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LMC7111 Tiny CMOS Operational Amplifier with Rail-to-Rail Input and Output

Check for Samples: LMC7111

FEATURES

- Tiny 5-Pin SOT-23 Package Saves Space
- Very Wide Common Mode Input Range
- Specified at 2.7V, 5V, and 10V
- Typical Supply Current 25 µA at 5V
- 50 kHz Gain-Bandwidth at 5V
- Similar to Popular LMC6462
- Output to Within 20 mV of Supply Rail at 100k
 Load
- Good Capacitive Load Drive

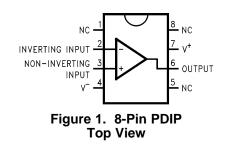
APPLICATIONS

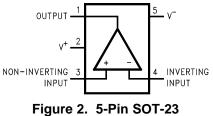
- Mobile Communications
- Portable Computing
- Current Sensing for Battery Chargers
- Voltage Reference Buffering
- Sensor Interface
- Stable Bias for GaAs RF Amps

Connection Diagram



The LMC7111 is a micropower CMOS operational amplifier available in the space saving SOT-23 package. This makes the LMC7111 ideal for space and weight critical designs. The wide common-mode input range makes it easy to design battery monitoring circuits which sense signals above the V⁺ supply. The main benefits of the Tiny package are most apparent in small portable electronic devices, such as mobile phones, pagers, and portable computers. The tiny amplifiers can be placed on a board where they are needed, simplifying board layout.





Top View

Figure 3. Actual Size

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LMC7111

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These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

Absolute Maximum Ratings ⁽¹⁾⁽²⁾

ESD Tolerance (3)	SOT-23 Package	2000V		
	PDIP Package	1500V		
Differential Input Voltage		±Supply Voltage		
Voltage at Input/Output Pin		(V ⁺) + 0.3V, (V [−]) − 0.3V		
Supply Voltage (V ⁺ – V ⁻)		11V		
Current at Input Pin		±5 mA		
Current at Output Pin ⁽⁴⁾		±30 mA		
Current at Power Supply Pin		30 mA		
Lead Temp. (Soldering, 10 sec.))	260°C		
Storage Temperature Range		−65°C to +150°C		
Junction Temperature ⁽⁵⁾		150°C		

(1) Absolute Maximum Ratings indicate limits beyond which damage to the device may occur. Operating Ratings indicate conditions for which the device is intended to be functional, but specific performance is not ensured. For ensured specifications and the test conditions, see the Electrical Characteristics.

(2) If Military/Aerospace specified devices are required, please contact the TI Sales Office/ Distributors for availability and specifications.

(3) Human Body Model is 1.5 kΩ in series with 100 pF.

(4) Applies to both single-supply and split-supply operation. Continuous short circuit operation at elevated ambient temperature can result in exceeding the maximum allowed junction temperature at 150°C.

(5) The maximum power dissipation is a function of $T_{J(MAX)}$, θ_{JA} and T_A . The maximum allowable power dissipation at any ambient temperature is $P_D = (T_{J(MAX)} - T_A)/\theta_{JA}$. All numbers apply for packages soldered directly into a PC board.

Operating Ratings ⁽¹⁾

Supply Voltage		2.5V ≤ V ⁺ ≤ 11V
Junction Temperature Range	LMC7111AI, LMC7111BI	-40°C ≤ T _J ≤ +85°C
Thermal Resistance (θ _{JA})	8-Pin PDIP	115°C/W
	5-Pin SOT-23	325°C/W

(1) Absolute Maximum Ratings indicate limits beyond which damage to the device may occur. Operating Ratings indicate conditions for which the device is intended to be functional, but specific performance is not specified. For ensured specifications and the test conditions, see the Electrical Characteristics.

2.7V DC Electrical Characteristics

Unless otherwise specified, all limits specified for $T_J = 25^{\circ}C$, $V^+ = 2.7V$, $V^- = 0V$, $V_{CM} = V_O = V^+/2$ and $R_L > 1 \text{ M}\Omega$. **Boldface** limits apply at the temperature extremes.

Symbol	Parameter	Conditions	Тур ⁽¹⁾	LMC7111AI Limit ⁽²⁾	LMC7111BI Limit ⁽²⁾	Units
V _{OS}	Input Offset Voltage	V ⁺ = 2.7V	0.9	3	7	mV
				5	9	max
TCV _{OS}	Input Offset Voltage Average Drift		2.0			µV/°C
Ι _Β	Input Bias Current	See ⁽³⁾	0.1	1 20	1 20	pA max
I _{OS}	Input Offset Current	See ⁽³⁾	0.01	0.5 10	0.5 10	pA max
R _{IN}	Input Resistance		>10			Tera Ω
+PSRR	Positive Power Supply Rejection Ratio	$2.7V \le V^+ \le 5.0V,$ $V^- = 0V, V_0 = 2.5V$	60	55 50	55 50	dB min
-PSRR	Negative Power Supply Rejection Ratio	$-2.7V \le V^- \le -5.0V,$ $V^- = 0V, V_0 = 2.5V$	60	55 50	55 50	dB min

(1) Typical Values represent the most likely parametric norm.

(2) All limits are specified by testing or statistical analysis.

(3) Bias Current specified by design and processing.



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2.7V DC Electrical Characteristics (continued)

Unless otherwise specified, all limits specified for $T_J = 25^{\circ}C$, $V^+ = 2.7V$, $V^- = 0V$, $V_{CM} = V_O = V^+/2$ and $R_L > 1 \text{ M}\Omega$. **Boldface** limits apply at the temperature extremes.

Symbol	Parameter	Conditions	Тур ⁽¹⁾	LMC7111AI Limit ⁽²⁾	LMC7111BI Limit ⁽²⁾	Units
	Input Common-Mode Voltage Range	V ⁺ = 2.7V For CMRR ≥ 50 dB	-0.10	0.0 0.40	0.0 0.40	V min
			2.8	2.7 2.25	2.7 2.25	V max
C _{IN}	Common-Mode Input Capacitance		3			pF
Vo	Output Swing	$V^+ = 2.7V$ R _L = 100 kΩ	2.69	2.68 2.4	2.68 2.4	V min
			0.01	0.02 0.08	0.02 0.08	V max
		$V^+ = 2.7V$ R _L = 10 k Ω	2.65	2.6 2.4	2.6 2.4	V min
			0.03	0.1 0.3	0.1 0.3	V max
I _{SC}	Output Short Circuit Current	Sourcing, $V_0 = 0V$	7	1 0.7	1 0.7	mA min
		Sinking, $V_0 = 2.7V$	7	1 0.7	1 0.7	mA min
A _{VOL}	Voltage Gain	Sourcing	400			V/mv min
		Sinking	150			V/mv min
I _S	Supply Current	$V^+ = +2.7V,$ $V_0 = V^+/2$	20	45 60	50 65	μA max

2.7V AC Electrical Characteristics

Unless otherwise specified, all limits specified for $T_J = 25^{\circ}C$, $V^+ = 2.7V$, $V^- = 0V$, $V_{CM} = V_0 = V^+/2$ and $R_L > 1 \text{ M}\Omega$. **Boldface** limits apply at the temperature extremes.

Symbol	Parameter	Conditions	Тур ⁽¹⁾	LMC7111AI Limit ⁽²⁾	LMC7111BI Limit ⁽²⁾	Units
SR	Slew Rate	See ⁽³⁾	0.015			V/µs
GBW	Gain-Bandwidth Product		40			kHz

(1) Typical Values represent the most likely parametric norm.

(2) All limits are specified by testing or statistical analysis.

(3) Connected as Voltage Follower with 1.0V step input. Number specified is the slower of the positive and negative slew rates. Input referred, V⁺ = 2.7V and R_L = 100 k Ω connected to 1.35V. Amp excited with 1 kHz to produce V_O = 1 V_{PP}.

3V DC Electrical Characteristics

Unless otherwise specified, all limits specified for $T_J = 25^{\circ}C$, $V^+ = 3V$, $V^- = 0V$, $V_{CM} = V_O = V^+/2$ and $R_L > 1 \text{ M}\Omega$. Boldface limits apply at the temperature extremes.

Symbol	Parameter	Conditions	Тур ⁽¹⁾	LMC7111AI Limit ⁽²⁾	LMC7111BI Limit ⁽²⁾	Units
V _{CM}	Input Common-Mode Voltage Range	V ⁺ = 3V For CMRR ≥ 50 dB	-0.25	0.0	0.0	V min
			3.2	3.0 2.8	3.0 2.8	V max

(1) Typical Values represent the most likely parametric norm.

(2) All limits are specified by testing or statistical analysis.



3.3V DC Electrical Characteristics

Unless otherwise specified, all limits specified for $T_J = 25^{\circ}C$, $V^+ = 3.3V$, $V^- = 0V$, $V_{CM} = V_O = V^+/2$ and $R_L > 1 \text{ M}\Omega$. **Boldface** limits apply at the temperature extremes.

Symbol	Parameter	Conditions	Тур ⁽¹⁾	LMC7111AI Limit ⁽²⁾	LMC7111BI Limit ⁽²⁾	Units
V _{CM}	Input Common-Mode Voltage Range	V ⁺ = 3.3V For CMRR ≥ 50 dB	-0.25	-0.1 0.00	-0.1 0.00	V min
			3.5	3.4 3.2	3.4 3.2	V max

(1) Typical Values represent the most likely parametric norm.

(2) All limits are specified by testing or statistical analysis.

5V DC Electrical Characteristics

Unless otherwise specified, all limits specified for $T_J = 25^{\circ}C$, $V^+ = 5V$, $V^- = 0V$, $V_{CM} = V_O = V^+/2$ and $R_L > 1 \text{ M}\Omega$. Boldface limits apply at the temperature extremes.

Symbol	Parameter	Conditions	Тур ⁽¹⁾	LMC7111AI Limit ⁽²⁾	LMC7111BI Limit ⁽²⁾	Units
V _{OS}	Input Offset Voltage	V ⁺ = 5V	0.9			mV max
TCV _{OS}	Input Offset Voltage Average Drift		2.0			µV/°C
IB	Input Bias Current	See ⁽³⁾	0.1	1 20	1 20	pA max
I _{OS}	Input Offset Current	See ⁽³⁾	0.01	0.5 10	0.5 10	pA max
R _{IN}	Input Resistance		>10			Tera Ω
CMRR	Common Mode Rejection Ratio	$0V \le V_{CM} \le 5V$	85	70	60	dB min
+PSRR	Positive Power Supply Rejection Ratio	$5V \le V^+ \le 10V$, $V^- = 0V$, $V_0 = 2.5V$	85	70	60	dB min
-PSRR	Negative Power Supply Rejection Ratio	$-5V \le V^- \le -10V,$ $V^- = 0V, V_0 = -2.5V$	85	70	60	dB min
V _{CM}	Input Common-Mode Voltage Range	V ⁺ = 5V For CMRR ≥ 50 dB	-0.3	-0.20 0.00	-0.20 0.00	V min
			5.25	5.20 5.00	5.20 5.00	V max
C _{IN}	Common-Mode Input Capacitance		3			pF
Vo	Output Swing	$V^{+} = 5V$	4.99	4.98	4.98	Vmin
		R _L = 100 kΩ	0.01	0.02	0.02	Vmax
		$V^{+} = 5V$	4.98	4.9	4.9	Vmin
		$R_L = 10 \ k\Omega$	0.02	0.1	0.1	Vmin
I _{SC}	Output Short Circuit Current	Sourcing, $V_0 = 0V$	7	5 3.5	5 3.5	mA min
		Sinking, $V_O = 3V$	7	5 3.5	5 3.5	mA min
A _{VOL}	Voltage Gain	Sourcing	500			V/mv min
		Sinking	200			V/mv min
I _S	Supply Current	$V^+ = +5V,$ $V_0 = V^+/2$	25			μA max

(1) Typical Values represent the most likely parametric norm.

(2) All limits are specified by testing or statistical analysis.

(3) Bias Current specified by design and processing.



5V AC Electrical Characteristics

Unless otherwise specified, all limits specified for $T_J = 25^{\circ}C$, $V^+ = 5V$, $V^- = 0V$, $V_{CM} = V_0 = V^+/2$ and $R_L > 1 \text{ M}\Omega$. **Boldface** limits apply at the temperature extremes.

Symbol	Parameter	Conditions	Тур ⁽¹⁾	LMC7111AI Limit ⁽²⁾	LMC7111BI Limit ⁽²⁾	Units
SR	Slew Rate	Positive Going Slew Rate ⁽³⁾	0.027	0.015	0.010	V/µs
GBW	Gain-Bandwidth Product		50			kHz

(1) Typical Values represent the most likely parametric norm.

(2) All limits are specified by testing or statistical analysis.

(3) Connected as Voltage Follower with 1.0V step input. Number specified is the slower of the positive slew rate. The negative slew rate is faster. Input referred, V⁺ = 5V and R_L = 100 k Ω connected to 1.5V. Amp excited with 1 kHz to produce V_O = 1 V_{PP}.

10V DC Electrical Characteristics

Unless otherwise specified, all limits specified for $T_J = 25^{\circ}$ C, $V^+ = 10$ V, $V^- = 0$ V, $V_{CM} = V_O = V^+/2$ and $R_L > 1$ M Ω . **Boldface** limits apply at the temperature extremes.

Symbol	Parameter	Conditions	Тур ⁽¹⁾	LMC7111AI Limit ⁽²⁾	LMC7111BI Limit ⁽²⁾	Units
V _{OS}	Input Offset Voltage	V ⁺ = 10V	0.9	3 5	7 9	mV max
TCV _{OS}	Input Offset Voltage Average Drift		2.0			µV/°C
IB	Input Bias Current		0.1	1 20	1 20	pA max
I _{OS}	Input Offset Current		0.01	0.5 10	0.5 10	pA max
R _{IN}	Input Resistance		>10			Tera Ω
+PSRR	Positive Power Supply Rejection Ratio	$5V \le V^+ \le 10V$, $V^- = 0V$, $V_0 = 2.5V$	80			dB min
-PSRR	Negative Power Supply Rejection Ratio	$-5V \le V^- \le -10V$, $V^- = 0V$, $V_0 = 2.5V$	80			dB min
V _{CM}	Input Common-Mode Voltage Range	V ⁺ = 10V For CMRR ≥ 50 dB	-0.2	-0.15 0.00	-0.15 0.00	V min
			10.2	10.15 10.00	10.15 10.00	V max
C _{IN}	Common-Mode Input Capacitance		3			pF
I _{SC}	Output Short Circuit Current ⁽³⁾	Sourcing, $V_0 = 0V$	30	20 7	20 7	mA min
		Sinking, $V_O = 10V$	30	20 7	20 7	mA min
A _{VOL}	Voltage Gain 100 kΩ Load	Sourcing	500			V/mv min
		Sinking	200			V/mv min
I _S	Supply Current	$V^+ = +10V,$ $V_0 = V^+/2$	25	50 65	60 75	μA max
Vo	Output Swing	V ⁺ = 10V	9.99	9.98	9.98	Vmin
		$R_L = 100 \text{ k}\Omega$	0.01	0.02	0.02	Vmax
		V ⁺ = 10V	9.98	9.9	9.9	Vmin
		$R_L = 10 \ k\Omega$	0.02	0.1	0.1	Vmin

(1) Typical Values represent the most likely parametric norm.

(2) All limits are specified by testing or statistical analysis.

(3) Bias Current specified by design and processing.

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10V AC Electrical Characteristics

Unless otherwise specified, all limits specified for $T_J = 25^{\circ}C$, $V^+ = 10V$, $V^- = 0V$, $V_{CM} = V_O = V^+/2$ and $R_L > 1 \text{ M}\Omega$. Boldface limits apply at the temperature extremes.

Symbol	Parameter	Conditions	Тур ⁽¹⁾	LMC7111AI Limit ⁽²⁾	LMC7111BI Limit ⁽²⁾	Units
SR	Slew Rate	See ⁽³⁾	0.03			V/µs
GBW	Gain-Bandwidth Product		50			kHz
φ _m	Phase Margin		50			deg
G _m	Gain Margin		15			dB
	Input-Referred Voltage Noise	f = 1 kHz V _{CM} = 1V	110			nV/ √Hz
	Input-Referred Current Noise	f = 1 kHz	0.03			pA/√Hz

(1) Typical Values represent the most likely parametric norm.

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 (2) All limits are specified by testing or statistical analysis.
 (3) Connected as Voltage Follower with 1.0V step input. Number specified is the slower of the positive and negative slew rates. Input referred, V⁺ = 10V and R_L = 100 kΩ connected to 5V. Amp excited with 1 kHz to produce V_O = 2 V_{PP}.

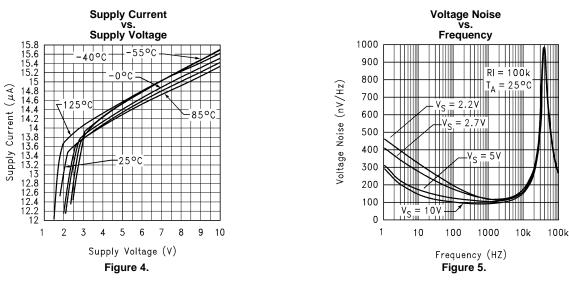


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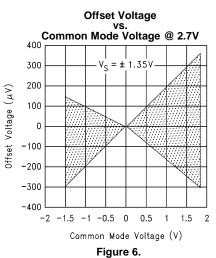
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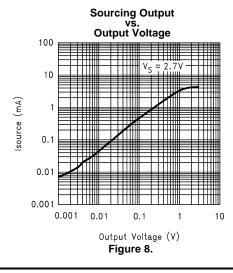
Typical Performance Characteristics

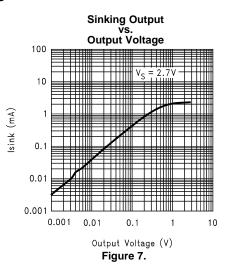
 T_{A} = 25°C unless specified, Single Supply



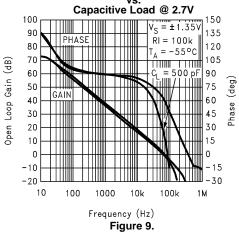
2.7V Performance







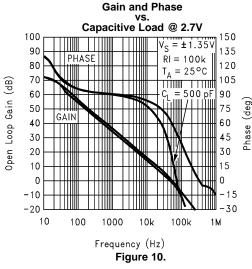
Gain and Phase

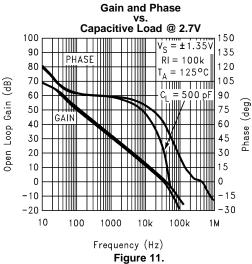


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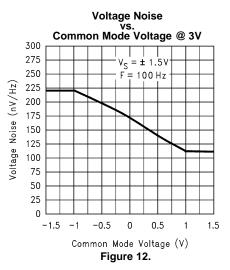


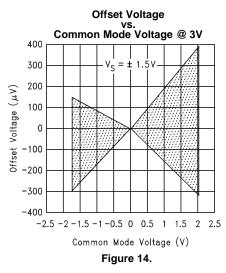


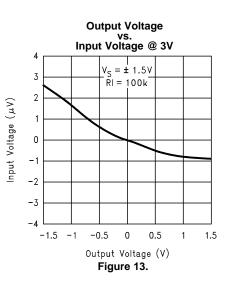




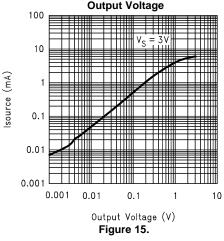
3V Performance







Sourcing Output vs. Output Voltage



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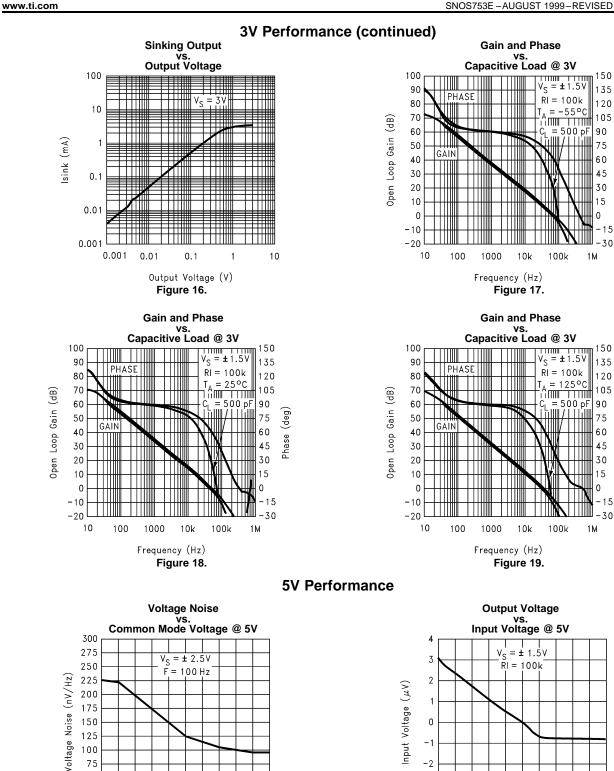


(deg)

Phase

(deg

Phase



50

25 0

-2.5

-1.5

-0.5

Common Mode Voltage (V)

Figure 20.

0.5

1.5

2.5

-4 -2.5 $-2^{-1.5}$ $-1^{-0.5}$ 0.5 1 1.5 2 2.5 1

Output Voltage (V)

Figure 21.

-3

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FEXAS NSTRUMENTS

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5V Performance (continued) **Offset Voltage** vs. Common Mode Voltage @ 5V 100 400 V_S = ± 2.5 v 300 10 200 Offset Voltage (μV) 100 lsource (mA) 1 0 0.1 -100 -200 0.01 -300 0.001 -400 - 4 -3 -2 - 1 0 1 2 3 4 0.001 0.01 0.1 Common Mode Voltage (V) Figure 22. **Sinking Output** vs. Output Voltage vs. 100 100 90 PHASE V, = 5V80 ТП 10 70 Open Loop Gain (dB) 60 lsink (mA) 1 50 40 30 0.1 20 10 0.01 0 -10 0.001 -20 0.001 0.01 0.1 1 10 10 100 1000 10k Output Voltage (V) Figure 24. Gain and Phase vs. Capacitive Load @ 5V 100 100 150 Ý_S = ±2.5V 90 135 90 RI = 100kHASE ASF 80 120 80 T¢ = 25°C 70 105 70 Open Loop Gain (dB) 60 90 60 00 р (deg) 50 75 50 GAIN GΔII 40 60 40 Phase 30 45 30 20 30 20 10 15 10 0 0 0 -10 -15 -10-20 -30 -20 10 100 1000 10k 100k 1M 10 100 1000 10k Frequency (Hz) Frequency (Hz) Figure 26.

Sourcing Output vs. Output Voltage V, = 5V 10 1 Output Voltage (V) Figure 23. Gain and Phase Capacitive Load @ 5V 150 $= \pm 2.5 V$ 135 $\overline{RI} = 100k$ 120 _ -55°C 105 90 4a 00 (deg 75 60 Phase 45 30 15 0 -15 -30 100k 1M Frequency (Hz) Figure 25. Gain and Phase vs. Capacitive Load @ 5V 150 $= \pm 2.5 V$ ۷_S 135 RI = 100k120 = 125°C 105 90 ÓÓ (deg) 75 60 Phase 45 30 15 0 -15

Figure 27.

Open Loop Gain (dB)

100k

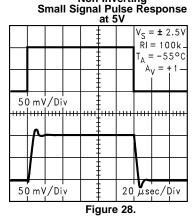
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1M



5V Performance (continued)



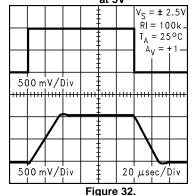


Non-Inverting Small Signal Pulse Response

					at 5	δV			
							V _S =	= ± 2 = 10	2.5V
							T _A =	= 12	5°C +1_
							A	v - 1 	_
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		****	+++++				++++		+++++
	T								
							T		
50	₽ DmV	/Di	V V			20	μ_{s}	ec/[Div
L				E 14		20			

Figure 30.

Non-Inverting Large Signal Pulse Response at 5V



Non-Inverting Small Signal Pulse Response at 5V

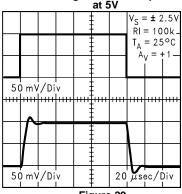
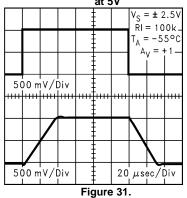
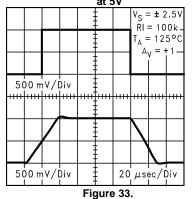


Figure 29.

Non-Inverting Large Signal Pulse Response at 5V

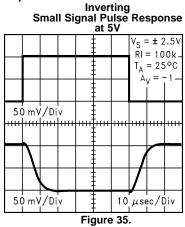


Non-Inverting Large Signal Pulse Response at 5V

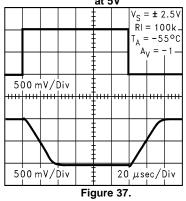




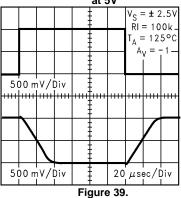
5V Performance (continued)



Inverting Large Signal Pulse Response at 5V

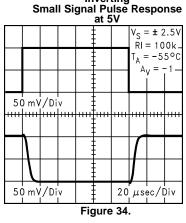








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Inverting Small Signal Pulse Response at 5V

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50	mV	/Di	v			20	μs	 ec/[l Div
Figure 36.									

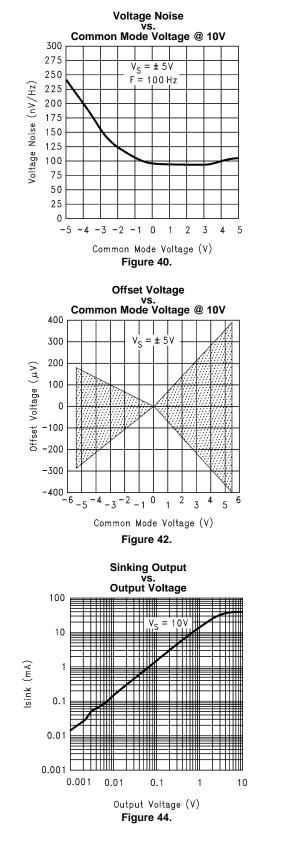
Inverting Large Signal Pulse Response at 5V

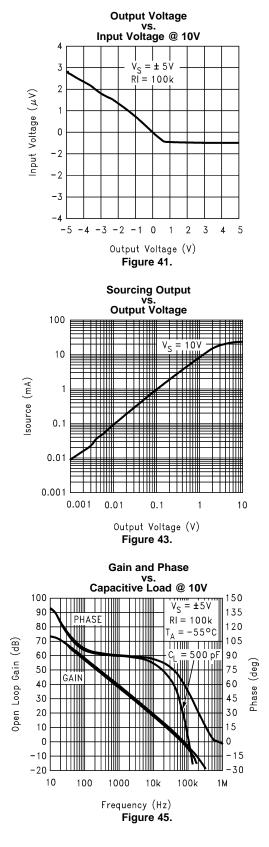
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	Figure 38.								



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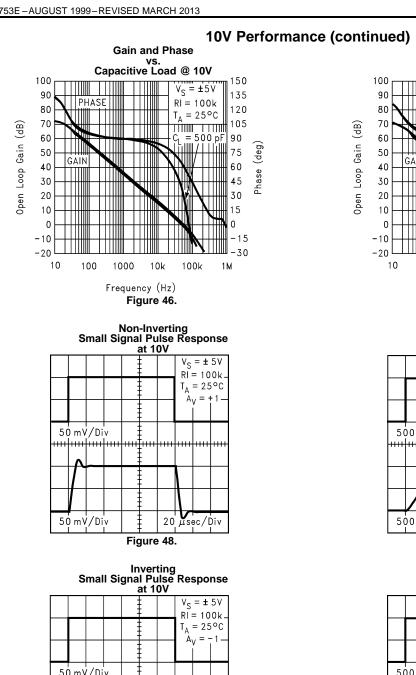




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vs. Capacitive Load @ 10V 150 $V_{S} = \pm 5V$ 135 PHASE RI = 100k120 $T_{A} = 125 °C$ 105 TIM 500 pF 90 (deg) 75 GAIN 60 Phase 45 30 15 Ш 0 -15 -30 1000 1E4 1E5 1E6 100 Frequency (Hz) Figure 47.

Gain and Phase



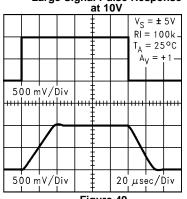


Figure 49.

Inverting Large Signal Pulse Response at 10V

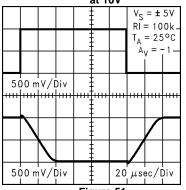


Figure 51.

50 mV/Div 50 mV/Div $10 \ \mu \text{sec}/\text{Div}$





APPLICATION INFORMATION

BENEFITS OF THE LMC7111 TINY AMP

Size

The small footprint of the SOT-23 packaged Tiny amp, (0.120 x 0.118 inches, 3.05 x 3.00 mm) saves space on printed circuit boards, and enable the design of smaller electronic products. Because they are easier to carry, many customers prefer smaller and lighter products.

Height

The height (0.056 inches, 1.43 mm) of the Tiny amp makes it possible to use it in PCMCIA type III cards.

Signal Integrity

Signals can pick up noise between the signal source and the amplifier. By using a physically smaller amplifier package, the Tiny amp can be placed closer to the signal source, reducing noise pickup and increasing signal integrity. The Tiny amp can also be placed next to the signal destination, such as a buffer for the reference of an analog to digital converter.

Simplified Board Layout

The Tiny amp can simplify board layout in several ways. First, by placing an amp where amps are needed, instead of routing signals to a dual or quad device, long pc traces may be avoided.

By using multiple Tiny amps instead of duals or quads, complex signal routing and possibly crosstalk can be reduced.

DIPs available for prototyping

LMC7111 amplifiers packaged in conventional 8-pin dip packages can be used for prototyping and evaluation without the need to use surface mounting in early project stages.

Low Supply Current

The typical 25 µA supply current of the LMC7111 extends battery life in portable applications, and may allow the reduction of the size of batteries in some applications.

Wide Voltage Range

The LMC7111 is characterized at 2.7V, 3V, 3.3V, 5V and 10V. Performance data is provided at these popular voltages. This wide voltage range makes the LMC7111 a good choice for devices where the voltage may vary over the life of the batteries.

INPUT COMMON MODE VOLTAGE RANGE

The LMC7111 does not exhibit phase inversion when an input voltage exceeds the negative supply voltage.

The absolute maximum input voltage is 300 mV beyond either rail at room temperature. Voltages greatly exceeding this maximum rating can cause excessive current to flow in or out of the input pins, adversely affecting reliability.

Applications that exceed this rating must externally limit the maximum input current to ± 5 mA with an input resistor as shown in Figure 52.

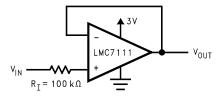


Figure 52. R_I Input Current Protection for Voltages Exceeding the Supply Voltage

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CAPACITIVE LOAD TOLERANCE

RESISTORS

or $R_1 C_{IN} \leq R_2 C_f$

which typically provides significant overcompensation.

Printed circuit board stray capacitance may be larger or smaller than that of a breadboard, so the actual optimum value for C_F may be different. The values of C_F should be checked on the actual circuit. (Refer to the LMC660 quad CMOS amplifier data sheet for a more detailed discussion.)



OUTPUT SWING

The output of the LMC7111 will go to within 100 mV of either power supply rail for a 10 k Ω load and to 20 mV of the rail for a 100 k Ω load. This makes the LMC7111 useful for driving transistors which are connected to the same power supply. By going very close to the supply, the LMC7111 can turn the transistors all the way on or all the way off.

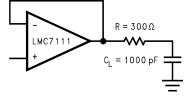


Figure 53. Resistive Isolation of a 330 pF Capacitive Load

When using very large value feedback resistors, (usually > 500 k Ω) the large feed back resistance can react with the input capacitance due to transducers, photodiodes, and circuit board parasitics to reduce phase margins. The effect of input capacitance can be compensated for by adding a feedback capacitor. The feedback capacitor

COMPENSATING FOR INPUT CAPACITANCE WHEN USING LARGE VALUE FEEDBACK

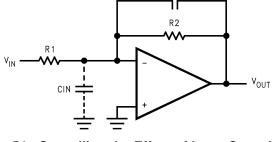
gain follower is the most sensitive configuration. Direct capacitive loading reduces the phase margin of op-amps. The combination of the op-amp's output impedance and the capacitive load induces phase lag. This results in either an underdamped pulse response or oscillation.

Capacitive load compensation can be accomplished using resistive isolation as shown in Figure 53. This simple technique is useful for isolating the capacitive input of multiplexers and A/D converters.

The LMC7111 can typically directly drive a 300 pF load with $V_{\rm S}$ = 10V at unity gain without oscillating. The unity

(as in Figure 54), C_f is first estimated by: $\frac{1}{2\pi R_1 C_{\rm IN}} \geq \frac{1}{2\pi R_2 C_{\rm f}}$

Cf



(2)

(1)



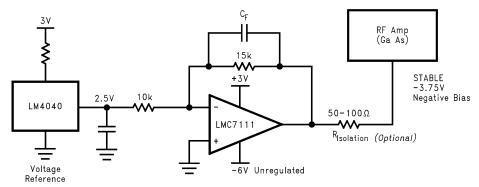
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BIASING GaAs RF AMPLIFIERS

The capacitive load capability, low current draw, and small size of the SOT-23 LMC7111 make it a good choice for providing a stable negative bias to other integrated circuits.

The very small size of the LMC7111 and the LM4040 reference take up very little board space.



C_F and R_{isolation} prevent oscillations when driving capacitive loads.

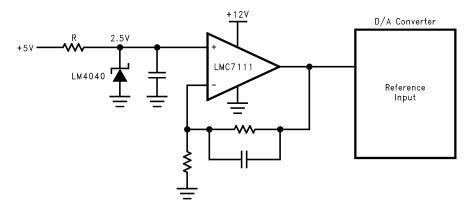
Figure 55. Stable Negative Bias

REFERENCE BUFFER FOR A-TO-D CONVERTERS

The LMC7111 can be used as a voltage reference buffer for analog-to-digital converters. This works best for Ato-D converters whose reference input is a static load, such as dual slope integrating A-to-Ds. Converters whose reference input is a dynamic load (the reference current changes with time) may need a faster device, such as the LMC7101 or the LMC7131.

The small size of the LMC7111 allows it to be placed close to the reference input. The low supply current (25 µA typical) saves power.

For A-to-D reference inputs which require higher accuracy and lower offset voltage, please see the LMC6462 datasheet. The LMC6462 has performance similar to the LMC7111. The LMC6462 is available in two grades with reduced input voltage offset.



DUAL AND QUAD DEVICES WITH SIMILAR PERFORMANCE

The LMC6462 and LMC6464 are dual and quad devices with performance similar to the LMC7111. They are available in both conventional through-hole and surface mount packaging. Please see the LMC6462/4 datasheet for details.

SPICE MACROMODEL

A SPICE macromodel is available for the LMC7111. This model includes simulation of:

- Input common-mode voltage range
- Frequency and transient response

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- · Quiescent and dynamic supply current
- Output swing dependence on loading conditions and many more characteristics as listed on the macro model disk. Visit the LMC7111 product page on http://www.ti.com for the spice model.

ADDITIONAL SOT-23 TINY DEVICES

Additional parts are available in the space saving SOT-23 Tiny package, including amplifiers, voltage references, and voltage regulators. These devices include—

- LMC7101 1 MHz gain-bandwidth rail-to-rail input and output amplifier—high input impedance and high gain, 700 µA typical current 2.7V, 3V, 5V and 15V specifications.
- LM7131 Tiny Video amp with 70 MHz gain bandwidth. Specified at 3V, 5V and ± 5V supplies.
- LMC7211 Comparator in a tiny package with rail-to-rail input and push-pull output. Typical supply current of 7 µA. Typical propagation delay of 7 µs. Specified at 2.7V, 5V and 15V supplies.
- **LMC7221** Comparator with an open drain output for use in mixed voltage systems. Similar to the LMC7211, except the output can be used with a pull-up resistor to a voltage different than the supply voltage.
- LP2980 Micropower SOT 50 mA Ultra Low-Dropout Regulator.
- LM4040 Precision micropower shunt voltage reference. Fixed voltages of 2.5000V, 4.096V, 5.000V, 8.192V and 10.000V.
- LM4041 Precision micropower shunt voltage reference 1.225V and adjustable.

Visit http://www.ti.com for more information.



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REVISION HISTORY

Ch	hanges from Revision D (March 2013) to Revision E	Page
•	Changed layout of National Data Sheet to TI format	18



PACKAGING INFORMATION

Orderable Device	Status	Package Type	•	Pins	Package	Eco Plan	Lead/Ball Finish	MSL Peak Temp	Op Temp (°C)	Top-Side Markings	Samples
	(1)		Drawing		Qty	(2)		(3)		(4)	
LMC7111BIM5	ACTIVE	SOT-23	DBV	5	1000	TBD	Call TI	Call TI	-40 to 85	A01B	Samples
LMC7111BIM5/NOPB	ACTIVE	SOT-23	DBV	5	1000	Green (RoHS & no Sb/Br)	CU SN	Level-1-260C-UNLIM	-40 to 85	A01B	Samples
LMC7111BIM5X	ACTIVE	SOT-23	DBV	5	3000	TBD	Call TI	Call TI	-40 to 85	A01B	Samples
LMC7111BIM5X/NOPB	ACTIVE	SOT-23	DBV	5	3000	Green (RoHS & no Sb/Br)	CU SN	Level-1-260C-UNLIM	-40 to 85	A01B	Samples

⁽¹⁾ The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

⁽²⁾ Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

Pb-Free (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

⁽³⁾ MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

⁽⁴⁾ Multiple Top-Side Markings will be inside parentheses. Only one Top-Side Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Top-Side Marking for that device.

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PACKAGE OPTION ADDENDUM

11-Apr-2013

PACKAGE MATERIALS INFORMATION

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TAPE AND REEL INFORMATION





QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



All dimensions are nominal												
Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
LMC7111BIM5	SOT-23	DBV	5	1000	178.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
LMC7111BIM5/NOPB	SOT-23	DBV	5	1000	178.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
LMC7111BIM5X	SOT-23	DBV	5	3000	178.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3
LMC7111BIM5X/NOPB	SOT-23	DBV	5	3000	178.0	8.4	3.2	3.2	1.4	4.0	8.0	Q3

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PACKAGE MATERIALS INFORMATION

26-Mar-2013



*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
LMC7111BIM5	SOT-23	DBV	5	1000	210.0	185.0	35.0
LMC7111BIM5/NOPB	SOT-23	DBV	5	1000	210.0	185.0	35.0
LMC7111BIM5X	SOT-23	DBV	5	3000	210.0	185.0	35.0
LMC7111BIM5X/NOPB	SOT-23	DBV	5	3000	210.0	185.0	35.0

DBV (R-PDSO-G5)

PLASTIC SMALL-OUTLINE PACKAGE



NOTES: A. All linear dimensions are in millimeters.

B. This drawing is subject to change without notice.

C. Body dimensions do not include mold flash or protrusion. Mold flash and protrusion shall not exceed 0.15 per side.

D. Falls within JEDEC MO-178 Variation AA.



DBV (R-PDSO-G5)

PLASTIC SMALL OUTLINE



NOTES:

A. All linear dimensions are in millimeters.B. This drawing is subject to change without notice.

- C. Customers should place a note on the circuit board fabrication drawing not to alter the center solder mask defined pad.
- D. Publication IPC-7351 is recommended for alternate designs.
- E. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Example stencil design based on a 50% volumetric metal load solder paste. Refer to IPC-7525 for other stencil recommendations.



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