
C_{LOAD}	Capacitive Load Drive Overshoot	100pF	16		%
I_S	Total Quiescent Current	$I_{OUT} = 0mA$	6.7	7.5	mA (max)

TYPICAL PERFORMANCE CHARACTERISTICS

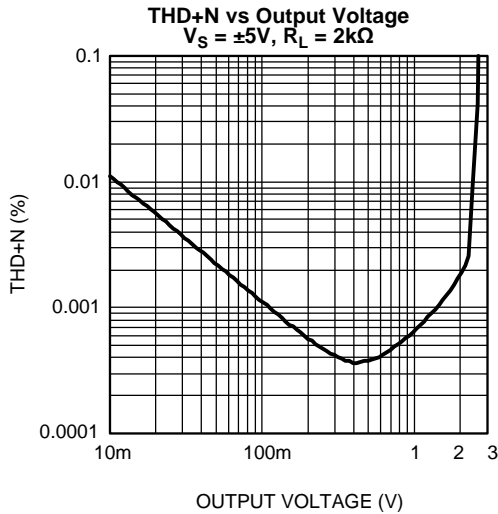


Figure 3.

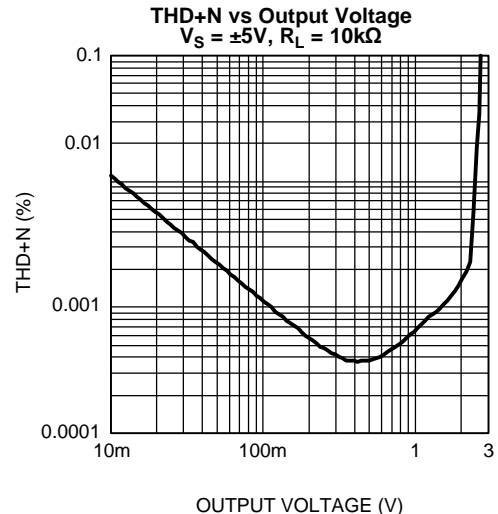


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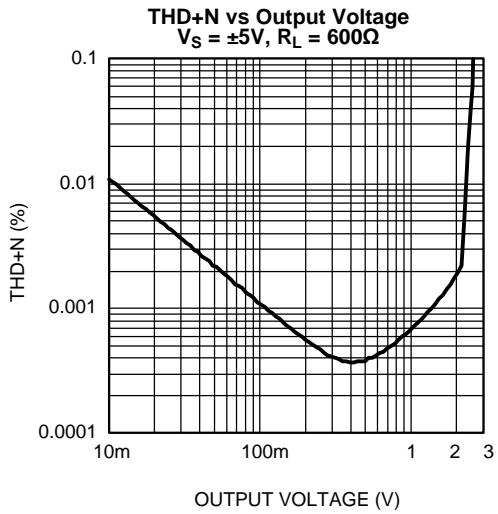


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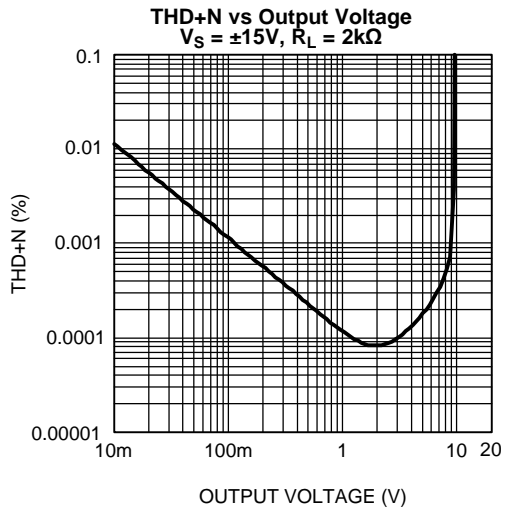


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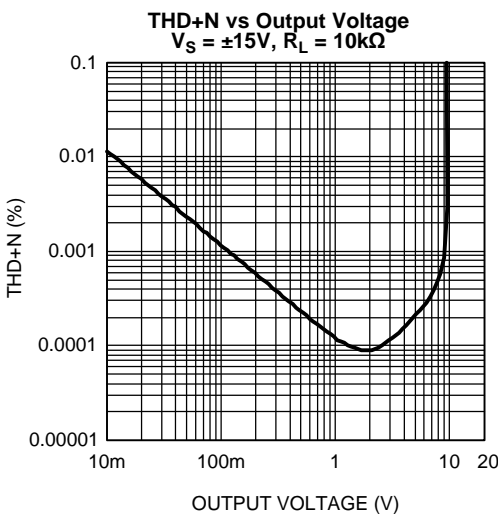


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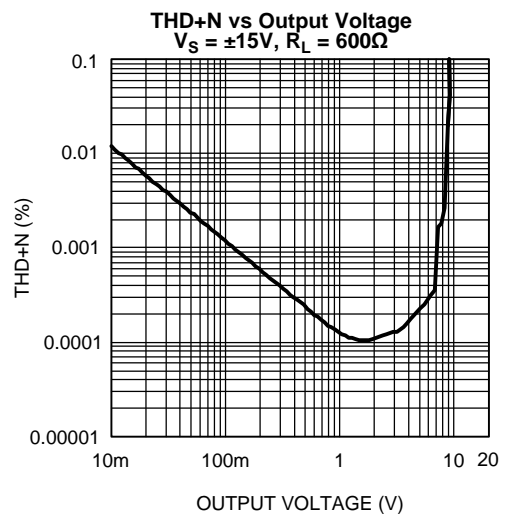


Figure 8.

TYPICAL PERFORMANCE CHARACTERISTICS (continued)

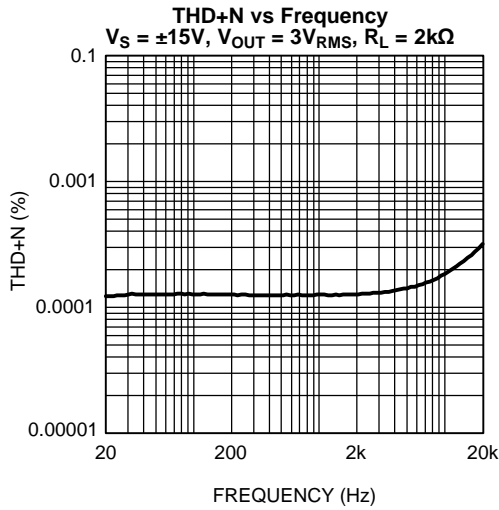


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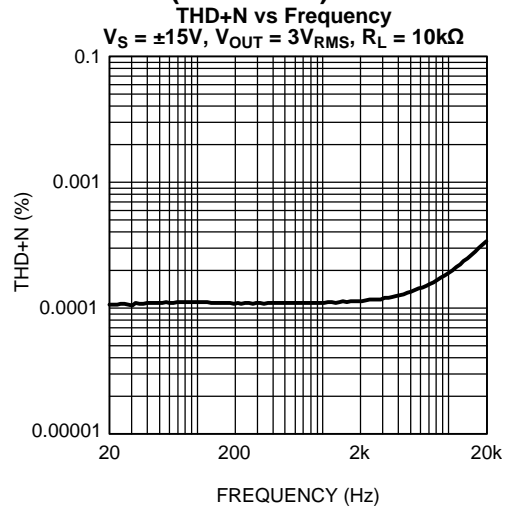


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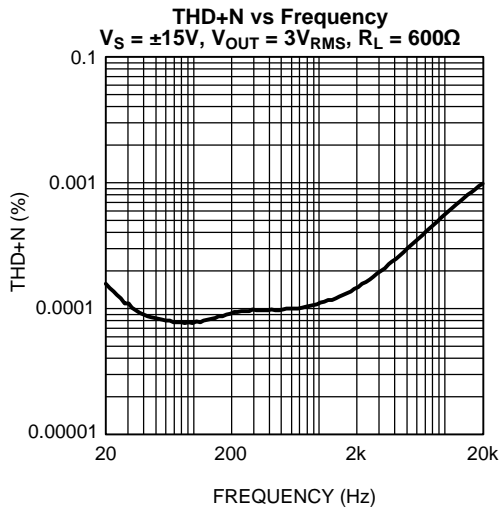


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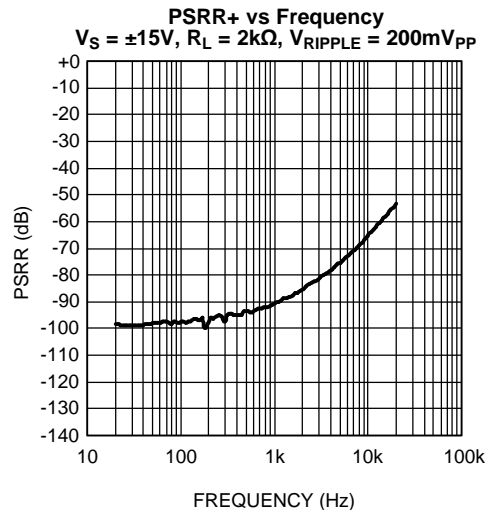


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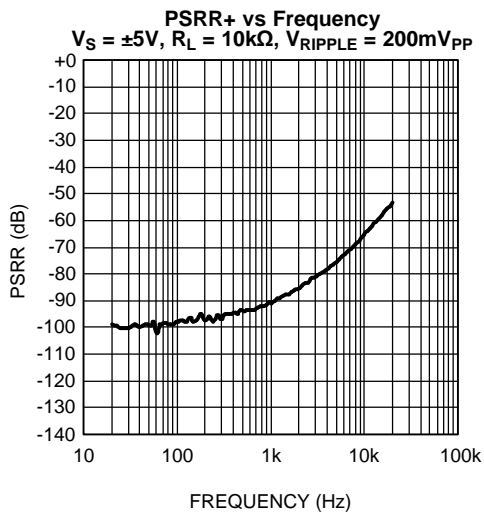


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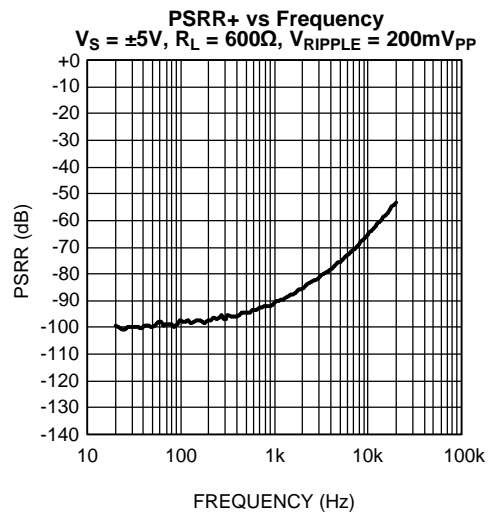


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TYPICAL PERFORMANCE CHARACTERISTICS (continued)

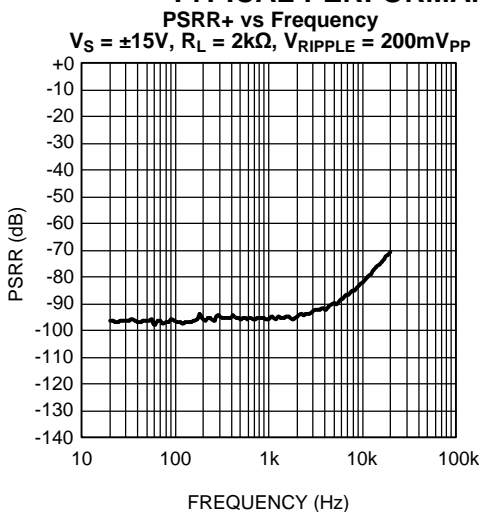


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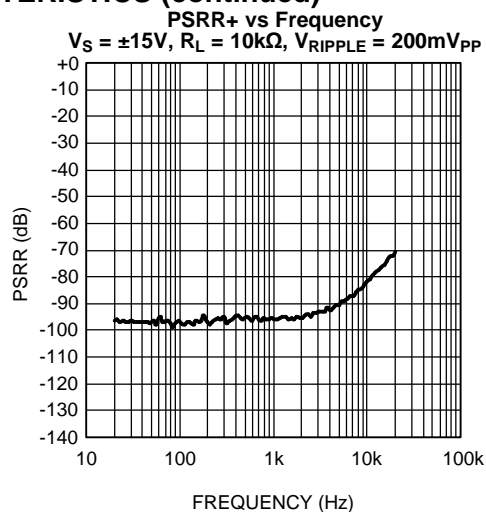


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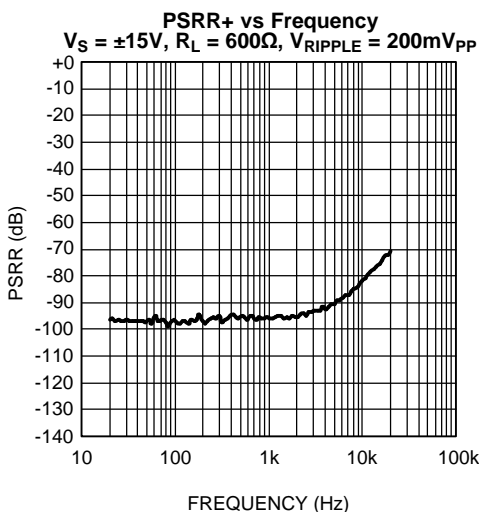


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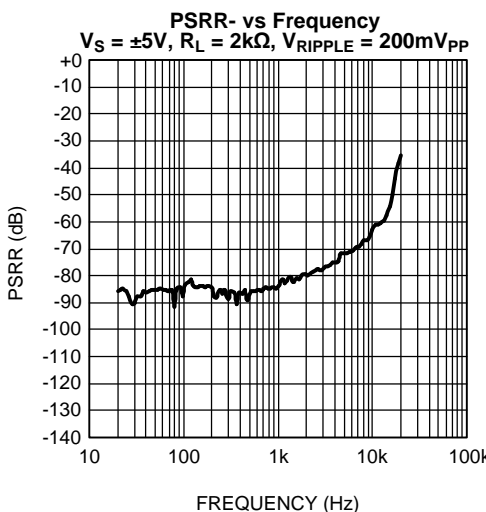


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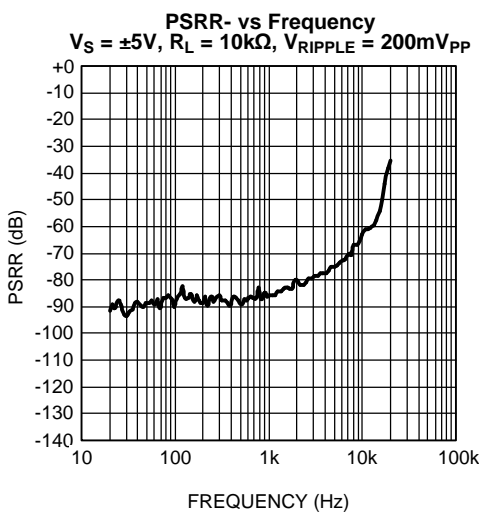


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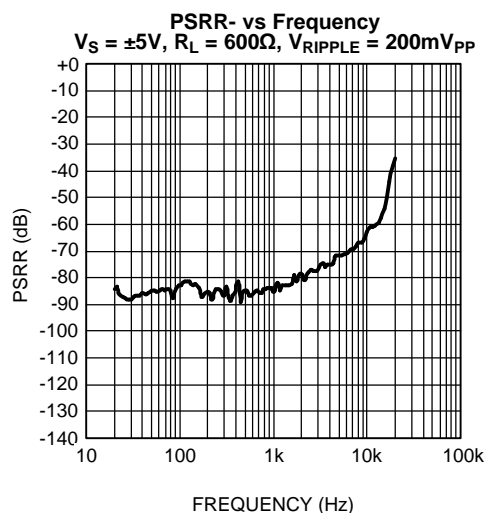


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TYPICAL PERFORMANCE CHARACTERISTICS (continued)

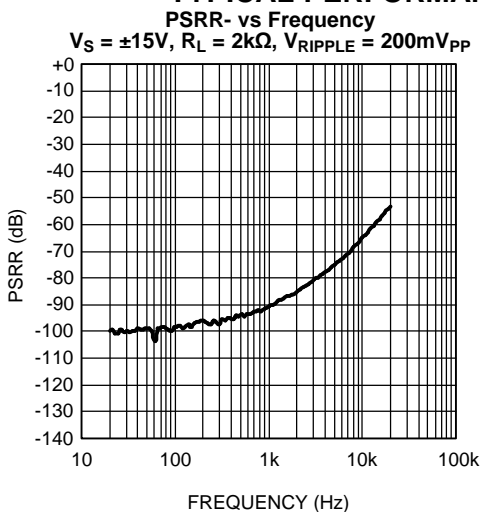


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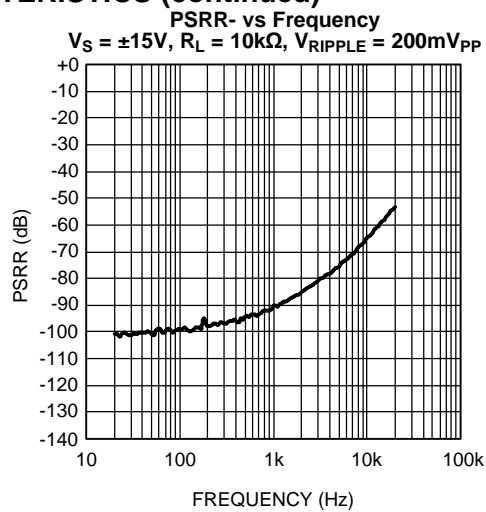


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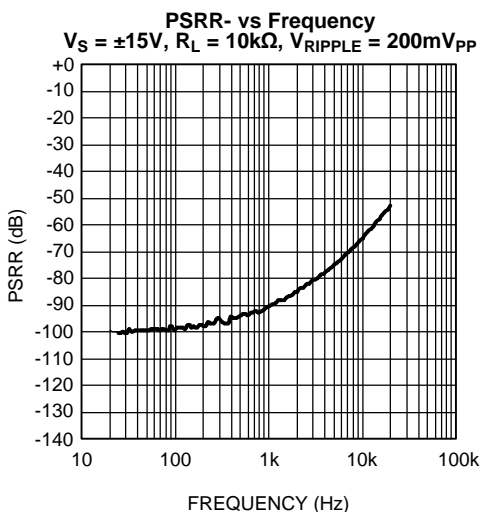


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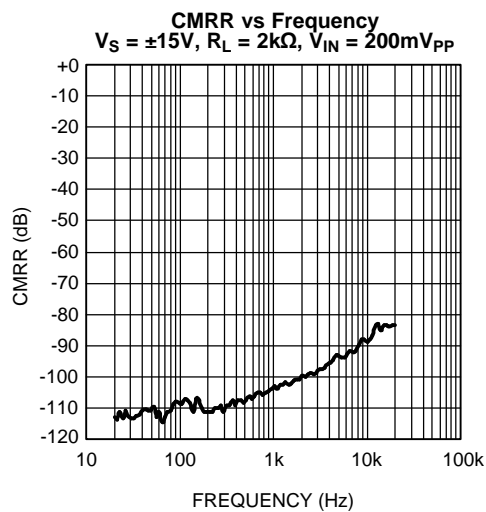


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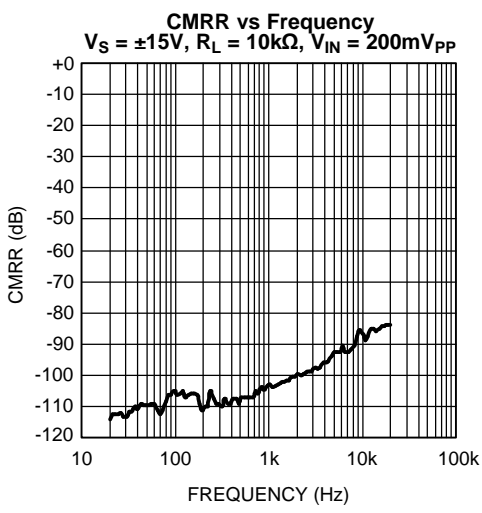


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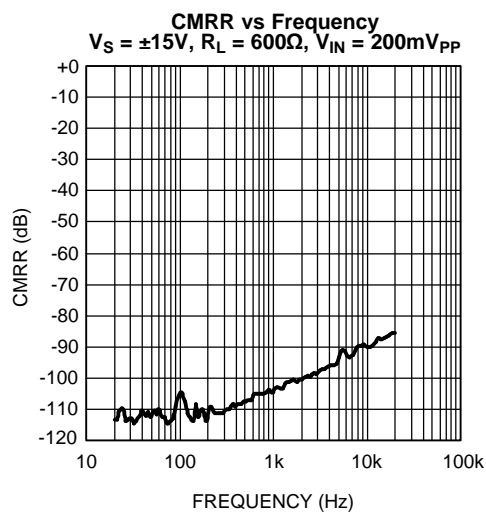


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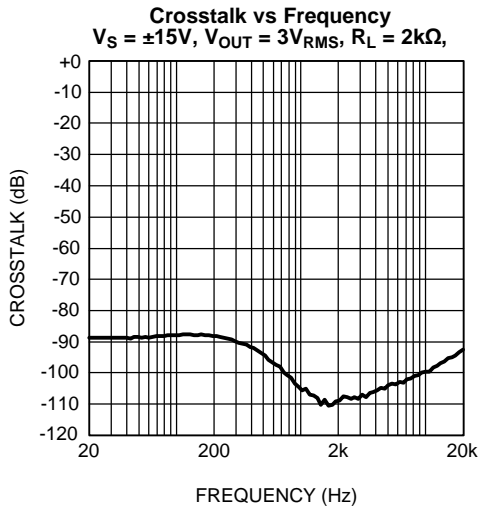


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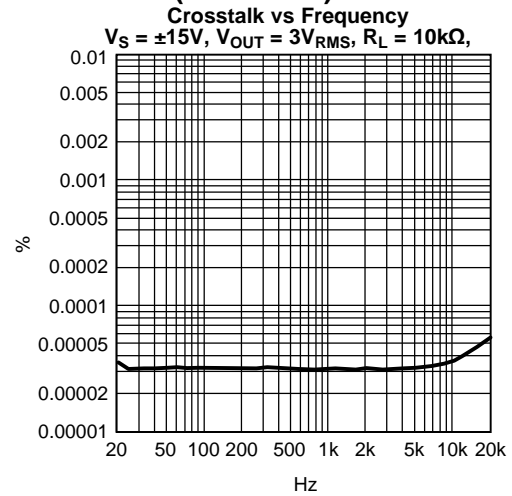


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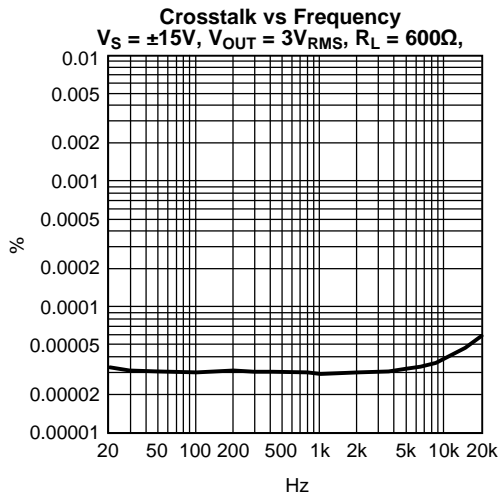


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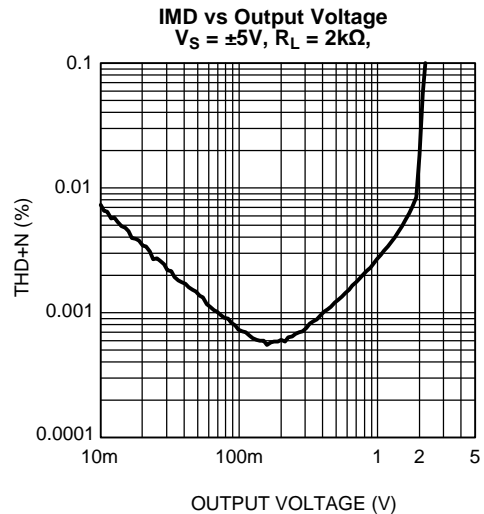


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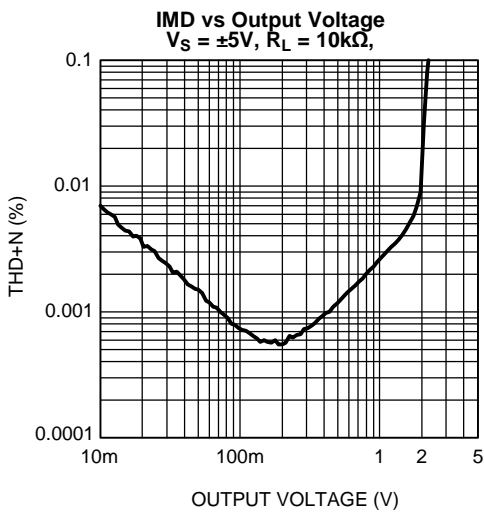


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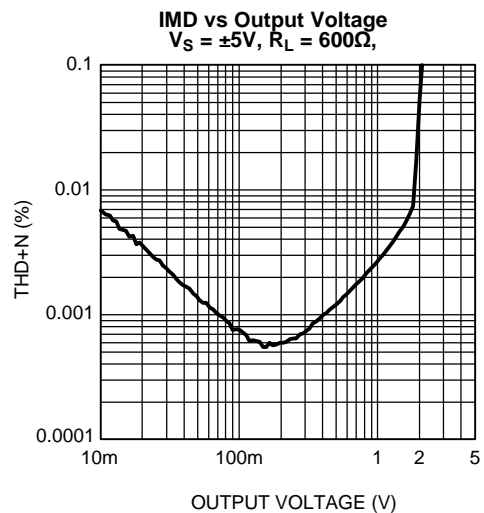


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TYPICAL PERFORMANCE CHARACTERISTICS (continued)

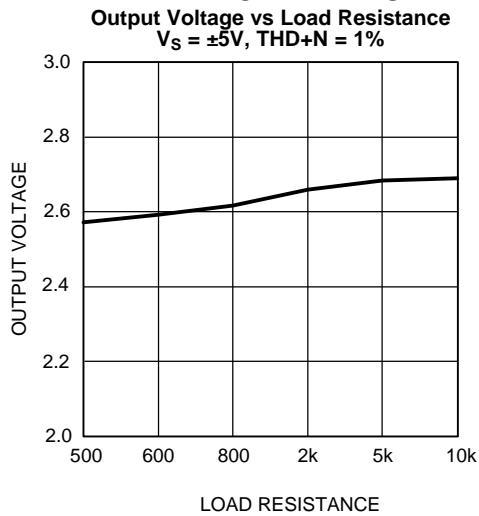


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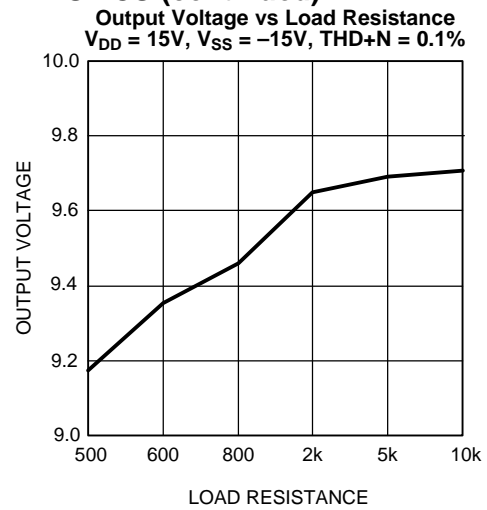


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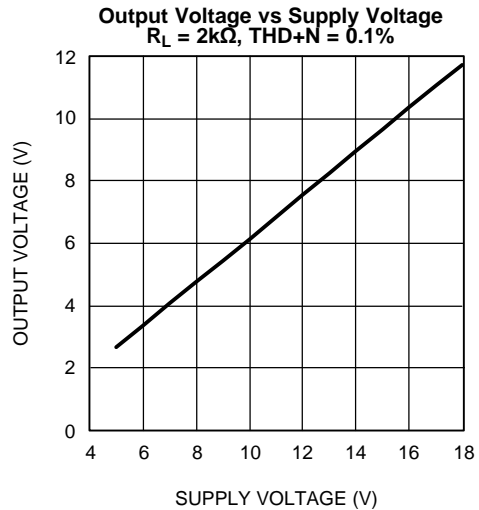


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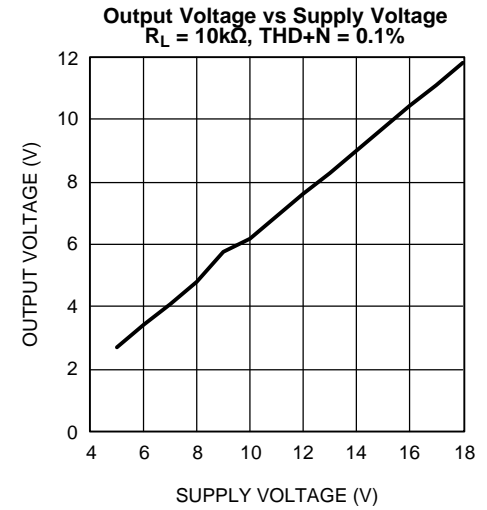


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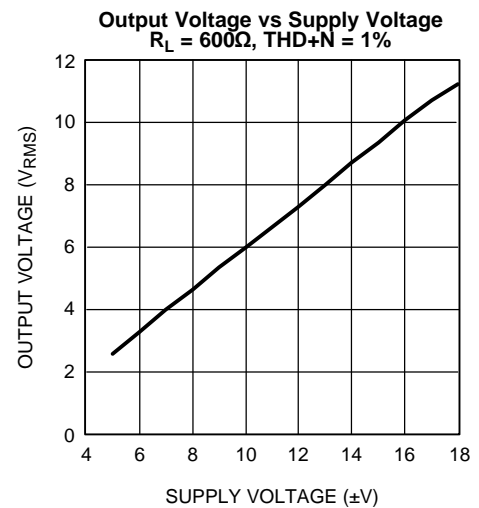


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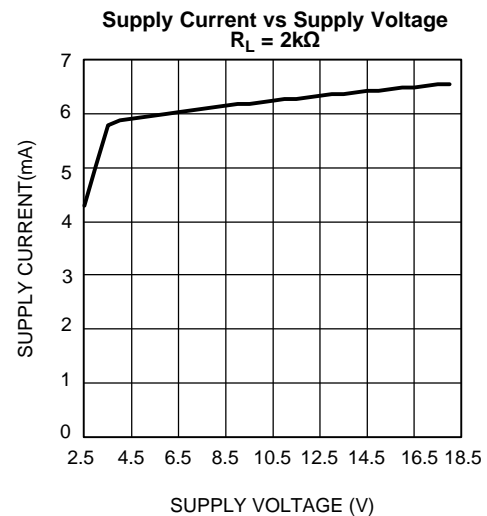


Figure 38.

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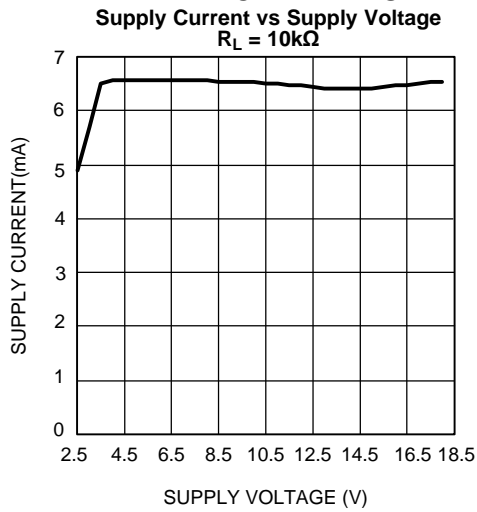


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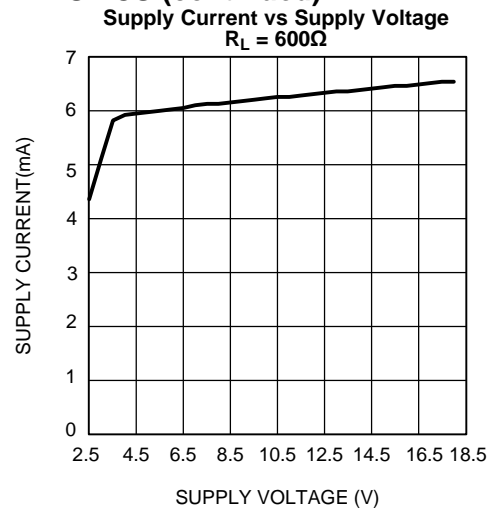
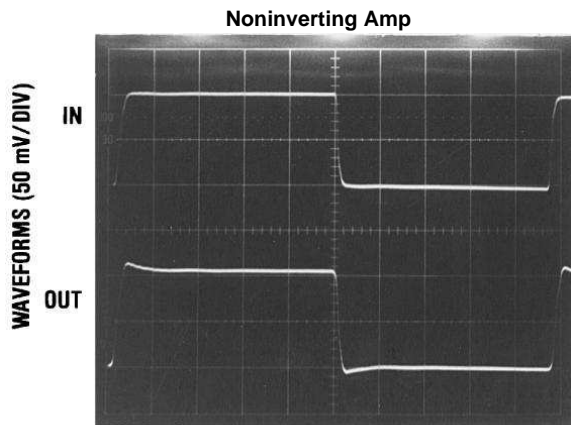
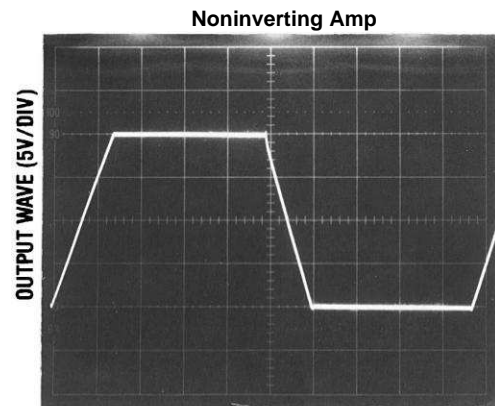


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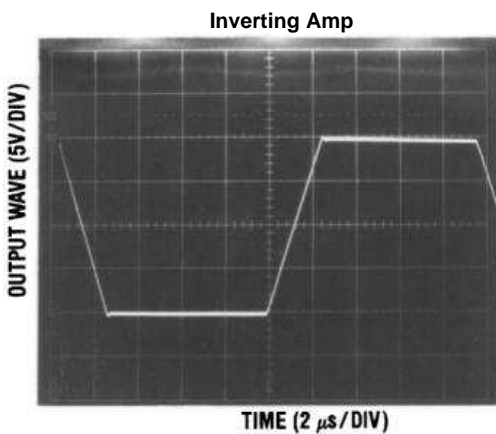
TIME (0.2 μ s/DIV)

Figure 41.



TIME (2 μ s/DIV)

Figure 42.



TIME (2 μ s/DIV)

Figure 43.

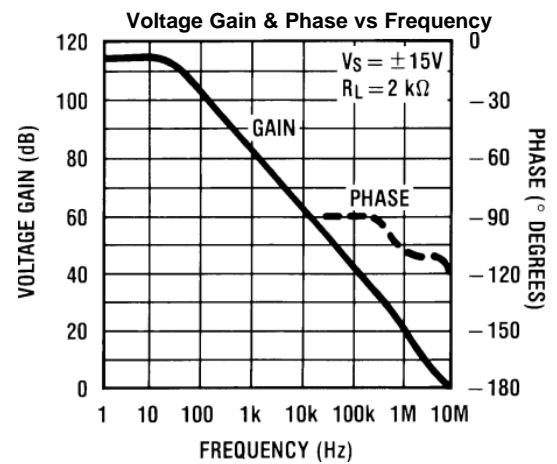


Figure 44.

TYPICAL PERFORMANCE CHARACTERISTICS (continued)

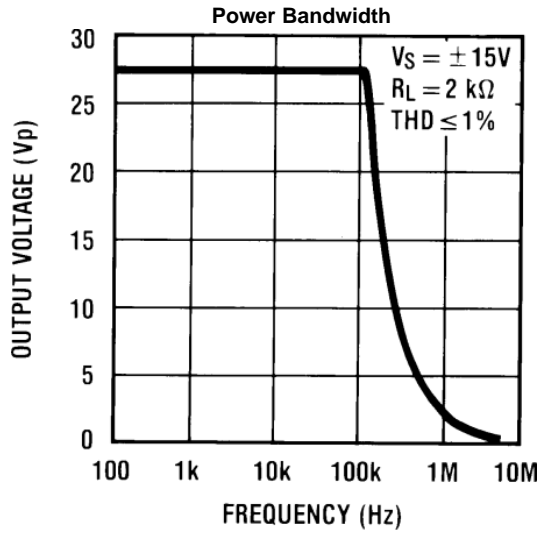


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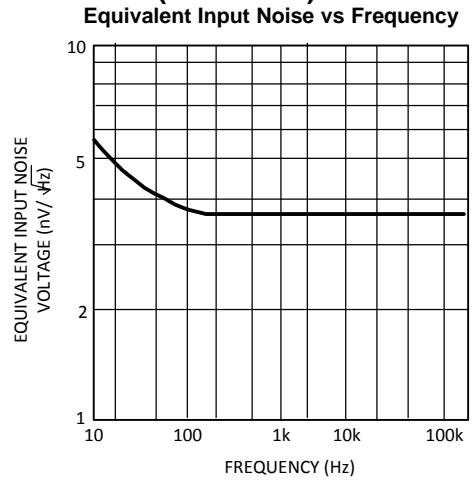


Figure 46.

APPLICATION INFORMATION

DISTORTION MEASUREMENTS

The vanishingly low residual distortion produced by LME49723 is below the capabilities of all commercially available equipment. This makes distortion measurements just slightly more difficult than simply connecting a distortion meter to the amplifier's inputs and outputs. The solution, however, is quite simple: an additional resistor. Adding this resistor extends the resolution of the distortion measurement equipment.

The LME49723's low residual distortion is an input referred internal error. As shown in Figure 47, adding the 10Ω resistor connected between the amplifier's inverting and non-inverting inputs changes the amplifier's noise gain. The result is that the error signal (distortion) is amplified by a factor of 101. Although the amplifier's closed-loop gain is unaltered, the feedback available to correct distortion errors is reduced by 101, which means that measurement resolution increases by 101. To ensure minimum effects on distortion measurements, keep the value of R1 low as shown in Figure 47.

This technique is verified by duplicating the measurements with high closed loop gain and/or making the measurements at high frequencies. Doing so produces distortion components that are within the measurement equipment's capabilities. This datasheet's THD+N and IMD values were generated using the above described circuit connected to an Audio Precision System Two Cascade.

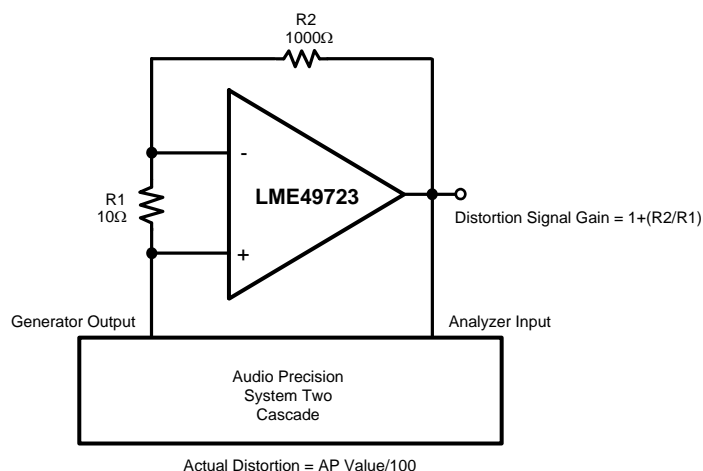
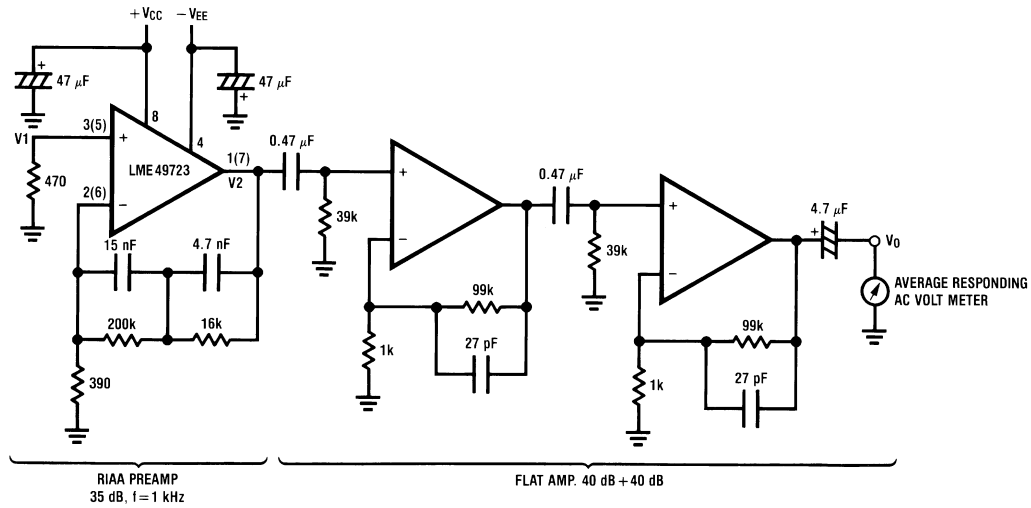


Figure 47. THD+N and IMD Distortion Test Circuit

The LME49723 is a high speed op amp with excellent phase margin and stability. Capacitive loads up to 100pF will cause little change in the phase characteristics of the amplifiers and are therefore allowable.

Capacitive loads greater than 100pF must be isolated from the output. The most straightforward way to do this is to put a resistor in series with the output. This resistor will also prevent excess power dissipation if the output is accidentally shorted.



Complete shielding is required to prevent induced pick up from external sources. Always check with oscilloscope for power line noise.

**Figure 48. Noise Measurement Circuit Total Gain: 115 dB @f = 1 kHz
Input Referred Noise Voltage: $e_n = V_0/560,000$ (V)**

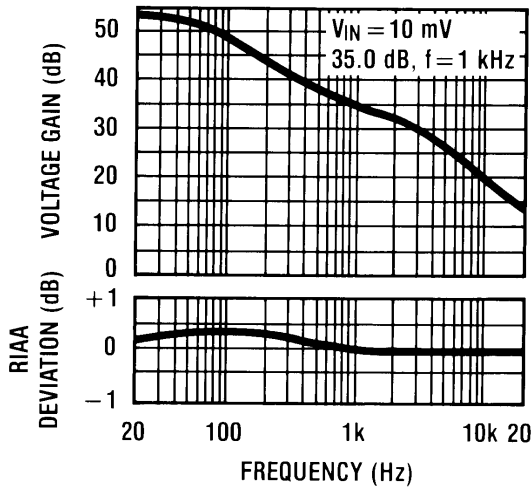


Figure 49. RIAA Preamp Voltage Gain, RIAA Deviation vs Frequency

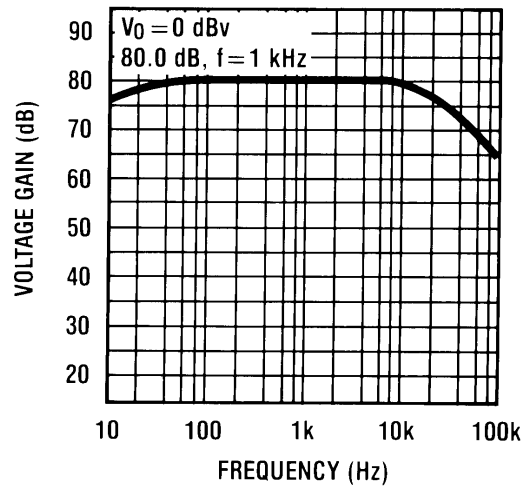
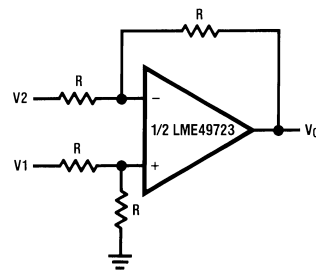


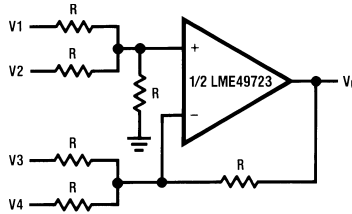
Figure 50. Flat Amp Voltage Gain vs Frequency

TYPICAL APPLICATIONS



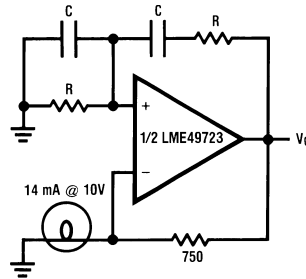
$V_O = V_1 - V_2$

Figure 51. Balanced to Single Ended Converter



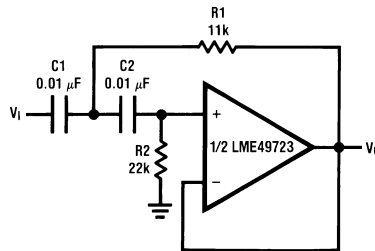
$$V_O = V_1 + V_2 - V_3 - V_4$$

Figure 52. Adder/Subtractor



$$f_o = \frac{1}{2\pi RC}$$

Figure 53. Sine Wave Oscillator



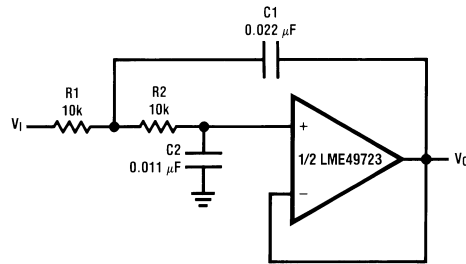
if $C_1 = C_2 = C$

$$R_1 = \frac{\sqrt{2}}{2\omega_o C}$$

$$R_2 = 2 \cdot R_1$$

Illustration is $f_o = 1 \text{ kHz}$

Figure 54. Second Order High Pass Filter (Butterworth)



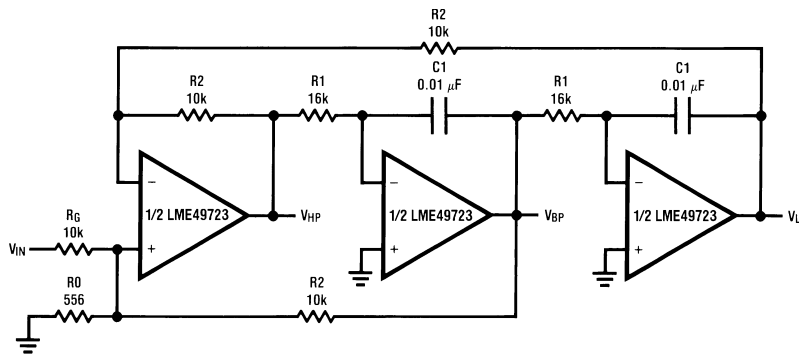
if $R1 = R2 = R$

$$C1 = \frac{\sqrt{2}}{\omega_0 R}$$

$$C2 = \frac{C1}{2}$$

Illustration is $f_0 = 1 \text{ kHz}$

Figure 55. Second Order Low Pass Filter (Butterworth)



$$f_0 = \frac{1}{2\pi C1 R1}, Q = \frac{1}{2} \left(1 + \frac{R2}{R0} + \frac{R2}{Rg} \right), A_{BP} = Q A_{LP} = Q A_{LH} = \frac{R2}{Rg}$$

Illustration is $f_0 = 1 \text{ kHz}, Q = 10, A_{BP} = 1$

Figure 56. State Variable Filter

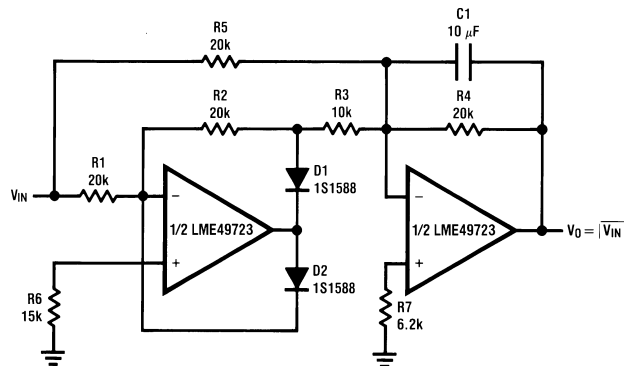


Figure 57. AC/DC Converter

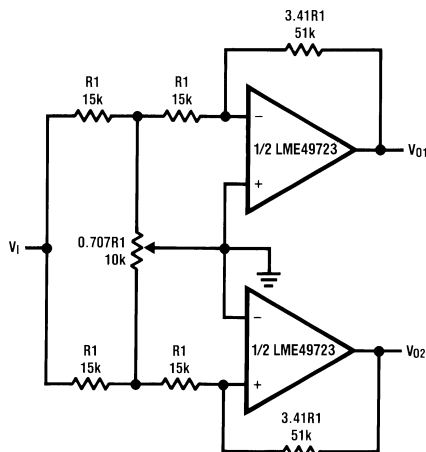


Figure 58. 2 Channel Panning Circuit (Pan Pot)

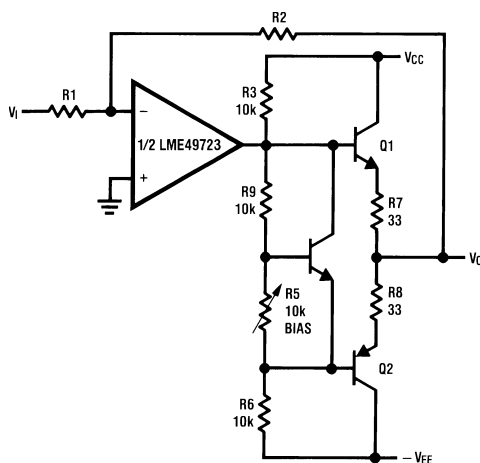
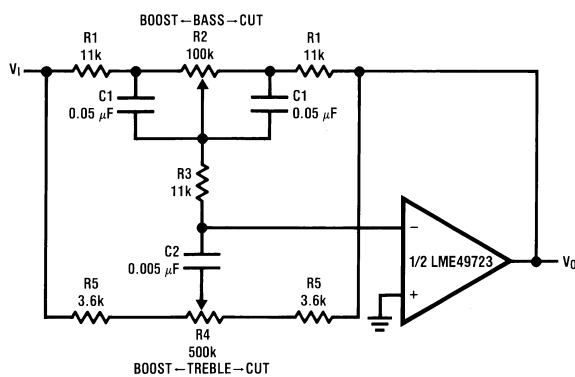


Figure 59. Line Driver



$$f_L = \frac{1}{2\pi R_2 C_1}, f_{LB} = \frac{1}{2\pi R_1 C_1}$$

$$f_H = \frac{1}{2\pi R_5 C_2}, f_{HB} = \frac{1}{2\pi (R_1 + R_5 + 2R_3) C_2}$$

Illustration is:

$$f_L = 32 \text{ Hz}, f_{LB} = 320 \text{ Hz}$$

$$f_H = 11 \text{ kHz}, f_{HB} = 1.1 \text{ kHz}$$

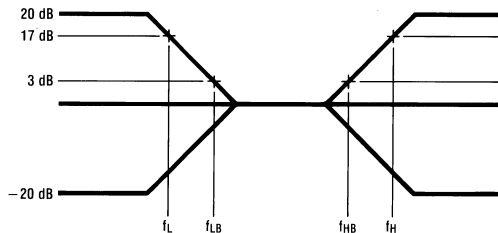
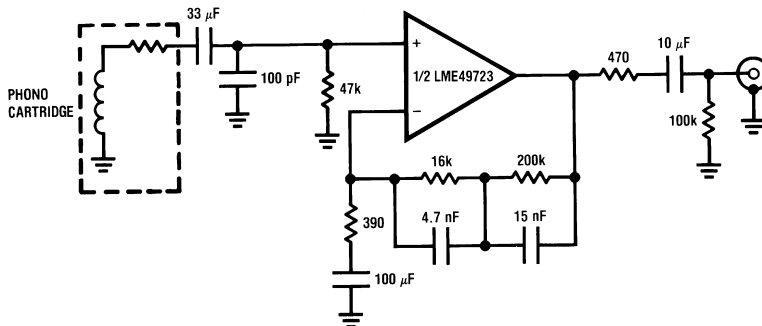
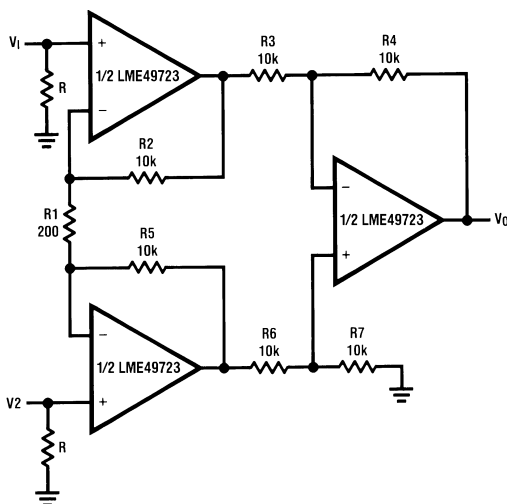


Figure 60. Tone Control



$A_v = 35 \text{ dB}$
 $E_n = 0.33 \mu\text{V}$
 $S/N = 90 \text{ dB}$
 $f = 1 \text{ kHz}$
 A Weighted
 A Weighted, $V_{IN} = 10 \text{ mV}$
 @ $f = 1 \text{ kHz}$

Figure 61. RIAA Preamp



If $R_2 = R_5, R_3 = R_6, R_4 = R_7$

$$V_0 = \left(1 + \frac{2R_2}{R_1}\right) \frac{R_4}{R_3} (V_2 - V_1)$$
 Illustration is:

$$V_0 = 101(V_2 - V_1)$$

Figure 62. Balanced Input Mic Amp

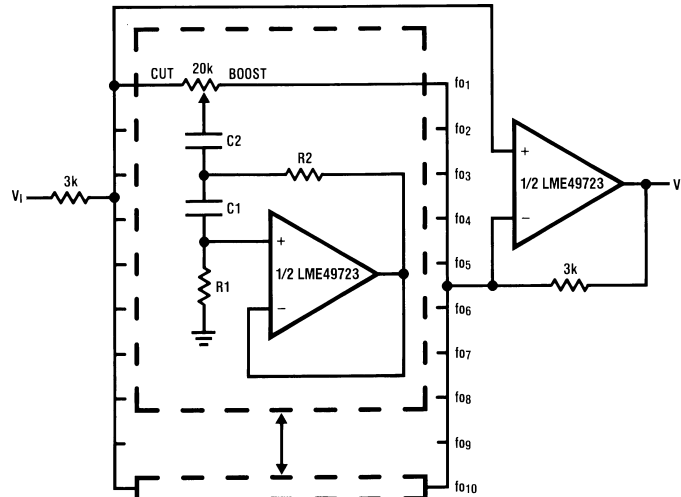


Figure 63. Band Graphic Equalizer

fo (Hz)	C ₁	C ₂	R ₁	R ₂
32	0.12μF	4.7μF	75kΩ	500Ω
64	0.056μF	3.3μF	68kΩ	510Ω
125	0.033μF	1.5μF	62kΩ	510Ω
250	0.015μF	0.82μF	68kΩ	470Ω
500	8200pF	0.39μF	62kΩ	470Ω
1k	3900pF	0.22μF	68kΩ	470Ω
2k	2000pF	0.1μF	68kΩ	470Ω
4k	1100pF	0.056μF	62kΩ	470Ω
8k	510pF	0.022μF	68kΩ	510Ω
16k	330pF	0.012μF	51kΩ	510Ω

REVISION HISTORY

Rev	Date	Description
1.0	01/07/08	Initial release.
1.01	02/11/08	Text edits.
B	04/04/13	Changed layout of National Data Sheet to TI format.

PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead/Ball Finish	MSL Peak Temp (3)	Op Temp (°C)	Top-Side Markings (4)	Samples
LME49723MA/NOPB	ACTIVE	SOIC	D	8	95	Green (RoHS & no Sb/Br)	CU SN	Level-1-260C-UNLIM	-40 to 85	L49723 MA	Samples
LME49723MAX/NOPB	ACTIVE	SOIC	D	8	2500	Green (RoHS & no Sb/Br)	CU SN	Level-1-260C-UNLIM	-40 to 85	L49723 MA	Samples

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check <http://www.ti.com/productcontent> for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

Pb-Free (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) Multiple Top-Side Markings will be inside parentheses. Only one Top-Side Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Top-Side Marking for that device.

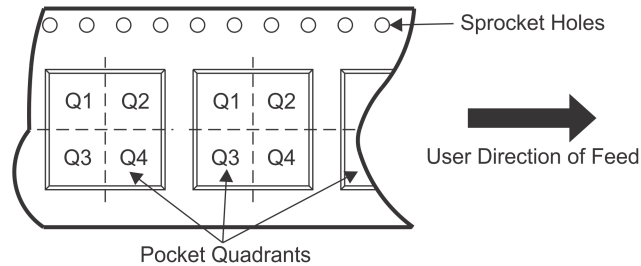
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TAPE AND REEL INFORMATION



QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
LME49723MAX/NOPB	SOIC	D	8	2500	330.0	12.4	6.5	5.4	2.0	8.0	12.0	Q1

TAPE AND REEL BOX DIMENSIONS

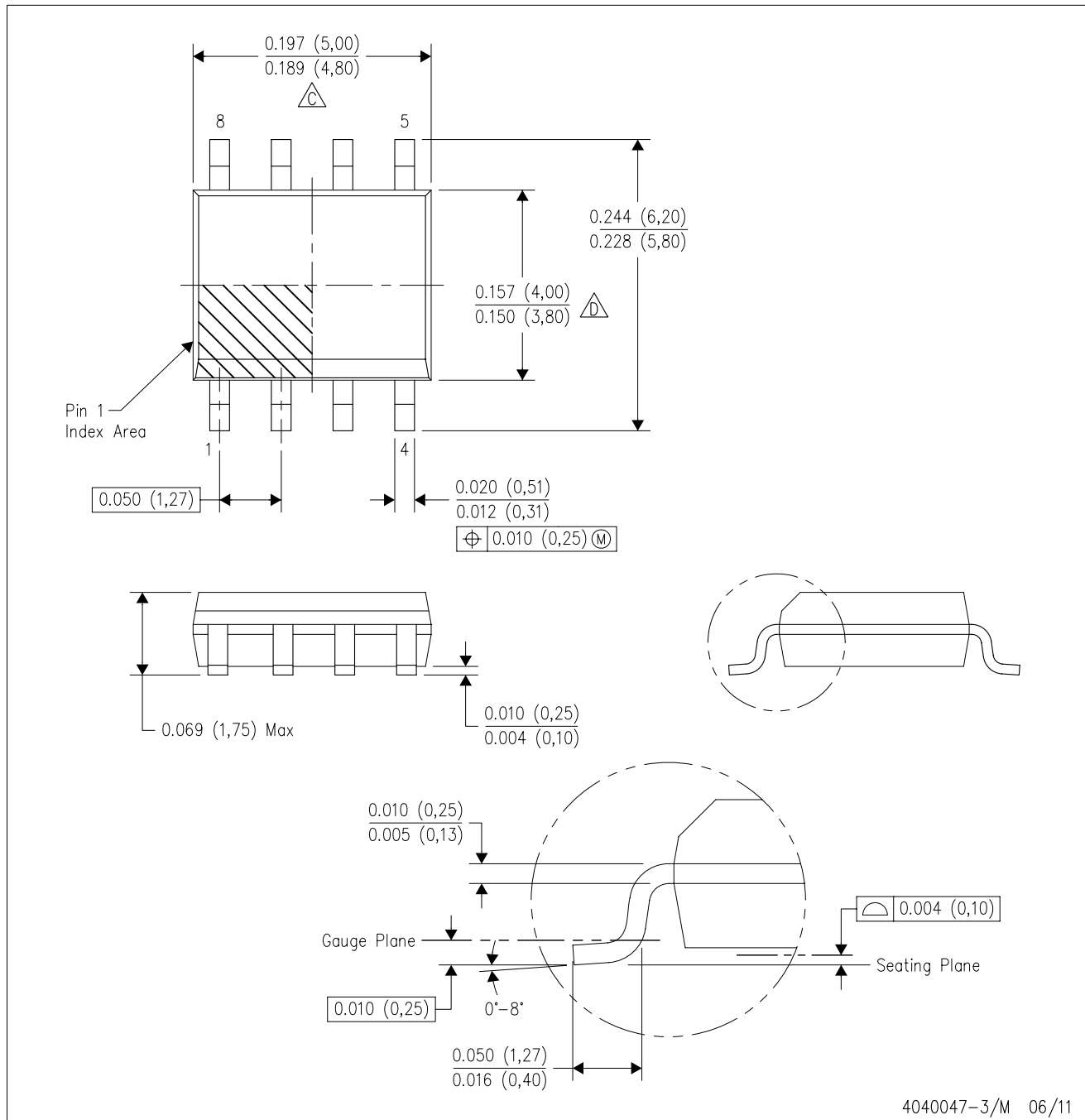


*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
LME49723MAX/NOPB	SOIC	D	8	2500	349.0	337.0	45.0

D (R-PDSO-G8)

PLASTIC SMALL OUTLINE



4040047-3/M 06/11

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